

DC circuit

Introduction :-

Q. What is electrical engineering?

Ans. Electrical engineering is a branch of engineering which deals with the aspect of electrical energy, its system, component & devices. Every equipment & component of an electrical system can be represented by an equivalent electrical network which consists of a very few elements such as energy source, resistor, capacitor & inductors.

* Electric Current $\Rightarrow$ It is the rate of flow of charges w.r.t. time.

$$I = \frac{dQ}{dt}$$

* Electric potential $\Rightarrow$ It is the amount of work done required to move a unit positive charge b/w two points.

$$V = \frac{W}{q} \quad (\text{Unit - Volts})$$

* Electro motive force $\Rightarrow$ It is the force or voltage that drives and electric current to flow in an electric circuit.

Electric power $\Rightarrow$ It is the rate of doing work per unit time.

$$P = \frac{W}{t} = VI = I^2 R$$

★ Ohm's Law :- A current flowing through the conductor is directly proportional to the potential applied across its ends with the physical condition that is temperature, pressure, Resistance etc. $I \propto V$

$$V = IR$$

Resistance :- It is the properties of the conductor by which it oppose the flow of current.

$$R = \frac{\rho l}{A}$$

Inductance :- It is the properties of the conductor by which change in the current throw it induces & EMF.

$$V = L \frac{di}{dt}$$

$$L = \frac{\mu N^2 A}{l}$$

Capacitance :- It is the ability of a conductor to store electric charge

$$C = \frac{Q}{V}$$

$$\text{Unit} - \frac{C}{V}$$

Basic circuit elements :-

- 1.) Active elements $\Rightarrow$ The circuit elements which supply energy to the circuit is known as Active elements.
eg - Battery, generator, cells etc.

2/ Passive element $\Rightarrow$ The elements which either dissipate or store energy are known as passive.

eg- Resistor, capacitor etc.

3/ linear element $\Rightarrow$ The elements whose output is directly proportional to the input are said to be linear element.

eg- pure resistor, inductor, capacitor etc.

4/ Non linear element $\Rightarrow$ The elements which are not linear are called non-linear.

eg- ~~Diode~~ Diode.

5/ Bilateral element $\Rightarrow$ Elements whose conduction of current are in both the directions.

eg- resistor, capacitor, inductor.

6/ Unilateral element $\Rightarrow$ Elements whose conduction of current is only one direction.

eg- Diode, transistor etc.

7/ Lumped element $\Rightarrow$ Elements which are pure and physically separable.

eg- pure resistor, inductor, capacitor etc.

8/ distributed element $\Rightarrow$ Distributed element are the elements which are not pure and can not be separated physically.

eg- transmission line.

* Faraday's first law $\Rightarrow$ When even the conductor line and emf is induced in conductor.

* Faraday's second law $\Rightarrow$ The magnitude of induced emf in a conductor is directly proportional to the rate of change of flux linkage.

* Lenz's Law $\Rightarrow$ According to this law and induced emf will give rise to a ~~current~~ current whose magnetic field opposes the change in original magnetic field.

Energy Source :- Energy sources may be some time called as network sources as these sources provide necessary excitation to an electric network. There are two types of energy source :-
(i) Voltage Source
(ii) Current Source

Further these sources classified as ideal and practical and dependent and independent

1) Ideal Voltage Source :- It is an energy source whose terminal voltage is constant and independent of current through it.

2) Ideal current source :- It is energy source which can supply a constant magnitude of current irrespective of the voltage across its terminal.

3) Practical Voltage source :- It is can be represented by an ideal voltage in series with an internal resistance r_i . This resistance accounts for the drop in voltage across its terminals.

4) Practical Current source :- It is can be represented by an ideal current source in parallel with an internal resistance r_i .

5) dependent energy source :- Dependent energy source are the voltage and current source whose output depends on the voltage or current of self certain elements of the network in which they are connected.

* Network Analysis $\Rightarrow$

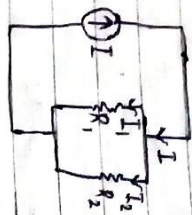
Voltage Divider Rule :-



$$V_1 = \frac{V \times R_1}{R_1 + R_2}$$

$$V_2 = \frac{V \times R_2}{R_1 + R_2}$$

Current Divider Rule :-



$$I_1 = I \times \frac{R_2}{R_1 + R_2}$$

$$I_2 = I \times \frac{R_1}{R_1 + R_2}$$

Kirchoff's Law :-

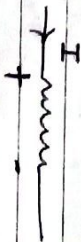
KCL :-



According to this law in an electric circuit the algebraic sum of all the current meeting at a point is equal to zero.

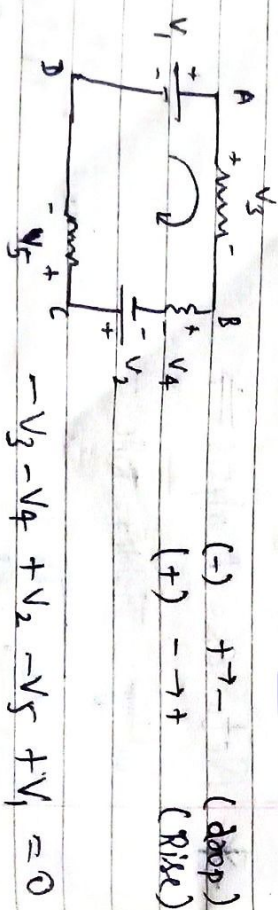
$$I_1 - I_2 + I_3 - I_4 + I_5 = 0$$

$$I_1 + I_3 + I_5 = I_2 + I_4$$

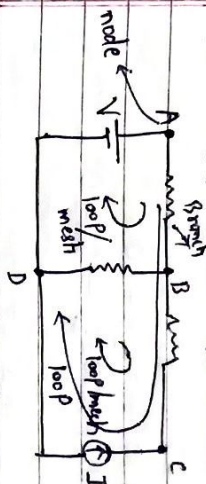


KVL :-

According to this law the algebraic sum of all the voltages in a closed path is equal to zero.



NETWORK ANALYSIS :-

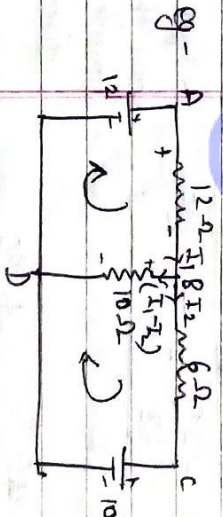


(i) NODE :- A node is the point at which two or more circuit elements are interconnected.

(ii) Branch :- A branch is a part of a network which lie b/w two points or nodes.

(iii) loop :- A loop in a network is any closed conducting path through two or more elements of the network.

(iv) mesh :- A mesh is the most elementary form of a loop which can not be further divided.



Apply KCL at node B
 $I_1 = I_2 + I_3$
 $I_3 = I_1 - I_2$

Apply KVL in ABDA loop :-

$$12 - I_1 \times 12 - (I_1 - I_2) \times 10 = 0$$

$$12 - 22I_1 + I_2 \times 10 = 0$$

$$10I_2 + 12 = 22I_1$$

$$5I_2 + 6 = 11I_1 \quad \text{--- (1)}$$

Apply KVL in BCDA

$$-6I_2 - 10 + 10(I_1 - I_2) = 0$$

$$-6I_2 + 10I_1 - 10I_2 - 10 = 0$$

$$10I_1 = 10 + 16I_2 \quad \text{--- (2)}$$

$$5I_1 = 5 + 8I_2 \quad \text{--- (1)}$$

from eqn (1) $(5I_2 + 6 = 11I_1)$

$$25I_1 = 25 + 40I_2$$

$$88I_1 = 42 + 40I_2$$

$$63I_1 = 83$$

$$I_1 = \frac{17}{63} \quad \boxed{I_1 = 0.365}$$

in eqn (1)

$$5I_2 + 6 = \frac{11 \times 23}{63} = \frac{253}{63}$$

$$5I_2 = \frac{253}{63} - \frac{378}{63} = -\frac{125}{63}$$

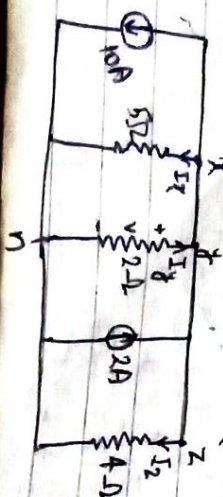
$$I_2 = \frac{-25}{63} = -0.397 \quad \text{(- sign is opposite direction)}$$

show that

$$V_1 = I_1 \times 12 = 12 \times 0.365 = 4.380$$

$$V_2 = I_2 \times 6 = 2.382$$

$$V_3 = 10 \times (I_1 - I_2) = 10(0.762) = 7.62$$



Apply KCL at node y

$$I_y - I_x - 10 + 2 - I_z = 0$$

$$-I_x + I_y - I_z = 8$$

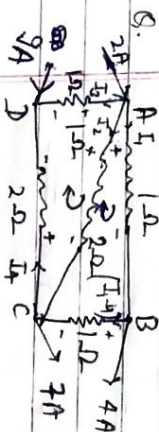
$$\frac{-V}{5} - \frac{V}{2} - \frac{V}{4} = 8$$

$$\frac{-V}{5} - \frac{3V}{4} = \frac{-4V - 15V}{20} = 8 \Rightarrow \frac{-19V}{20} = 8$$

$$-19V = 160 \Rightarrow V = \frac{-160}{19} = -8.42$$

$$I_x = \frac{-8.42}{5} = -1.684$$

$$I_y = -\left(\frac{-8.42}{2}\right) = 4.21, \quad I_z = \frac{-8.42}{4} = -2.105$$



Apply KCL at node A

$$I_1 + I_2 + I_3 = 2$$

$$I_3 = 2 - I_1 - I_2 \quad \text{--- (1)}$$

$$I_1 - 4 = I_5$$

Apply KCL at node C

$$I_2 + I_1 - 4 = 7 + I_4$$

$$I_4 = I_1 + I_2 - 11 \quad \text{--- (2)}$$

loop Apply KVL in loop ABCA

$$-I_1 \times 1 - 1 \times (I_1 - 4) + 2 \times I_2$$

$$-I_1 - I_1 + 4 + 2I_2$$

$$2I_2 - 2I_1 + 4 = 0$$

$$I_1 - I_2 = 2 \quad \text{--- (3)}$$

loop KVL in loop ACDA

$$I_3 \times 1 - 2 \times I_2 - (I_1 + I_2 - 11) \times 2 = 0$$

$$I_3 - 2I_1 - 2I_2 + 2I_2$$

$$I_3 - 4I_1 - 8I_2 + 2I_2$$

$$I_3 + 2I_2 = 4I_1 + 2I_2$$

$$2 - I_1 - I_2 + 2I_2 = 4I_1 + 2I_2$$

$$3I_2 + 5I_1 = 26 \quad (4)$$

$$(I_1 = I_2 + 2A)$$

$$3I_2 + 5I_2 + 10 = 26$$

$$8I_2 = 16 \Rightarrow I_2 = 2A$$

$$I_1 = 4A$$

$$I_3 = 2 - 2 - 4 \Rightarrow I_3 = -4A$$

$$I_4 = I_1 + I_2 - 11 = -5A = I_4$$

$$I_5 = I_1 - 4 = 0 = I_5$$

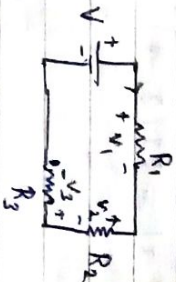
Voltage - current source change $\Rightarrow$



Series Parallel network:-

Series network $\Rightarrow$

$$V = V_1 + V_2 + V_3$$



$$V = I(R_1 + R_2 + R_3)$$

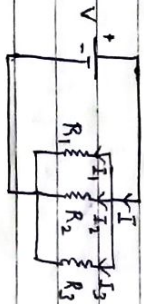
$$\Rightarrow V = I R_{eq}$$

$$I = \frac{V}{R_{eq}}$$

$$V_1 = I R_1 = \frac{V}{R_{eq}} \times R_1 = V_1$$

$$V_2 = \frac{V}{R_{eq}} \times R_2$$

Parallel network $\Rightarrow$



$$I = I_1 + I_2 + I_3$$

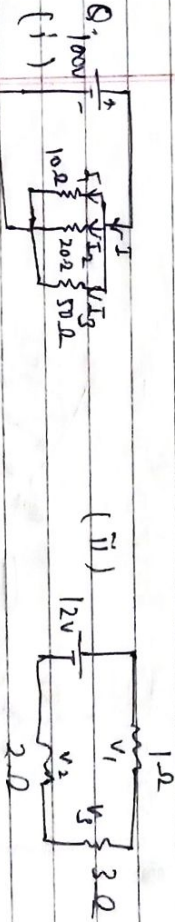
$$I = V \left[\frac{1}{R_1} + \frac{1}{R_2} + \frac{1}{R_3} \right] \Rightarrow \frac{1}{R_{eq}} = \frac{1}{R_1} + \frac{1}{R_2} + \frac{1}{R_3}$$

$$R_{eq} = G_1 + G_2 + G_3$$

$$I_1 = \frac{V}{R_1} = V \times G_1 \Rightarrow I_1 = \frac{V}{R_{eq}} \times G_1$$

$$I_2 = \frac{V}{R_2} \times G_2$$

$$I_3 = \frac{V}{R_3} \times G_3$$



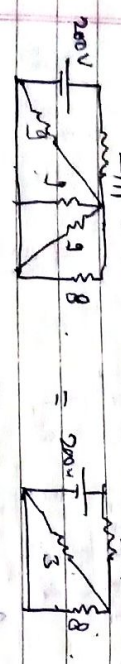
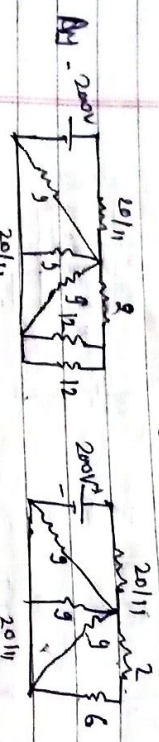
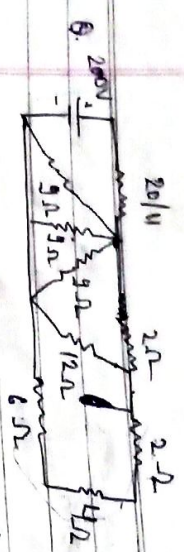
Q. (ii) $V_1 = \frac{V}{R_1 + R_2 + R_3} \times R_1 = \frac{12 \times 1}{6} = 2 \text{ Volt}$

$$V_2 = \frac{12}{6} \times 3 = 6 \text{ Volt}$$

$$V_3 = \frac{12}{6} \times 2 = 4 \text{ Volt}$$

(iii) $I_1 = \frac{I \times G_1}{G_{eq}} = \frac{100 \times \frac{1}{10}}{10} = 10A$

$$I_2 = \frac{100 \times \frac{1}{20}}{20} = 5A, \quad I_3 = \frac{100 \times \frac{1}{20}}{20} = 2A$$



$$\frac{1}{3} + \frac{1}{6} = \frac{2+3}{24} = \frac{11}{24}$$

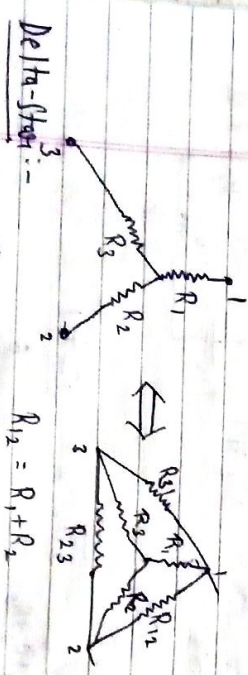
$$\frac{24}{11} + \frac{20}{11} = \frac{44}{11} = 4$$

$$R_{eq} = 4\Omega$$

$$V = IR = 4 \times 20 = 200$$

$$I = 50A$$

Star-Delta Transformation :-



$$R_{12} = R_{12} \parallel (R_{23} + R_{31})$$

$$R_1 + R_2 = \frac{R_{12} (R_{23} + R_{31})}{R_{12} + R_{23} + R_{31}}$$

$$R_2 + R_3 = \frac{R_{23} (R_{31} + R_{12})}{R_{12} + R_{23} + R_{31}}$$

$$R_3 + R_1 = \frac{R_{31} (R_{12} + R_{23})}{R_{12} + R_{23} + R_{31}}$$

Adding - ①, ② & ③

$$R_1 + R_2 + R_3 = \frac{R_{12} R_{23} + R_{23} R_{31} + R_{31} R_{12}}{R_{12} + R_{23} + R_{31}} \quad (4)$$

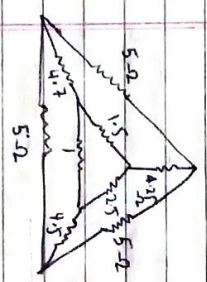
$$R_1 = \frac{R_{12} R_{31}}{\sum R_{eq}}, \quad R_2 = \frac{R_{12} R_{23}}{\sum R_{eq}}, \quad R_3 = \frac{R_{23} R_{31}}{\sum R_{eq}}$$

Star-Delta :-

$$R_{12} = R_1 + R_2 + \frac{R_1 R_2}{R_3}$$

$$R_{23} = R_2 + R_3 + \frac{R_2 R_3}{R_1}$$

$$R_{31} = R_1 + R_3 + \frac{R_1 R_3}{R_2}$$

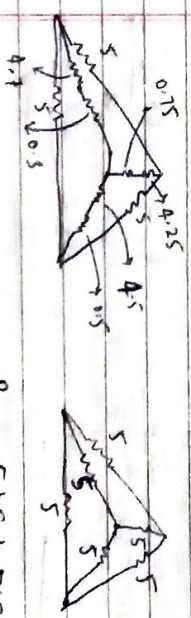


$$R_{12} = 2.5, \quad R_{23} = 1, \quad R_{31} = 15$$

$$R_1 = \frac{2.5 \times 15}{\sum R_{eq}}$$

$$R_1 = \frac{2.5 \times 15}{5} = 0.75$$

$$R_2 = \frac{2.5}{5} = 0.5, \quad R_3 = \frac{15}{5} = 3 = 0.3\Omega$$



$$R_{12} = 5 + 5 + \frac{5 \times 5}{5} = 15$$

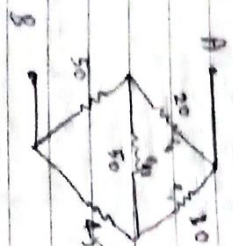
$$R_1 = \frac{1}{\frac{1}{5} + \frac{1}{5}} = \frac{3+1}{15} = \frac{4}{15} \Rightarrow R_1 = \frac{15}{4} = 3.75$$

$$R_{eq} = \frac{1}{\frac{1}{15} + \frac{1}{5} + \frac{1}{7.5}} = \frac{5+15+10}{25} = \frac{30}{25} = 1.2$$



$$R_{eq} = \frac{75}{30} = 2.5 \Omega$$

8.

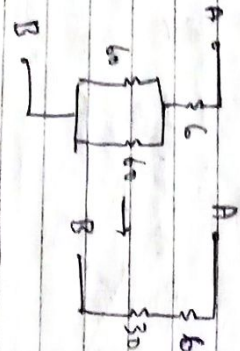
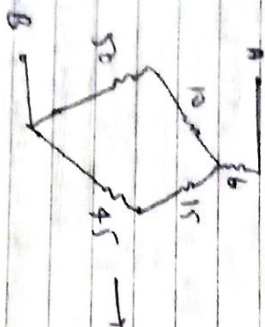


$R_{AB} = ?$

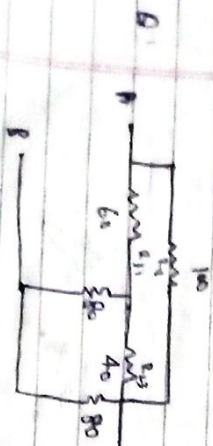
$$R_1 = \frac{30 \times 20}{100} = 6$$

$$R_2 = \frac{50 \times 30}{100} = 15$$

$$R_3 = \frac{50 \times 20}{100} = 10$$



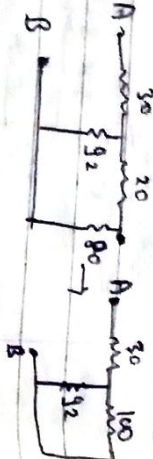
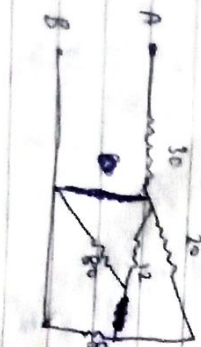
$$R_{eq} = 36 \Omega$$



$$R_1 = \frac{100 \times 60}{200} = 30$$

$$R_2 = \frac{40 \times 100}{200} = 20$$

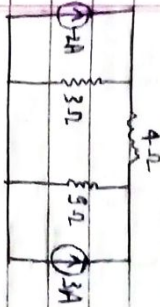
$$R_3 = \frac{60 \times 20}{80} = 15$$



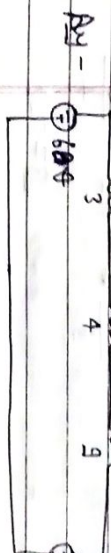
$$R_{eq} = \frac{30 + 50}{12} = \frac{80}{12} = 6.67 \Omega$$

$$R_{eq} = \frac{30 + 50}{12} = \frac{80}{12} = 6.67 \Omega$$

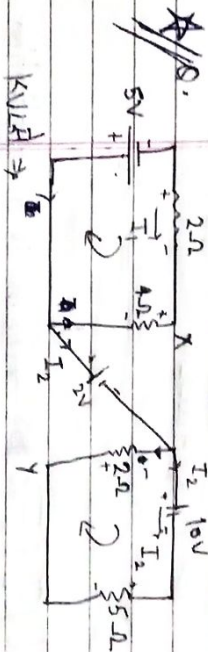
9.



find current through 4 ohm resistor using source conversion.



$$I = \frac{30}{16}$$



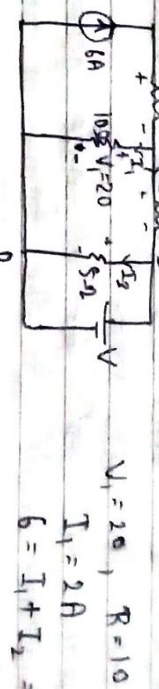
find $V_{xy} = ?$

$$-2I_1 - 4I_1 - 5 = 0 \Rightarrow 6I_1 = -5 \Rightarrow I_1 = -5/6$$

$$-10 - 5I_2 - 2I_2 = 0 \Rightarrow I_2 = -10/7$$

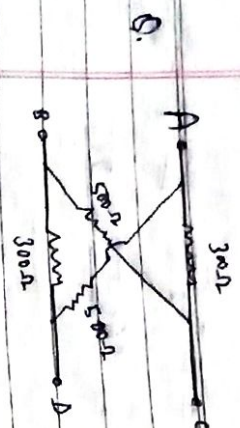
$$4x - (-5) - 2 + 1(-10/7) = \frac{20}{6} - \frac{24}{7} = \frac{140 - 240}{42} = \frac{-100}{42}$$

$$= \frac{-100}{42} = -2.38$$



$$V_1 = 20, R = 10$$

$$-2I_2 - 10V_{BD} + 20 = 0 \Rightarrow V_{BD} = 20 - 10I_2 \Rightarrow V_{BD} = 12V$$



Case-I when C-D are short.
Case-II when C-D are open.

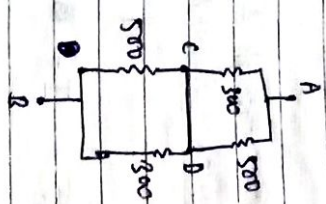
At- (I)

$$R = \frac{1}{\frac{1}{300} + \frac{1}{300} + \frac{1}{300}} = \frac{800}{3}$$

$$R = \frac{1500}{8}$$

$$R = \frac{1500}{8} \times 2 = \frac{1500}{4}$$

$$375 \Omega$$



(II)

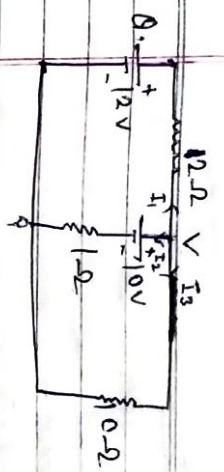
$$R = 400 \Omega$$

NODAL ANALYSIS $\Rightarrow$ This theorem is a application of KCL and computes all the node voltages.

STEP to solve the circuit using nodal analysis:-

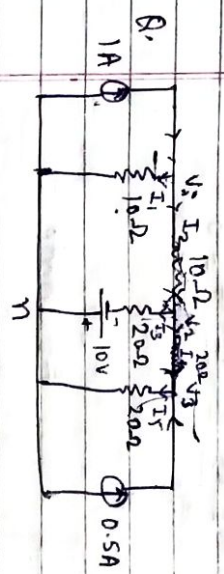
- (i) Identify all the nodes in the network and assume one node as a reference node.
- (ii) Assume node voltages at each node accept the reference node because voltage of reference node is zero volt.
- (iii) Right KCL eqⁿ for all node accept the reference node in the form of node voltage and solve them.

'n' node eqⁿ (n-1 eqⁿ)



16V = 160

V = 10 Volt



Solve $\Rightarrow$ KCL Rule:- (At node 1)

1 = $\frac{V_1}{10} + \frac{V_1 - V_2}{10}$
 $2V_1 - V_2 = 10$ — (1)

At node 2

$I_3 + I_4 = I_2$
 $\frac{V_2 + 10}{20} + \frac{V_2 - V_3}{20} = \frac{V_1 - V_2}{10}$
 $4V_2 + 10 = 2V_1 + V_3$
 $8V_1 - 40 + 10 = 2V_1 + V_3$

At node 3

$I_4 + 0.5 = I_5$
 $\frac{V_2 - V_3}{20} + 0.5 = \frac{V_3}{20}$
 $6V_1 = V_3 + 30$ — (2)

$V_2 - V_3 + 10 = V_3$

$V_2 + 10 = 2V_3$ — (3)

$2V_1 - 10 + 10 = 2V_3$

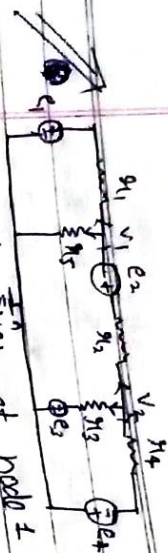
$6V_3 = V_3 + 30$

$V_3 = 6$

$V_1 = V_3$

$V_1 = V_3 = 6 \text{ Volt}$

$V_2 = 2 \text{ Volt}$



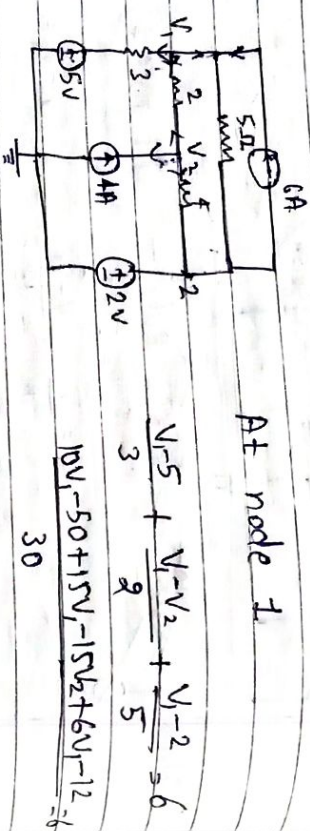
Apply KCL at node 1

$$\frac{V_1 - E_1}{R_1} + \frac{V_1}{R_1} + \frac{V_1 + E_2 - V_2}{R_2} = 0$$

At node 2

$$\frac{V_2 - E_3 - V_1}{R_3} + \frac{V_2 - E_4}{R_4} + \frac{V_2 + E_4}{R_5} = 0$$

Q.



At node 1

$$\frac{V_1 - 5}{3} + \frac{V_1 - V_2}{2} + \frac{V_1 - 2}{5} = 0$$

$$3V_1 - 15V_2 = 180 + 50 + 12$$

$$3V_1 - 15V_2 = 242 \text{ --- (1)}$$

At node 2

$$\frac{V_2 - V_1}{2} + \frac{V_2 - 2}{4} = 4$$

$$\frac{2V_2 - 2V_1 + V_2 - 2}{4} = 4 \Rightarrow 3V_2 - 2V_1 = 18 \text{ --- (2)}$$

$$V_2 = \frac{18 + 2V_1}{3}$$

$$3V_1 - 15 \left(\frac{18 + 2V_1}{3} \right) = 242$$

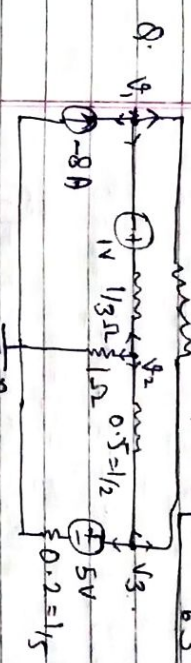
$$3V_1 - 90 - 10V_1 = 242$$

$$2V_1 = 332$$

$$V_1 = \frac{332}{2} = 166$$

$$3V_2 - \frac{664}{2} = 18 \Rightarrow 3V_2 = 18 + \frac{664}{2}$$

$$V_2 = \frac{1092}{6.3} = 16.5 \text{ V or H}_3$$



At node 1

$$\frac{V_1 + 1 - V_2}{1/3} + \frac{V_1 - V_3}{1/3} + 8 = 0$$

$$3V_1 + 3 - 3V_2 + 4V_1 - 4V_3 + 8 = 0$$

$$7V_1 - 3V_2 - 4V_3 + 11 = 0 \text{ --- (1)}$$

At node 2: ---

$$\frac{V_2 - 1 - V_1}{1/3} + \frac{V_2}{1} + \frac{V_2 - V_3}{1/2} = 0 \Rightarrow 3V_2 - 3 - 3V_1 + 2V_2 - V_3 = 0$$

At node 3: ---

$$\frac{V_3 - V_1}{1/4} + \frac{V_3 - V_2}{1/2} + \frac{V_3 - 5}{1/5} = 0$$

$$4V_3 - 4V_1 + 2V_3 - 2V_2 + 5V_3 - 25 = 0$$

$$11V_3 - 4V_1 - 2V_2 = 25 \text{ --- (3)}$$

$$3V_1 + 2V_3 + 3 = 6V_2$$

$$7V_1 - 3 \left(\frac{3V_1 + 2V_3 + 3}{2} \right) - 4V_3 + 11 = 0$$

$$7V_1 - (3V_1 + 2V_3 + 3) - 4V_3 + 11 = 0$$

$$14V_1 - 3V_1 - 2V_3 - 3 - 8V_3 + 22 = 0$$

$$11V_1 - 10V_3 = -19 \text{ --- (4)}$$

from eqn (3)
 $11V_3 - 4V_1 - 25 = 2V_3$

$$7V_1 - 3\left(\frac{11V_3 - 4V_1 - 25}{2}\right) - 4V_3 + 11 = 0$$

$$14V_1 - 33V_3 + 12V_1 + 75 - 8V_3 + 22 = 0$$

$$26V_1 - 41V_3 = -97$$

$$41V_3 - 26V_1 = 97 \quad (5)$$

(eqn-4)

$$\begin{array}{r} 410V_3 - 260V_1 = 970 \\ -410V_3 + 451V_1 = -779 \\ \hline 191V_1 = 191 \end{array}$$

$$V_1 = 1$$

$$V_3 = \frac{-19 - 11}{-10} = 3 \quad V_3 = 3$$

$$\begin{array}{r} 41 \\ 3 \\ \hline 163 \\ 121 \\ \hline 41 \end{array}$$

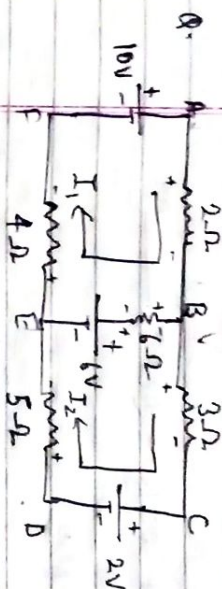
$$7 - 3V_2 - 12 + 11 = 0$$

$$V_2 = 2$$

* Mesh current method $\Rightarrow$ This method is based on KVL. In this method instead of assuming branch current mesh current is assumed with do not split but complete their path around the mesh.

1) Step to solve current using mesh analysis $\Rightarrow$ Identify all the mesh and assume mesh current in any direction.

2) Write KVL eqn for each mesh and solve it.
 Note $\rightarrow$ This method is not applicable in a mesh containing current source.



In mesh ABCE

$$-2I_1 - 6I_1 - 6V - 4I_1 + 10 + 6I_2 = 0$$

$$6I_2 - 12I_1 + 4 = 0 \quad \text{--- (1)} \quad 3I_2 - 6I_1 = -2 \quad \text{--- (1) } \times 3$$

$$12I_1 - 6I_2 = 4 \quad \text{--- (2) } \times 3$$

In mesh BCDE

$$-3I_2 - 2 - 5I_2 + 6 - 6I_2 + 6I_1 = 0$$

$$6I_1 - 14I_2 + 4 = 0$$

$$3I_1 - 7I_2 = -2 \quad \text{--- (2) } \times 4$$

$$24I_1 - 42I_2 = -8$$

$$\begin{array}{r} 24I_1 - 42I_2 = -8 \\ 3I_1 - 7I_2 = -2 \\ \hline -42I_2 + 18I_1 = -12 \\ \hline -33I_2 = -18 \\ I_2 = \frac{18}{33} = \frac{6}{11} \end{array}$$

$$12I_1 - 28I_2 = -8$$

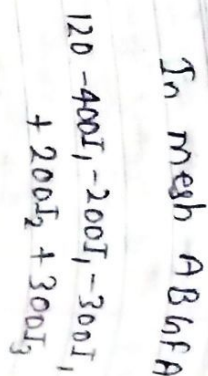
$$12I_1 - 6I_2 = 4$$

$$-24I_2 = -12$$

$$I_1 = 20$$

$$I_1 = \frac{20}{33}$$

$$I_2 = \frac{6}{11}$$



In mesh BCDGB

$$I_1 = \frac{1}{2} I_2 \quad \text{--- (2)}$$

$$-400I_3 - 300I_3 + 300I_4 = 0$$

$$3I_1 = 7I_3$$

$$I_1 = \frac{1}{2} I_2 \quad \text{--- (3)}$$

$$0I_3 = \frac{3I_2}{2} \quad \text{--- (4)}$$

$$120 + 200 I_2 + 300 \times \frac{1}{2} I_2 = 900 \times \frac{1}{2} I_2$$

$$I_2 = \frac{240}{1000} \quad 240 + 400 I_2 + 900 I_2 = 6300 I_2$$

5000 T₂ = 2400

$$I_1 = 7 \times 10^{-04} \text{ g}$$

7x0.024

~~$$I_1 = \frac{7}{2} \times 4.5 = 7 \times 2.25 = 15.75$$~~

II 0.168



$$I_3 = \frac{3}{2} \times 0.048 = 3 \times 0.024 = 6.072 = I_3$$

$$I_3 - I_2 = -0.048 + 0.072 = 0.024$$



$$144 - 3I_1 - 4I_2 - 2I_3 + 4I_4 + 2I_5$$

In mesh CDEF

$$-2I_3 - 8I_3 - 12I_3 + 2I_1 + 8I_2$$

$$8I_2 + 2I_1 = 22I_3 \quad \text{--- (2)}$$

In math $BCDTHG$

$$-4I_1 - 2I_2 + 4I_3 + 2I_4 = 0$$

$$8I_1 + 4I_2 = 8I_2 - (1) \Rightarrow 4I_1 + 8I_2 = 11I_2 \quad (2)$$

in eqn - (1)

$$144 + 4I_3 - I_1 = 9I_1 + 4I_3 \text{ from eqn } - \textcircled{A} + \textcircled{B}$$

$$H_1 = 101$$

I, _____

$$10. \quad 2I_1 - 4I_2 = 22I_3 + 8I_4$$

History

$$H_1 - H_2 = 225.483$$

from eqn ② & ③

$$2I_1 + 4I_3 = 22I_3 - 2I_1$$

$$I_1 = \frac{18}{4} I_3 = \frac{9}{2} I_3 \quad \text{--- (4)}$$

from eqn ① & ③

$$144 - 11I_1 + 11I_3 - I_1 + 2I_3 = 0$$

$$144 + 13I_3 = 10I_1$$

$$144 + 13I_3 = \frac{90}{8} I_3 \Rightarrow 288 = (90 - 26)I_3$$

$$I_3 = 4.5$$

$$I_1 = \frac{9}{2} \times 4.5 = 20.25$$

$$I_1 = 20.25$$

In eqn ③

$$2I_1 + 4I_3 = 8I_2$$

$$2 \times 20.25 + 4 \times 4.5 = 8I_2$$

$$40.5 + 18 = 8I_2$$

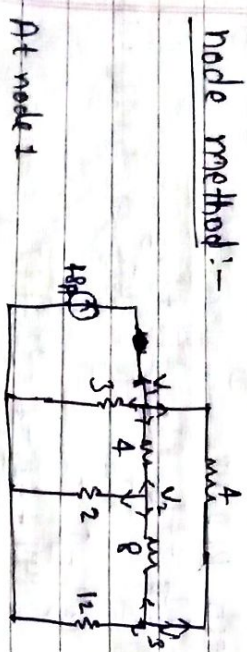
$$\frac{58.5}{8} = I_2$$

$$I_2 = 7.31$$

In 'ED' Branch

$$V = I_3 \times R = 4.5 \times 12 = 54 \text{ Volts}$$

node method:-



At node 1

$$\frac{V_1 - V_2}{4} + \frac{V_1 - V_2}{4} + \frac{V_1}{3} = 48$$

$$3V_1 - 3V_2 + 3V_1 - 3V_2 + 4V_1 = 576$$

$$70V_1 - 3V_2 - 3V_2 = 576 \quad \text{--- (1)}$$

At node 2

$$\frac{V_2 - V_1}{4} + \frac{V_2 - V_3}{8} + \frac{V_2}{2} = 0$$

$$24 - 2V_1 + V_2 - V_3 + 4V_2 = 0 \Rightarrow 7V_2 = 2V_1 + V_3 \quad \text{--- (2)}$$

At node 3

$$\frac{V_3}{12} + \frac{V_3 - V_2}{8} + \frac{V_3 - V_1}{4} = 0$$

$$2V_3 + 3V_3 - 3V_2 + 6V_3 - 6V_1 = 0 \Rightarrow 11V_3 = 3V_2 + 6V_1 \quad \text{--- (3)}$$

In eqn ② & ③

$$21V_2 - 6V_1 = 3V_3$$

$$3V_2 + 6V_1 = 11V_3$$

$$\frac{84V_2}{14} = V_3 \Rightarrow \frac{12}{7} V_2 = V_3$$

$$10V_1 - 3V_2 - 3V_2 = 576 \quad 10V_1 - 6V_2 = 576$$

$$10V_1 - 6 \left(\frac{12}{7} V_2 \right) = 576$$

$$70V_1 - 67V_2 = 4032 \quad 40V_1 - 7V_2 - 12V_3 = 2304$$

$$40V_1 - 19V_3 = 2304 \quad \text{--- (4)}$$

$$eqn ③ \quad \frac{78}{12} V_3 = 24 + V_3$$

$$(49 - 12) V_3 = 24 \times 12$$

$$37V_3 = 24 \times 12$$

$$V_3 = \frac{37}{24} V_3$$

$$\left(\frac{40 \times 37}{24 \times 12} - 19 \right) V_3 = 2304$$

$$(185 - 57) V_3 = 2304 \quad V_3 = \frac{2304}{128}$$

Network Theorem

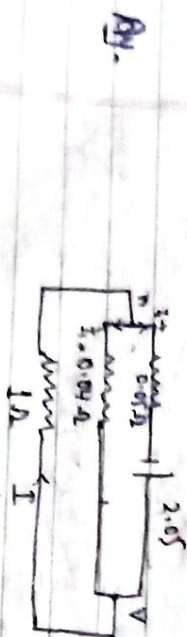
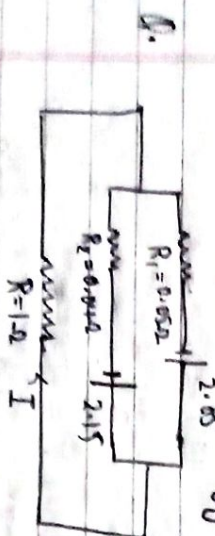
Superposition Theorem $\Rightarrow$

In a linear resistor network containing one or more energy sources, the total current in

any part of a circuit is equal to the algebraic sum of all the currents produced by each source separately

Steps for solving superposition theorem $\Rightarrow$

1. Take any one of the independent voltage or current source. Replace the other energy sources by their internal resistances that is voltage sources become short circuits and current sources become open circuits.
2. Obtain various branch current.
3. Repeat the above two step for each of the independent energy sources.
4. At all the currents obtain in each branch for individual energy source.



$$\frac{dI}{dV} = \frac{1}{R}$$

$$\frac{V - 2.05}{0.05} + \frac{V}{0.04} - \frac{4}{1} = 0$$

$$\frac{100(V - 2.05)}{5} + \frac{V \times 100}{4} - 400 = 0$$

$$20(V - 2.05) + 25V - 4 = 0$$

$$20V - 41 + 25V - 4 = 0$$

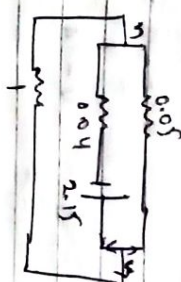
$$45V - 45 = 0$$

$$V = 1V$$

$$920V = 820$$

$$V = \frac{820}{920} = \frac{41}{46} = 0.891$$

$$I = 0.081A$$



$$\frac{V}{0.05} + \frac{V}{1} + \frac{V - 2.15}{0.04} = 0$$

$$\frac{100V}{5} + \frac{V}{1} + \frac{100V - 215}{4} = 0$$

$$20V + 20V + 25V - 53.75 = 0$$

$$45V = 33.75$$

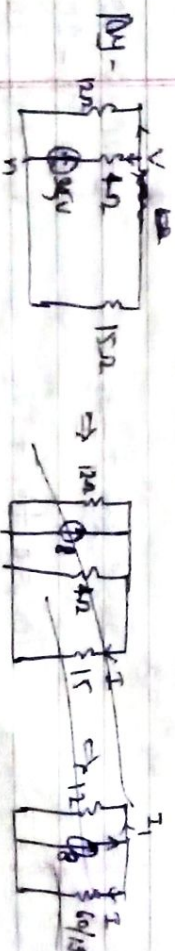
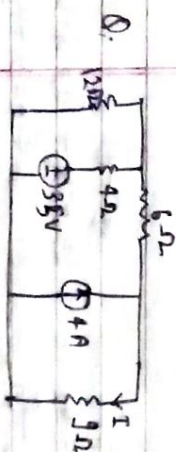
$$V = 0.75V$$

$$920V = 1075$$

$$V = \frac{1075}{920} = 1.16$$

$$V_{total} = 0.891 + 1.16 = 2.051$$

$$I_{total} = 2.051$$

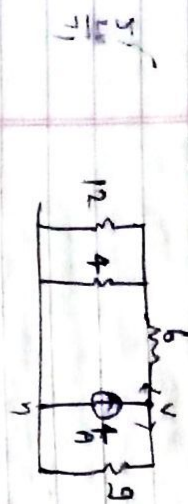


$$\frac{V}{12} + \frac{V - 35}{4} + \frac{V}{15} = 0$$

$$V = 20$$

$$54 \times 12 = 648 + 4V = 19V = 384$$

$$V = I \times R \rightarrow \frac{20}{15} = I \rightarrow I = \frac{4}{3} A$$



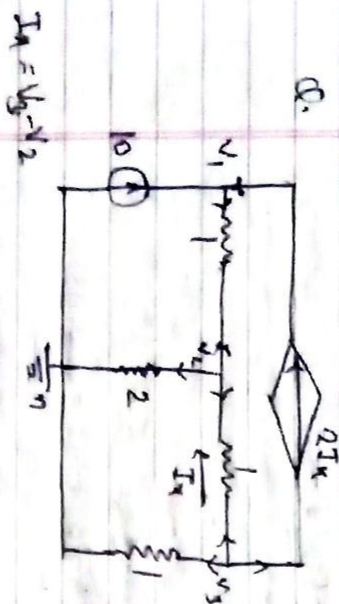
$$V = \frac{9}{2} \times \frac{4}{3} = 6V$$

$$\frac{V}{9} + \frac{V}{9} = 4 \rightarrow V = \frac{9}{2} \times 4 = 18$$

$$I = \frac{V}{R} = \frac{18}{9} = 2$$

$$I_{b1} = \frac{4}{3} + 2 = 1.332 = 3.52$$

dependent



$$2I_x + 10 = V_1 - V_2 \quad \text{--- (1)}$$

$$V_2 - V_1 + \frac{V_2}{2} + V_2 - V_3 = 0$$

$$5V_2 = 2V_1 + 2V_3 \quad \text{--- (2)}$$

$$V_3 - V_2 + 2I_x + V_3$$

$$2V_3 + 2V_3 - 2V_2 = V_2$$

$$4V_3 = 3V_2 \quad \text{--- (3)}$$

eqn - (2)

$$5V_2 = 2V_1 + \frac{3}{2}V_2$$

$$10V_2 = 4V_1 + 3V_2$$

$$7V_2 = 4V_1$$

$$\frac{7 \times 8}{4} = V_1$$

$$V_1 = \frac{56}{4} = 14$$

$$V_2 = \frac{4 \times 14}{7} = 8V$$

$$V_2(1 - \frac{1}{2}) = -20$$

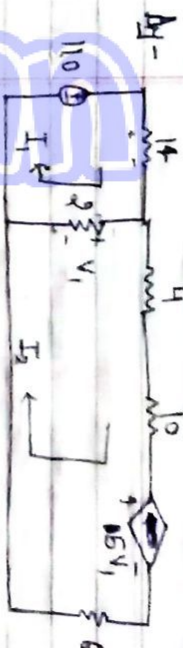
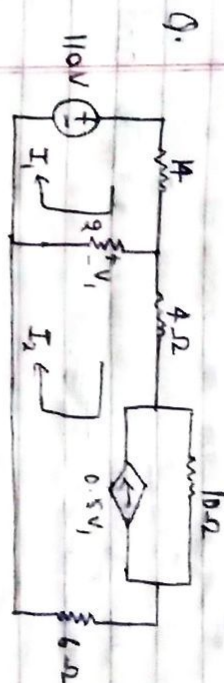
$$V_2(-\frac{1}{2}) = -20$$

$$V_2 = 8V$$

$$4V_3 = 3 \times 8$$

$$V_3 = 6$$

$$I_x = V_3 - V_2 = 6 - 8 = -2A = 2A$$



$$110 - 14I_1 - 2I_1 + 2I_2 = 0$$

$$110 - 16I_1 + 2I_2 = 0 \quad \text{--- (1)}$$

$$-2I_2 - 14I_2 - 6I_2 + 2I_1 - 45V_1 = 0$$

$$5V_1 = 2I_1 - 22I_2 - 5(I_1 - I_2) \times 2$$

$$2I_1 - 22I_2 - 10I_1 + 10I_2$$

$$-8I_1 - 12I_2 = 0$$

$$-12I_2 = 8I_1$$

$$-3I_2 = 2I_1$$

eqn - (1)

$$110 + 24I_1 + 2I_2 = 0$$

$$26I_2 = -110$$

$$I_2 = \frac{-55}{13} = -4.23$$

$$-3(-4.23) = 2I_1$$

$$I_1 = \frac{12.69}{2} = 6.34$$

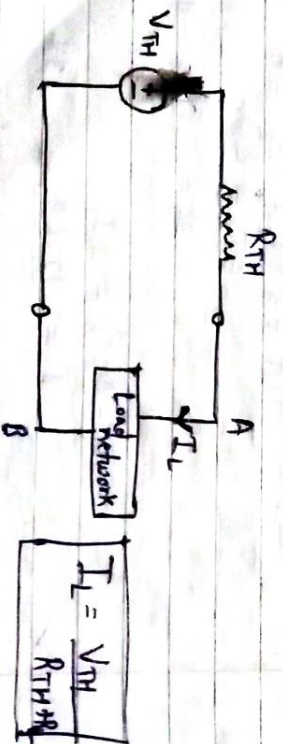
$$V_1 = 6.34 + 4.23$$

$$V_1 = 10.57V$$

Thevenin Theorem:— Any linear resistive network with constant & voltage sources can be replaced by an equivalent circuit containing of a single independent voltage source V_{TH} and series resistance R_{TH} .

Step for solving thevenin theorem $\rightarrow$

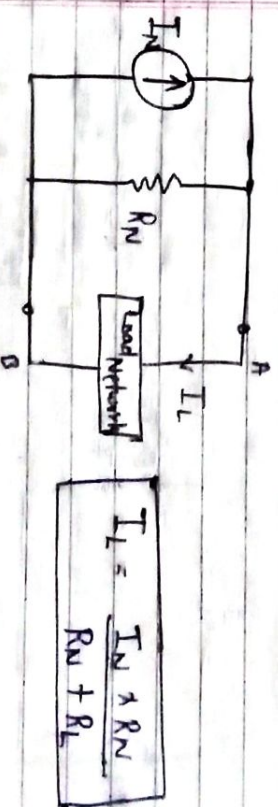
- 1] Open the load resistor
- 2] Calculate the open circuit voltage or thevenin voltage that is V_{TH} .
- 3] Replace the sources by their internal resistances i.e. current source become open circuited and voltage become a short circuited.
- 4] Calculate the thevenin resistance i.e. R_{TH}
- 5] Now re-draw the circuit with measured value of V_{TH} & R_{TH} and connect the load resistor. This is the equivalent thevenin circuit.
- 6] Find the total circuit current I_L flowing through the load resistor.

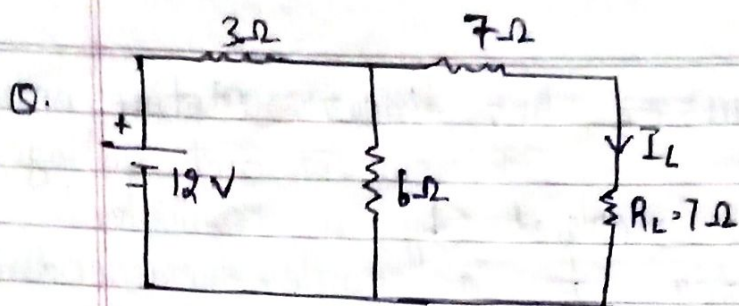


Norton's theorem $\Rightarrow$ Any linear resistive network with constant and voltage sources can be replaced by an equivalent circuit containing of a single independent current source I_N and a parallel resistance R_N .

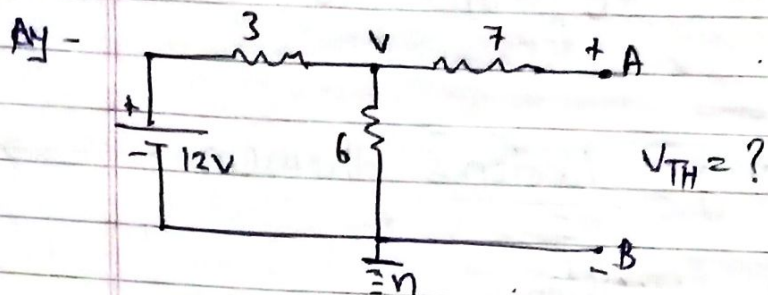
Step for solving Norton's theorem $\rightarrow$

- 1] Short the load resistor.
- 2] Calculate the norton's current i.e. I_N
- 3] Replace the source by their internal resistance i.e. current source becomes short circuited and voltage source become open circuited.
- 4] Calculate the norton's resistance, i.e. R_N
- 5] Now re-draw the circuit with measured value of I_N & R_N and connect the load resistor. This is the equivalent norton's circuit
- 6] Find out the total load current flowing through the load resistor.





Find $I_L = ?$ using
Thevenin theorem



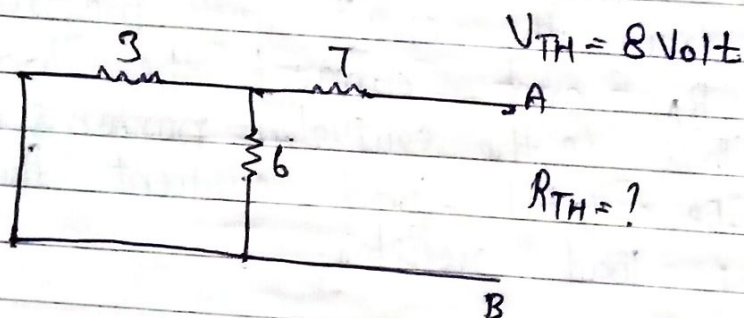
By nodal theorem:-

$$\frac{V-12}{3} + \frac{V}{6} = 0$$

$$\frac{2V-24}{6} = 0$$

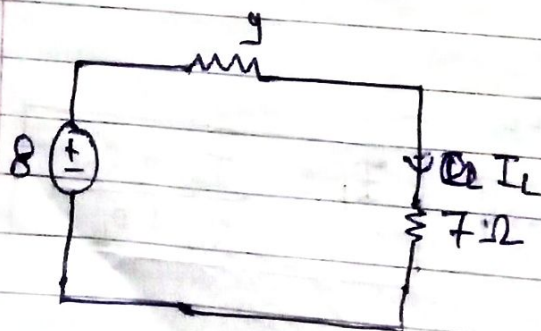
$$3V = 24$$

$$V = 8 \text{ Volt}$$



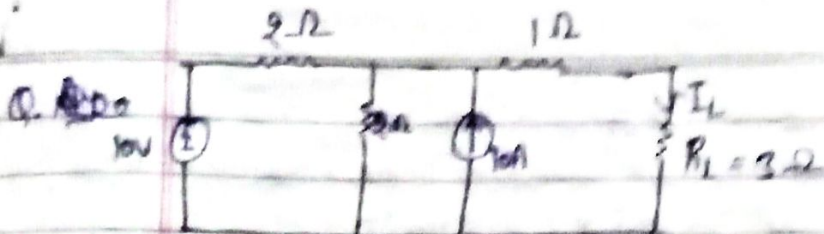
$$R_{TH} = 3 + 7$$

$$R_{TH} = 10 \Omega$$

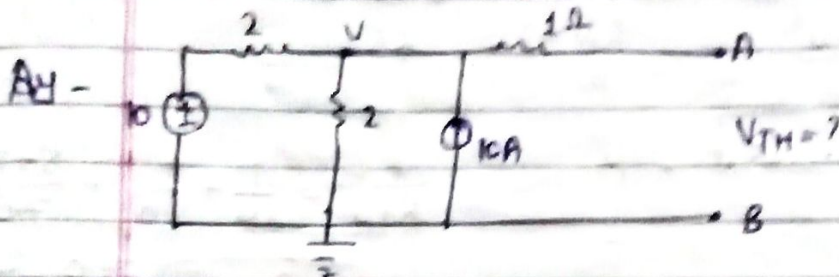


$$I_L = \frac{8}{10+7} = \frac{8}{17} = \frac{1}{2}$$

$$I_L = 0.5 \text{ A}$$

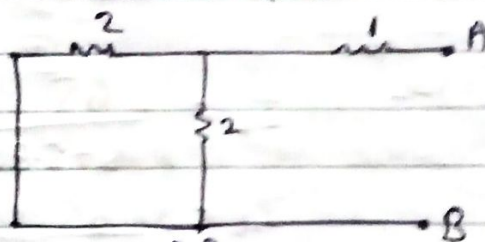


find I_L ?

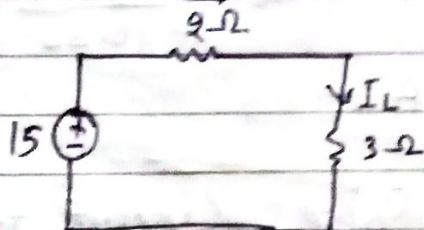


$$\frac{V-10}{2} + \frac{V}{2} = 10$$

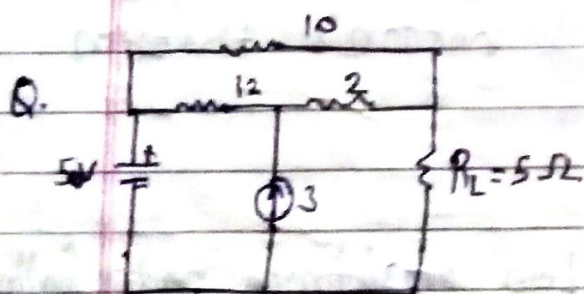
$$2V = 30 \rightarrow V = 15V_{th} \quad V_{TH} = 15V$$



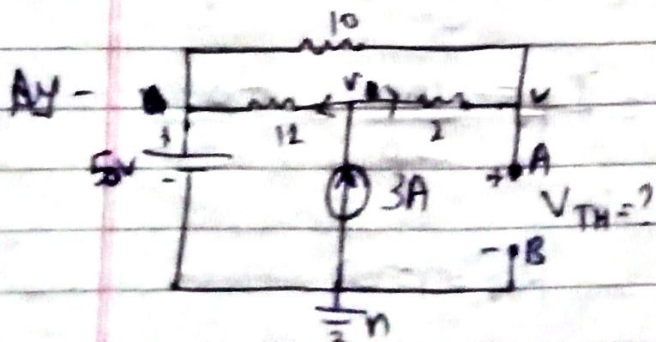
$$R_{TH} = 2\Omega$$



$$I_L = \frac{15}{2+3} = 3A$$



find I_L ?



$$\frac{V-50}{12} + \frac{V-50}{12} = 3$$

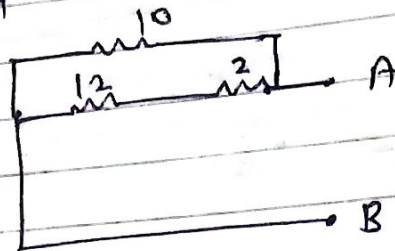
$$2V = 36 + 100$$

$$V = \frac{136}{2} = 68V_{th}$$

$$I = \frac{68 - 80}{12} = \frac{18}{12} = 1.5 \text{ A}$$

$$V' = 1.5 \times 2 = 3 \text{ Volt}$$

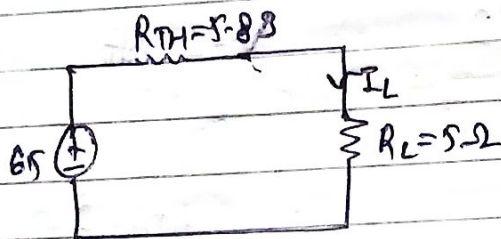
$$V_{TH} = V - V' = 68 - 3 = 65$$



$$\frac{1}{R_{TH}} = \frac{1}{14} + \frac{1}{10} = \frac{5+7}{70}$$

$$R_{TH} = \frac{70}{12} = 5.83$$

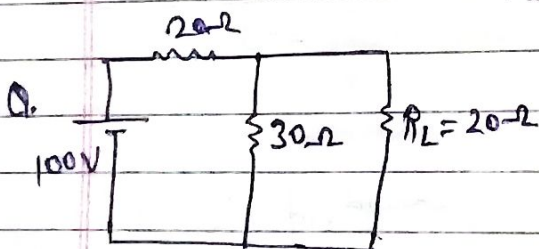
$$R_{TH} = 5.83 \Omega$$



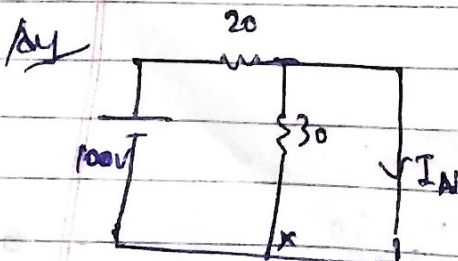
$$I_L = \frac{65}{5 + 5.83}$$

$$I_L = \frac{65}{10.83} = \frac{6500}{1083}$$

$$I_L = 6 \text{ A}$$

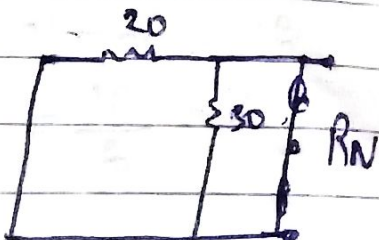


Solve circuit to
norton's theorem



(Low resistance path follow)

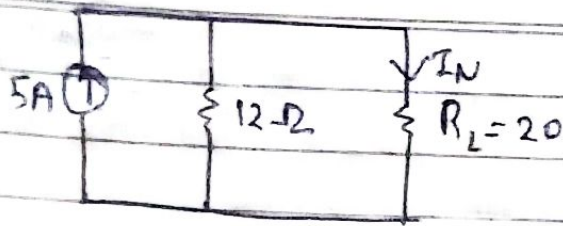
$$I_N = \frac{100}{20} = 5 \text{ A}$$



$$R_N = 12 \Omega$$

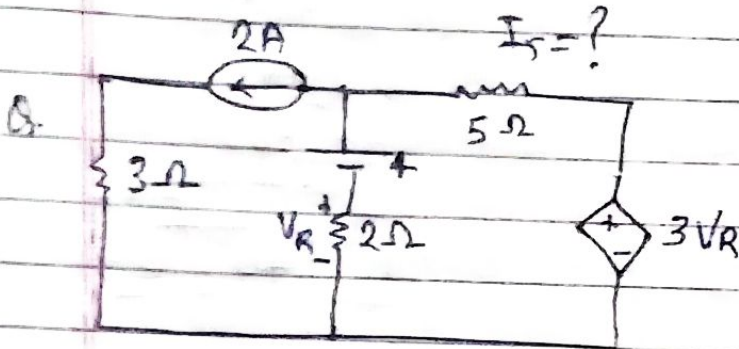
$$\frac{60}{5} = 12$$

$$\frac{1}{20} + \frac{1}{30} = \frac{5}{60}$$

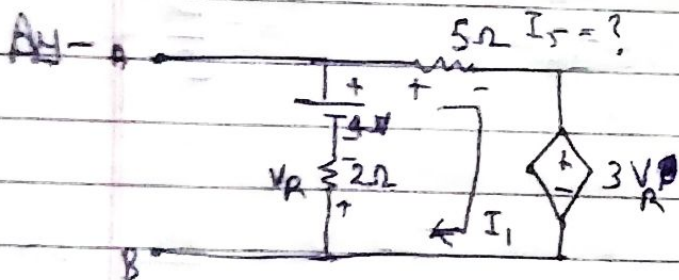


$$I_L = \frac{5 \times 12}{12 + 20} = \frac{60}{32} = \frac{15}{8}$$

$$= \frac{15}{8} = 1.875 \text{ A}$$



find out at
5 ohm Resistance
Using superposition
theorem



By mesh :-

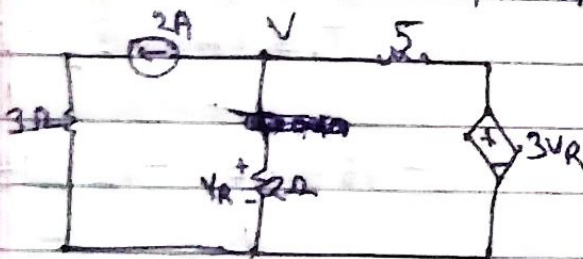
$$-5I_1 - 3V_R - 2I_1 + 4 = 0$$

$$-5I_1 + 6I_1 - 2I_1 + 4 = 0$$

$$I_1 = 4 \text{ A}$$

$$V_R = 2I_1$$

$$V_R = -2I_1$$



$$\therefore V = V_R$$

$$2 + \frac{V}{2} + \frac{V - 3V_R}{5} = 0$$

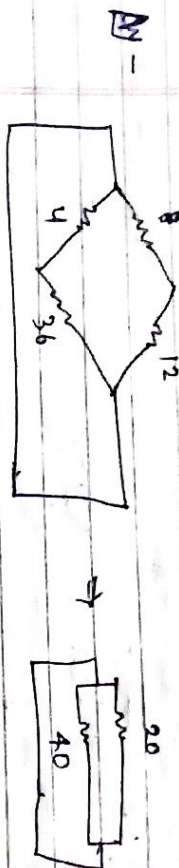
$$2 + \frac{V_R}{2} + \frac{V_R - 3V_R}{5} = 0$$

$$\frac{20 + 5V_R - 4V_R}{10} = 0$$

$$V_R = -20 \text{ Volt}$$

$$I_2 = \frac{V - 3V_R}{5} = \frac{-2V_R}{5} = \frac{-2(-20)}{5} = 8 \text{ A}$$

$$I = 8 + 4 \Rightarrow \boxed{12 \text{ A} = I_{\text{total}}}$$



$$R_{TH} = \frac{1}{\frac{1}{20} + \frac{1}{40}} = \frac{2+1}{3} = R_{TH} = \frac{40}{3} \Omega$$

$$I = \frac{V}{R} = \frac{40}{40/3} = 3A$$

$$I_1 = 3 \times \frac{20}{60} = 1A, I_2 = 2A$$

$$I_2 = 3 \times \frac{40}{60} = 2A$$

$$V_{TH} = V_{AC} + V_{BC} = -16 + 4 = -12V$$

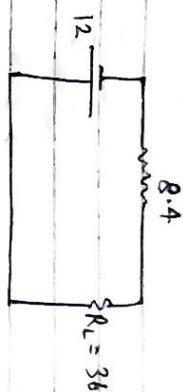
$$I_L = \frac{V_{TH}}{R_{TH} + R_L} = \frac{-12}{\frac{40}{3} + 36} = \frac{-12}{148} = -\frac{3}{37}A$$

$$I_L = \frac{-36}{148} = -\frac{9}{37}A$$

$$\frac{1}{8} + \frac{1}{12} = \frac{1}{4} \left(\frac{1}{2} + \frac{1}{3} \right) = \frac{1}{4} \times \frac{5}{6} = \frac{5}{24}$$

$$\frac{1}{4} \left(\frac{1}{2} + \frac{1}{3} \right) = \frac{1}{4} \times \frac{5}{6} = \frac{5}{24}$$

$$R_{TH} = \frac{18+24}{5} = \frac{42}{5} = 8.4 \Omega$$

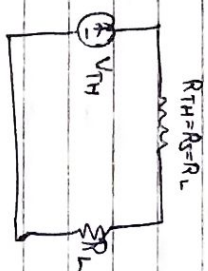


$$I_L = \frac{-12}{8.4 + 36} = \frac{-12}{44.4} = -\frac{1}{3.7}A$$

$$I_L = \frac{+120}{444} = \frac{+60}{222} = \frac{+30}{111}$$

Maximum Power Transfer Theorem:-

$$R_S = R_L$$



$$I_L = \frac{V}{R_S + R_L}$$

Power transfer.

$$P_L = I_L^2 R_L$$

$$P_L = \frac{V^2}{(R_S + R_L)^2} \times R_L$$

$$P_L = \frac{V^2}{R_S^2 + R_L^2 + 2R_S R_L} \times R_L$$

$$P_L = \frac{V^2}{R_S^2 + R_L^2 + 2R_S R_L}$$

$$\left(\frac{R_S^2}{R_L} + R_L + 2R_S \right)$$

for max. power transfer $P_{max} = \frac{dP_L}{dR_L} = 0$

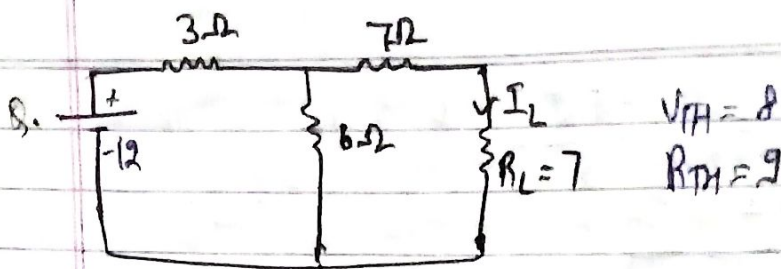
$$-\frac{R_S^2}{R_L^2} + 1 = 0$$

$$R_S = R_L$$

The max. power transfer theorem is used to determine the condition under which max. power can be transferred from a source network to load. According to this theorem max. power will be transferred from a source network to a load. If the

load resistance R_L equal to the equivalent resistance R_S of the source network

i.e. $R_S = R_L$



$$I_L = \frac{V_{TH}}{(2R_S)}$$

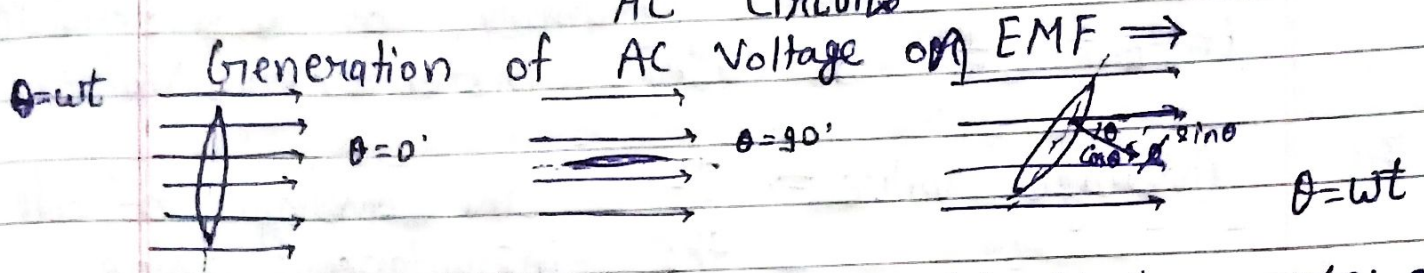
$$P = I_L^2 R$$

$$P = \frac{V^2}{4R^2} \times R = \frac{8 \times 8}{4 \times 9} = \frac{64}{36}$$

$$P = \frac{16}{9} = 1.77$$

Unit - 2

AC Circuits



$$e = -\frac{d(N\phi)}{dt} = N \frac{d(-\phi \cos \theta)}{dt} = \omega N \phi \sin \omega t$$

$$e = N \phi \omega \sin \omega t$$

$$e = E_{\max} \sin \omega t$$

According to the Faraday's Law of electro-magnetic induction when a moving conductor is placed in a magnetic field emf is generated. It is important to note that max. of flux is linked with the conductor when the plane of the conductor is $\perp$ to the flux lines. And it is zero when the plane of the coil is in parallel. At any other angle θ the flux linked with the conductor is equal to the component of ϕ which is $\perp$ to the coil.

- | | | |
|--------------------------|-------------------|----------------------|
| 1.] Alternating Quantity | 3.] Cycle | 4.] Time period |
| 2.] Waveform | 6.] Phase angle | 7.] Phase difference |
| 5.] Frequency | 9.] Average Value | 10.] peak value |
| 8.] Instantaneous Value | | |

1.] Alternating Quantity $\Rightarrow$ It is the Quantity which changes periodically in magnitude and direction w.r.t. time.

2.1 Instantaneous Value $\Rightarrow$ The value of alternating quantity at any particular instant t is called instantaneous value.

2.2 Average Value $\Rightarrow$ It is the average of all the instantaneous values over one complete cycle.

2.3 Time period $\Rightarrow$ Time taken by an alternating cycle is called time period and it is denoted by 'T'.

2.4 Frequency $\Rightarrow$ The number of cycles completed first, second by an alternating quantity is called 'frequency' and denoted by 'f'.

2.5 Peak value $\Rightarrow$ It is the maximum magnitude of an alternating quantity.

2.6 Wave form $\Rightarrow$ The curve obtained by the plotting instantaneous values over time is called 'wave form'.

2.7 Cycle $\Rightarrow$ One complete set of positive and negative values of an alternating wave is called cycle.

3.1 Phase angle $\Rightarrow$ It is the angle of an AC wave at $t = 0$.

3.2 Phase difference $\Rightarrow$ It is the angular difference b/w the phase angle of the two waves.

Average Value and RMS Value $\Rightarrow$

Average value:- It is the average of all the instantaneous values of an alternating quantity over one complete cycle.

$I_{av} = \frac{\text{Area under full cycle}}{\text{length of one cycle}}$

$$I_{av} = \frac{1}{2\pi} \int_0^{2\pi} i \, d\theta$$



Note:- In case of periodic alternative quantity that is one where two half cycles are same whether it is sinusoidal or non-sinusoidal. The average value over one complete cycle is zero.

RMS Value:- It is the square root of mean of square instantaneous value.

The RMS value of an alternating quantity is given by that steady voltage or current which when flow through a resistor of known resistance for a given periodic time

produces the same amount of heat when the alternating current or voltage is applied to the same resistor for the same period of time.

$$I_{rms} = \sqrt{\frac{1}{2\pi} \int_0^{2\pi} i^2(t) \cdot dt}$$

Form vector (k_f) $\Rightarrow$ It is the ratio of effective value or rms value to the average value.

$$k_f = \frac{I_{rms}}{I_{av}}$$

Note:- For sinusoidal wave $k_f = 1.11$

For square wave $k_f = 1$

For triangular wave $k_f = 1.547$

Peak vector (k_p) $\Rightarrow$ It is the ratio of max value to the rms value.

$$k_p = \frac{I_{max}}{I_{rms}}$$

Note:- For sin wave $k_p = 1.414$

For square wave $k_p = 1$

For triangular wave $k_p = 1.732$

Q. Find out average value, rms value, form value and peak value for sin wave.

Ans-



$$I_{av} = \frac{1}{\pi} \int_0^{\pi} i \, dt = \frac{1}{\pi} \int_0^{\pi} I_{max} \sin \omega t \, dt$$

$$= \frac{I_{max}}{\pi} [-\cos t]_0^{\pi} = \frac{I_{max}}{\pi} [-\cos \pi + \cos 0]$$

$$I_{qv} = \frac{2 I_{max}}{\pi}$$

$$I_{rms}^2 = \frac{1}{2\pi} \int_0^{2\pi} I_{max}^2 \sin^2 t \, dt$$

$$I_{rms}^2 = \frac{I_{max}^2}{2\pi} \int_0^{\pi} (1 - \cos 2t) \cdot dt$$

$$I_{rms}^2 = \frac{I_{max}^2}{2\pi} \left[t - \frac{\sin 2t}{2} \right]_0^{\pi} = \frac{I_{max}^2}{2\pi} (\pi)$$

$$I_{rms} = \frac{I_{max}}{\sqrt{2}}$$

$$k_f = \frac{I_{rms}}{I_{av}} = \frac{\frac{I_{max}}{\sqrt{2}} \times \frac{\pi}{2 I_{max}}}{\frac{2 I_{max}}{\pi}} = \frac{\pi}{2\sqrt{2}} = \frac{3.14}{2.82} = 1.11$$

$$k_p = \frac{I_{max}}{I_{rms}} = \frac{I_{max}}{\frac{I_{max}}{\sqrt{2}}} \times \sqrt{2} = \sqrt{2} = 1.414$$



$$V_{av} = \frac{1}{2\pi} \int_0^{2\pi} v(t) \cdot dt = \frac{1}{2\pi} \left[\int_0^{\pi} V_m \sin t \cdot dt + \int_{\pi}^{2\pi} 0 \cdot dt \right]$$

$$V_{av} = \frac{V_m}{2\pi} [-\cos t]_0^{\pi} = \frac{V_m}{\pi}$$

$$V_{rms} = \sqrt{\frac{1}{8\pi} \int_0^{2\pi} V^2 dt}$$

$$V_{rms} = \sqrt{\frac{1}{2\pi} \left[\int_0^{\pi} V_m^2 \sin^2 t dt + 0 \right]}$$

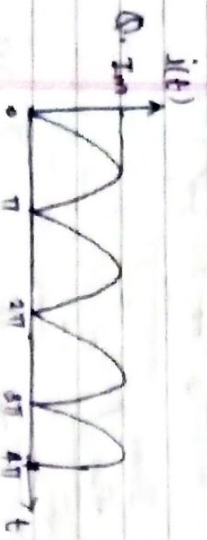
$$V_{rms}^2 = \frac{1}{8\pi} V_m^2 \int_0^{\pi} (1 - \cos 2t) dt$$

$$V_{rms} = \frac{V_m}{\sqrt{2}} \quad (\pi - 0) = \frac{V_m}{\sqrt{2}}$$

$$V_{rms} = \frac{V_m}{\sqrt{2}}$$

$$k_p = \frac{V_{rms}}{V_{av}} = \frac{V_m}{2} \times \frac{\pi}{V_m} = \frac{3.14}{2} = 1.57$$

$$k_p = \frac{V_{rms}}{V_{avg}} = \frac{V_m}{2} \times 2 = 2$$



$$I_{av} = \frac{1}{\pi} \int_0^{\pi} I_m \sin t dt$$

$$I_{av} = \frac{2I_m}{\pi} \quad \text{Avg}$$

$$I_{rms}^2 = \frac{1}{\pi} \int_0^{\pi} I_m^2 \sin^2 t dt$$

$$I_{rms} = \frac{I_0}{\sqrt{2}}$$

$$k_f = \frac{I_{rms}}{I_{av}} = \frac{I_0}{2I_m} = \frac{1}{\sqrt{2}}$$

$$k_p = \frac{I_{rms}}{I_{avg}} = \frac{I_m}{2I_m} = \frac{1}{2}$$

Q.



$$i(t) = \begin{cases} 0 & 0 \text{ to } \pi/4 \\ 100 \sin t & \pi/4 \text{ to } \pi \\ 0 & \pi \text{ to } 2\pi \end{cases}$$

$$I_{av} = \frac{1}{2\pi} \left[\int_0^{\pi/4} 0 + \int_{\pi/4}^{\pi} 100 \sin t dt + \int_{\pi}^{2\pi} 0 \right]$$

$$I_{av} = \frac{1}{2\pi} \left[-100 [\cos t]_{\pi/4}^{\pi} \right]$$

$$I_{av} = \frac{-100}{2\pi} \left(-1 - \frac{1}{\sqrt{2}} \right) = \frac{50}{\pi} \times \frac{1.71}{\sqrt{2}}$$

$$I_{av} = \frac{50}{3.14} \times \frac{2.41}{1.41} = \frac{12.05}{4.214} = 2.86$$

$$I_{rms}^2 = \frac{1}{2\pi} \left[\int_0^{\pi/4} 0 + \int_{\pi/4}^{\pi} 100^2 \sin^2 t dt + 0 \right]$$

$$I_{rms}^2 = \frac{5000}{\pi} \left[\int_{\pi/4}^{\pi} \frac{1 - \cos 2t}{2} dt \right]$$

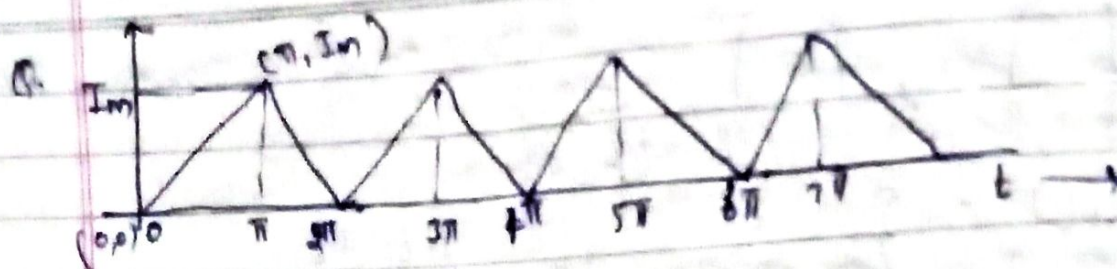
$$= \frac{5000}{\pi \times 2} \left[t - \frac{\sin 2t}{2} \right]_{\pi/4}^{\pi}$$

$$= \frac{5000}{2\pi} \left(\frac{2\pi}{4} - \left(0 - \frac{1}{2} \right) \right)$$

$$I_{rms} = \frac{5000 \times 1}{8} + \frac{5000}{2\pi} = 5000 \left(\frac{3}{8} + \frac{100}{1256} \right)$$

$$5000 \times 0.3 = 1500$$

$$I_{rms} = \sqrt{1500^2 + 1256^2} = 1976$$



$$I_{av} = \frac{\frac{1}{2} \times \pi \times I_m}{\pi} = \frac{I_m}{2}$$

$$y = mx + c$$

$$i(t) = \frac{y_2 - y_1}{x_2 - x_1} \times t + c$$

$$i(t) = \frac{I_m}{\pi} \times t$$

$$I_{av} = \frac{1}{\pi} \int_0^{\pi} \frac{I_m}{\pi} \cdot t \cdot dt$$

$$I_{av} = \frac{1}{\pi} \times \frac{I_m}{\pi} \times \frac{\pi^2}{2} = \frac{I_m}{2}$$

$$I_{av} = \frac{I_m}{2}$$

$$I_{rms}^2 = \frac{1}{\pi} \int_0^{\pi} \frac{I_m^2}{\pi^2} \times t^2 \cdot dt$$

$$I_{rms}^2 = \frac{1}{\pi} \times \frac{I_m^2}{\pi^2} \times \frac{\pi^3}{3} = \frac{I_m^2}{3}$$

$$I_{rms} = \frac{I_m}{\sqrt{3}}$$

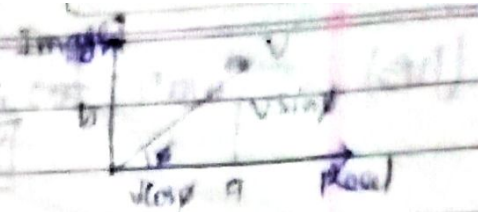
$$I_{rms} = \frac{0.577 I_m}{\sqrt{3}} = \frac{I_m}{\sqrt{3}}$$

$$I_{rms}^2 = \frac{1}{\pi} \int_0^{\pi} \frac{I_m^2}{\pi^2} t^2 \cdot dt$$

$$I_{rms}^2 = \frac{1}{\pi} \times \frac{I_m^2}{\pi^2} \left(\frac{\pi^3}{3} \right) = \frac{I_m^2}{3}$$

$$I_{rms} = \frac{I_m}{\sqrt{3}}$$

Representation of Phasor:—



(i) Rectangular form:—

$$V = a + jb$$

Magnitude, $V = \sqrt{a^2 + b^2}$

phase angle $\phi = \tan^{-1}\left(\frac{b}{a}\right)$

(ii) Triangular form:—

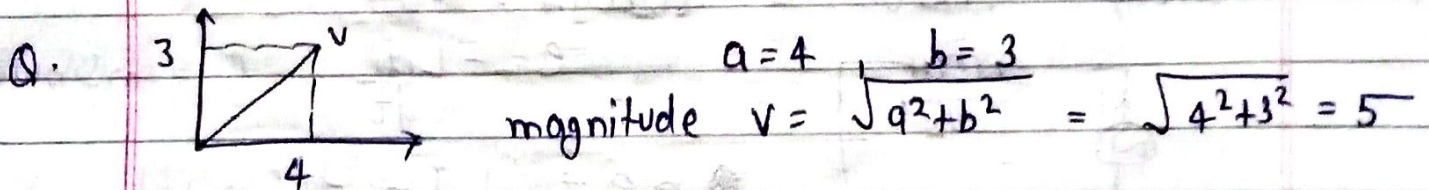
$$a = V \cos \phi$$

$$b = V \sin \phi$$

$$V = \underset{\text{magnitude}}{V} [\cos \phi + j \sin \phi]$$

(iii) Polar form:—

$$V = \underset{\text{magnitude}}{V} \angle \phi$$



Rectangular form

$$V = 4 + j3$$

$$\phi = \tan^{-1}\left(\frac{3}{4}\right) = 37^\circ$$

$$\cos \phi = \frac{4}{5}, \sin \phi = \frac{3}{5}$$

(ii) Triangular form

$$V = 5 \left[\frac{4}{5} + j \frac{3}{5} \right] = 5 [\cos \phi + j \sin \phi]$$

$$V = 5 [\cos 37^\circ + j \sin 37^\circ]$$

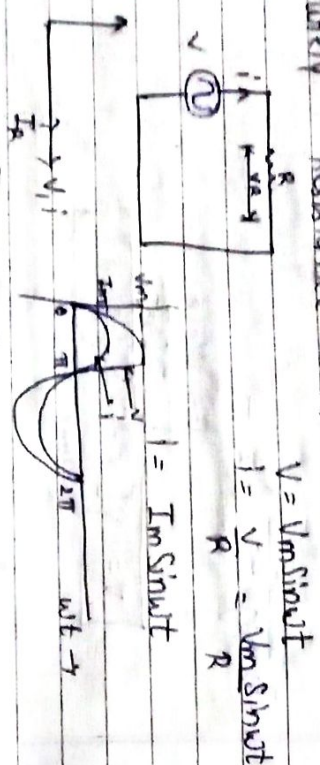
(i)

$V = 5 \angle 31^\circ$
 $V = 5 \tan^{-1}(\frac{3}{4})$

$V = 5 \angle 31^\circ$

Analysis of AC circuit $\Rightarrow$

1) Purely Resistance circuit:-



Active power $P = VI \cos \phi$

$P = VI$

(ii) Purely inductive circuit:-



$e = -L \frac{di}{dt}$ ($i = I_m \sin \omega t$)

$e = -I_m \frac{d(\sin \omega t)}{dt}$

$V = \omega L I_m \sin \omega t = V_m \sin(\omega t + \pi/2)$



$V_m = L I_m \omega = \omega L I_m$
 $V_m = X_L I_m$

$X_L = \omega L = 2\pi f L$

$X_L = \frac{V_m}{I_m}$

$P = VI \cos \phi = VI \cos 90^\circ$

$P = 0$

(iii) Purely capacitor circuit:-



$i = C \frac{dv}{dt}$

where $V = V_m \sin \omega t$

$i = C \frac{d(V_m \sin \omega t)}{dt}$

$i = I_m \cos \omega t = I_m \sin(\omega t + \pi/2)$



$I_m = \omega C V_m$
 $\frac{1}{\omega C} = \frac{V_m}{I_m} = X_C$

$X_C = \frac{V_m}{I_m}$

$X_C = \frac{1}{\omega C} = \frac{1}{2\pi f C}$

$P = 0$

Q. $V = 75 \sin(200\pi t - 0.25)$ find (i) amplitude and

rms value (ii) Time & frequency (iii) Phase angle ϕ

(i) 75, $V_{rms} = \frac{75}{\sqrt{2}}$

(ii) $2\pi f = 200\pi$

$f = 100 \text{ Hz}$

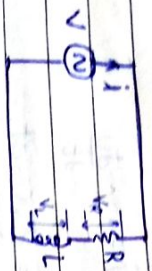
$T = \frac{1}{100} \Rightarrow 0.01 \text{ sec} = T$

(iii) $0.25 \times \frac{180}{\pi} = \frac{1}{4} \times \frac{180}{\pi} = \frac{45}{\pi} = \frac{4500}{314}$

$= 14.33^\circ$

R-L series circuit:-

Apply KVL in the loop.

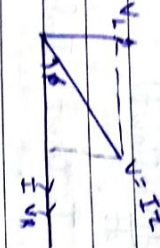


$$V = V_R + V_L$$

$$V = IR + jX_L I$$

$$V = I(R + jX_L)$$

$$V = IZ$$

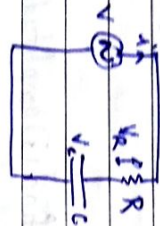


$$Z = R + jX_L$$

$$Z = \sqrt{R^2 + X_L^2}$$

$$\phi = \tan^{-1}\left(\frac{X_L}{R}\right)$$

R-C series circuit:-



$$V = V_R + V_C$$

$$V = IR + jIX_C$$

$$V = I(R + jX_C)$$

$$V = IZ$$

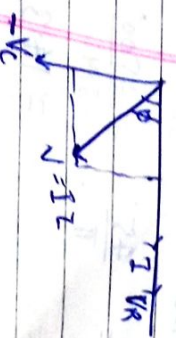
$$V = I\left(R + j\frac{1}{\omega C}\right)$$

$$V = IZ$$

$$Z = R + j\frac{1}{\omega C}$$

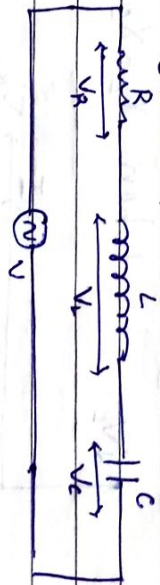
$$Z = \sqrt{R^2 + \left(\frac{1}{\omega C}\right)^2}$$

$$\phi = \tan^{-1}\left(\frac{1}{\omega CR}\right)$$



$$\cos \phi = \frac{V_R}{V}$$

Series RLC circuit



Applying KVL

$$V = V_R + V_L + V_C$$

$$V = IR + jIX_L - jIX_C$$

$$V = I(R + jX_L - jX_C)$$

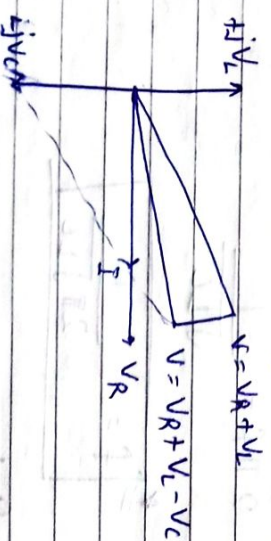
$$V = IZ$$

$$Z = R + jX_L - jX_C$$

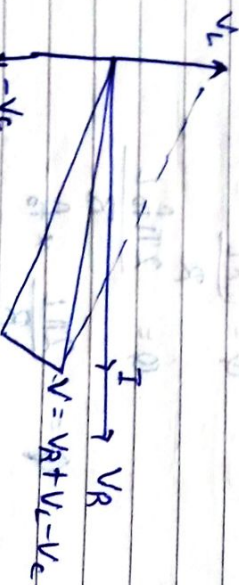
$$Z = \sqrt{R^2 + (X_L - X_C)^2}$$

$$\phi = \tan^{-1}\left(\frac{X_L - X_C}{R}\right)$$

Case-I when $X_L > X_C$



Case-II when $X_C > X_L$



Case - III

When $X_L = X_C$



At resonance

$$X_L = X_C$$

$$\omega L = \frac{1}{\omega C}$$

$$\omega^2 = \frac{1}{LC}$$

$$\omega = \frac{1}{\sqrt{LC}}$$

$$2\pi f = \frac{1}{\sqrt{LC}}$$

$$f = \frac{1}{2\pi\sqrt{LC}}$$

Q factor

$$Q = \frac{X_L}{R}$$

$$Q = \frac{2\pi f L}{R}$$

$$Q = \frac{2\pi L}{R} \times f$$

$$Q = \frac{2\pi L}{R} \times \frac{1}{2\pi\sqrt{LC}}$$

$$Q = \frac{1}{R} \sqrt{\frac{L}{C}}$$

$$Q = \frac{1}{R} \sqrt{\frac{L}{C}}$$

FORMULA

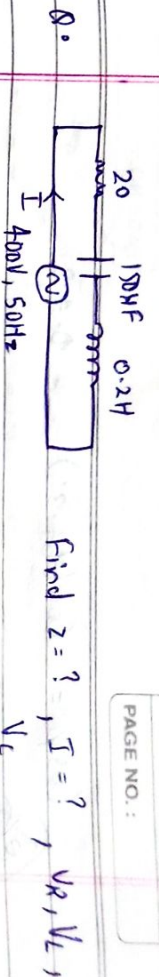
$$V_R = IR, \quad V_L = jIX_L, \quad V_C = jIX_C, \quad I = \frac{V_L}{Z_L}, \quad V = \sqrt{V_R^2 + V_C^2 + V_L^2}$$

$$X_L = 2\pi fL, \quad X_C = \frac{1}{2\pi fC}, \quad V = \sqrt{V_R^2 + V_C^2 + V_L^2}$$

$$V_R = \sqrt{V_R^2 + V_C^2}, \quad Z = R + jX_L, \quad Z = R - jX_C, \quad Z = R + j(X_L - X_C)$$

$$\phi = \tan^{-1} \left(\frac{X_L - X_C}{R} \right), \quad \phi = \tan^{-1} \frac{X_L}{R}$$

$$\phi = \tan^{-1} \frac{X_C}{R}, \quad P = VI \cos \phi, \quad Q = VI \sin \phi$$



$$Z = \sqrt{(X_L - X_C)^2 + R^2}$$

$$Z = \sqrt{(0.2 - 0.00015)^2}$$

$$X_L = 2\pi fL = 2 \times \pi \times 50 \times 0.2 = 20\pi = 62.8$$

$$X_C = \frac{1}{2\pi fC} = \frac{10^6 \times 150}{2 \times 314 \times 50 \times 150} = \frac{100000}{471} = 21.23$$

$$Z = R + j(X_L - X_C)$$

$$Z = 20 + j(62.8 - 21.2) = 41.6$$

$$Z = \sqrt{400^2 + 1730.56} = 4130.56$$

$$\phi = \tan^{-1} \left(\frac{62.8 - 21.2}{20} \right) = \frac{61.6}{20} = 71.308$$

$$\phi = 64.32^\circ = 64.3^\circ$$

$$I = \frac{V_L}{Z_L} = \frac{400}{41.6} = 9.61$$

$$I = \frac{400 \angle 0^\circ}{46.15 \angle 64.3^\circ} = 8.66 \angle -64.3^\circ$$

$$V_R = IX_R = 8.66 \angle -64.3^\circ \times 20 \angle 0^\circ = 173.2 \angle -64.3^\circ$$

$$V_C = jIX_C = (8.66 \angle -64.3^\circ) (21.23 \angle 90^\circ) = 183.8 \angle 25.7^\circ$$

$$V_C = IX_C = j(8.66 \angle -64.3^\circ) (21.23) = 183.8 \angle 25.7^\circ$$

$$\cos \phi = \cos(64.3^\circ) = 0.43$$

$$P = VI \cos \phi$$

$$P = 400 \times 8.66 \times 0.43$$

$$P = 1489.5 \text{ W}$$

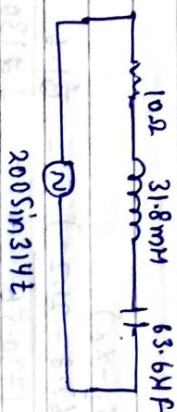
Real Power

$$Q = VI \sin \phi$$

$$Q = 400 \times 8.66 \times (-0.9)$$

$$Q = -3117.6 \text{ VAR}$$

Total Power Unit (VA)



$$X_L = 10 = 31.8 \times 10^{-3} \times 314 = 9.98 \approx 10 \Omega$$

$$X_C = \frac{1}{\omega C} = \frac{1}{63.6 \times 314} = 50 \Omega$$

$$Z = \sqrt{(10)^2 + (10 - 50)^2} = 10 \sqrt{17} = 41.23 \Omega$$

$$\phi = \tan^{-1} \left(\frac{-40}{10} \right) = -75.96^\circ = -76^\circ$$

$$I = \frac{V}{Z} = \frac{200}{41.23} = 4.85 \angle -76^\circ \text{ Amp}$$

$$\cos \phi = \cos(76^\circ) = 0.24$$

$$P = VI \cos \phi = 200 \times 4.85 \times 0.24$$

$$P = 234.66 \text{ watt}$$

$$Q = VI \sin \phi = 200 \times 4.85 \times 0.97$$

$$= 941.19 \text{ VAR}$$



$$V_{avg} = \frac{1}{2} \left[\int_0^1 4 \cdot dt + \int_1^2 (-4t + 4) \cdot dt \right] = \frac{1}{2} \left[(4t) \Big|_0^1 + (-2t^2 + 4t) \Big|_1^2 \right]$$

$$V_{avg} = \frac{1}{2} [(4 - 0) + (-8 + 8) + (-4 + 2)] = 1 \text{ volt}$$

$$V_{rms}^2 = \frac{1}{2} \left[16 \int_0^1 dt + 16 \int_1^2 (t^2 + 1 - 2t) \cdot dt \right]$$

$$V_{rms}^2 = 8 \left[[t]_0^1 + \left[\frac{t^3}{3} + t - t^2 \right]_1^2 \right]$$

$$V_{rms}^2 = 8 \left[(1 - 0) + \left(\frac{8}{3} + 2 - 4 \right) - \left(\frac{1}{3} + 1 - 1 \right) \right]$$

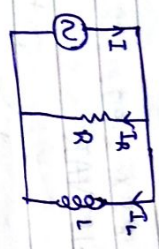
$$V_{rms}^2 = 8 \left[1 + \frac{8}{3} - 2 - \frac{1}{3} \right]$$

$$V_{rms}^2 = \frac{32}{3} \text{ volt}$$

$$V_{rms} = \sqrt{10.67}$$

$$V_{rms} = 3.27 \text{ volt}$$

Parallel RL circuit: $\rightarrow$



Apply KCL

$$I = I_R + I_L$$

$$I = \frac{V}{R} + \frac{V}{jX_L}$$

$$I = V \left(\frac{1}{R} - \frac{j}{X_L} \right)$$

$$I = VY$$

where Y is admittance

$$Z = \frac{1}{Y} \text{ impedance}$$

$$Y = \frac{1}{R} - \frac{j}{X_L}$$

$$Y = G + jB$$

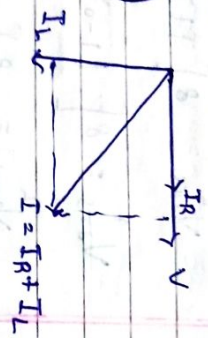
Conductance

Susceptance

$$G = \frac{1}{R}$$

$$B = \frac{-1}{\omega L}$$

$$\phi = \tan^{-1} \left(\frac{B}{G} \right)$$



Parallel RC circuit: $\rightarrow$



$$I = I_R + I_C$$

$$I = \frac{V}{R} + \frac{V}{-jX_C}$$

$$I = \frac{V}{R} + j \frac{V}{X_C}$$

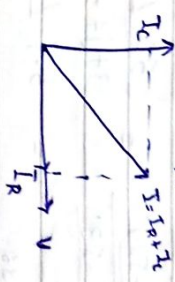
$$I = V \left(\frac{1}{R} + \frac{j}{X_C} \right)$$

$$I = VY$$

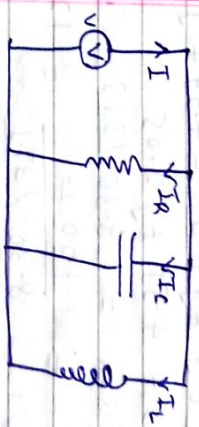
$$Y = \frac{1}{R} + \frac{j}{X_C}$$

$$Y = \frac{1}{R} + j\omega C = G + jB$$

$$\phi = \tan^{-1} \left(\frac{B}{G} \right)$$



Parallel RLC circuit: $\rightarrow$



Apply KCL

$$I = I_R + I_C + I_L$$

$$I = \frac{V}{R} - \frac{V}{jX_C} + \frac{V}{jX_L}$$

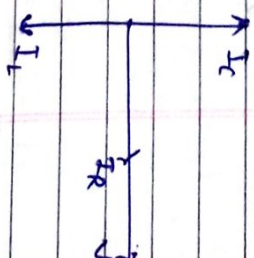
$$I = V \left(\frac{1}{R} + \frac{j}{X_C} - \frac{j}{X_L} \right)$$

$$I = VY$$

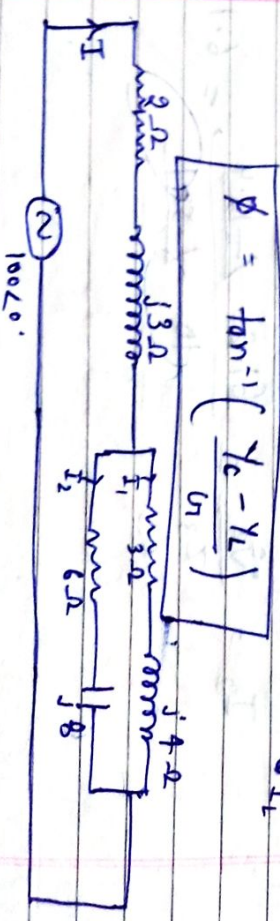
$$Y = \frac{1}{R} + j\omega C - \frac{j}{\omega L}$$

$$Y = G + j(Y_C - Y_L)$$

$$\phi = \tan^{-1} \left(\frac{Y_C - Y_L}{G} \right)$$



Q.



Calculate (i) V_1, P, V_2 (ii) I, P, I_2

AN - $V_1 = I Z_1$, $Z_{23} = \frac{Z_2 \times Z_3}{Z_2 + Z_3}$

$Z_1 = 2 + j3 = 3.6 \angle 56.3^\circ$

$Z_2 = 3 + j4 = 5 \angle 53.1^\circ$

$Z_3 = 6 - j8 = 10 \angle -53.1^\circ$

$Z_{23} = \frac{5 \angle 53.1^\circ \times 10 \angle -53.1^\circ}{5 \angle 53.1^\circ + 10 \angle -53.1^\circ} = \frac{50 \angle 0^\circ}{3 + j4 + 6 - j8} = \frac{50 \angle 0^\circ}{9 - j4}$

$Z_{23} = \frac{50 \angle 0^\circ}{9 - j4} = \frac{50 \angle 0^\circ}{9.84 \angle -23.9^\circ} = 5.1 \angle 23.9^\circ$

$Z_{23} = 5.1 \angle 23.9^\circ = 5.1 [0.9 + j0.4] = 4.59 + j2.05$

$Z = Z_1 + Z_{23} = 2 + j3 + 4.59 + j2.05 = 6.59 + j5.05 = 8.30 \angle 37.3^\circ$

$I = \frac{V \angle 0^\circ}{Z \angle 37.3^\circ} = \frac{100 \angle 0^\circ}{8.30 \angle 37.3^\circ} = 12 \angle -37.3^\circ$

$V_1 = I Z_1 = 12 \angle -37.3^\circ \times 3.6 \angle 56.3^\circ$

$V_1 = 43.2 \angle 19^\circ$

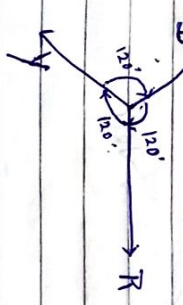
$V_2 = I Z_2 = 12 \angle -37.3^\circ \times 5.1 \angle 53.1^\circ$

$V_2 = 61.2 \angle -13.4^\circ$

$I_1 = \frac{V_1}{Z_1} = \frac{43.2 \angle 19^\circ}{3.6 \angle 56.3^\circ} = 12 \angle -37.3^\circ$

$I_2 = \frac{V_2}{Z_2} = \frac{61.2 \angle -13.4^\circ}{10 \angle -53.1^\circ} = 6.12 \angle 39.7^\circ$

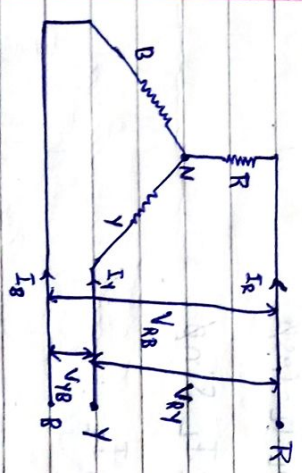
Generation of 3 phase EMF :-



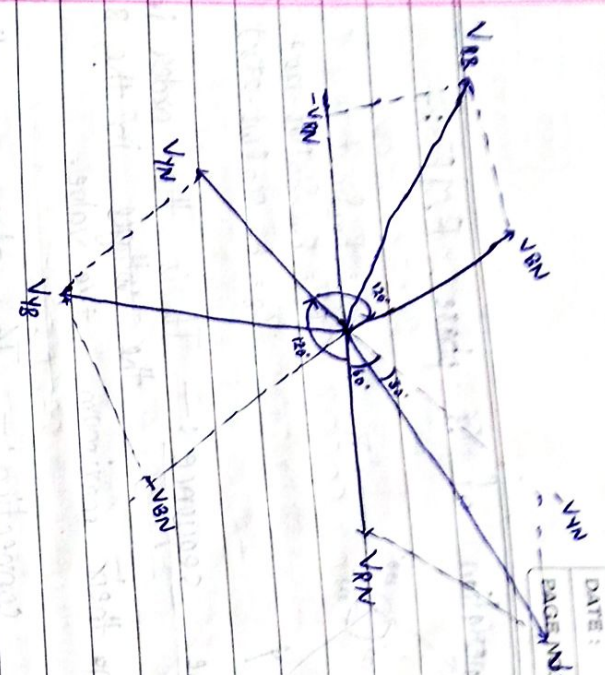
$E_R = E_m \sin \omega t$
 $E_Y = E_m \sin(\omega t - 120^\circ)$
 $E_B = E_m \sin(\omega t - 240^\circ)$

* Phase sequence :- It is the order in which the voltages in the 3 phase attain their maximum +ve value.

* Star connection :- In star connection the similar ends either star or finish of 3 windings are connected to a common point called 'neutral point'. And the 3 line conductors are run from the remaining three terminals called line conductor. The current flowing in through each phase is called phase current I_P , and the current flowing through each conductor is called line current I_L .



$I_{RN} = I_{BN} = I_{YN} = I_P$
 $I_R = I_Y = I_B = I_L$
 $V_{RN} = V_{BN} = V_{YN} = V_P$
 $V_{RY} = V_{YB} = V_{BR} = V_L$



$$V_{RY}^2 = V_{RN}^2 + V_{YN}^2 + 2|V_{RN}||V_{YN}|\cos 60^\circ$$

$$V_L^2 = V_p^2 + V_p^2 + 2V_p^2 \times \frac{1}{2}$$

$$V_L^2 = \frac{3V_p^2}{V_L = \sqrt{3}V_p} \quad [I_L = I_p]$$

$$P = 3V_p I_p \cos \phi$$

$$P = \sqrt{3} \times \sqrt{3}V_p I_p \cos \phi$$

$$P = \sqrt{3} V_L I_L \cos \phi$$

$$Q = \sqrt{3} V_L I_L \sin \phi$$

$$S = \sqrt{3} V_L I_L$$

8. A 3 phase star connected generator has a terminal voltage of 415 volt & deliver a full load current of 200 Amp at 0.8 power factor lagging. find
- Voltage per phase
 - full load current per phase
 - Real power deliver
 - impedance (per phase)

Ans (i) $V_p = \frac{415}{\sqrt{3}} = 239.6$ (ii) $I_L = I_p = 200 \text{ Amp}$

(iii) $S = \sqrt{3} V_L I_L = \sqrt{3} \times 415 \times 200 = 143100$ (iv) $P = \sqrt{3} \times 415 \times 200 \times 0.8 = 115000 \text{ W}$

(iv) $Z = \frac{V_p}{I_p} = \frac{239.6}{200} = 1.19 \angle 36.8^\circ$

(v) $Z = \frac{V}{I} = \frac{400}{230.9} = 1.73 \angle 36.8^\circ$

Q. A balanced star connected load of $8+j6$ ohm per phase is connected to a 3 phase 400 volt supply. find the line currents, power, vector, real power deliver and kVA rating.

Ans- $I_p = \frac{400}{\sqrt{3} \times 10} = 23.09 \angle 36.8^\circ$ $V_p = 230.9 \text{ V}$

$\theta = \tan^{-1}(\frac{6}{8}) = 36.8^\circ$ $P = \sqrt{3} V_p I_L \cos \phi = \sqrt{3} \times 400 \times 23.09 \times \frac{4}{5}$

$$I_p = 23.09 \angle -36.8^\circ$$

$$V_p = 230.9$$

$$\phi = 36.8^\circ$$

$$\cos \phi = 0.8$$

total power

$$P = \sqrt{3} \times 400 \times 23.09 \times 0.8$$

$$P = 12797.77 \text{ Watt}$$

$$P = 12.79 \text{ kWatt}$$

$$\theta = \frac{-\sqrt{3} \times 400 \times 23.09 \times 3}{5}$$

$$\theta = -9.59 \text{ KVAR}$$

$$Z = \frac{V}{I \angle 36.8^\circ} = \frac{230.9}{23.09 \angle 36.8^\circ} = 10 \angle -36.8^\circ$$

$$S = \sqrt{3} \times 400 \times 23.09 = 16.08 \text{ KVA}$$

$$V_R = 230.9 \angle 0^\circ$$

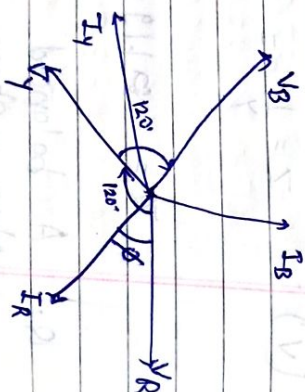
$$V_Y = 230.9 \angle -120^\circ$$

$$V_B = 230.9 \angle -240^\circ$$

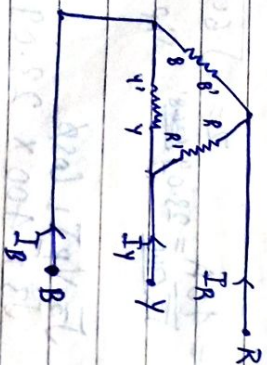
$$I_R = 23.1 \angle 36.8^\circ$$

$$I_Y = 23.1 \angle -156.8^\circ$$

$$I_B = 23.1 \angle -276.8^\circ$$

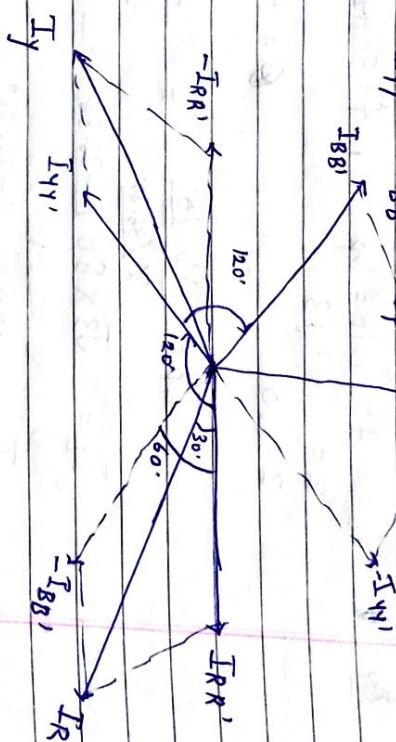


Delta connection:



$$I_R = I_Y = I_B = I_L$$

$$I_{RR'} = I_{YY'} = I_{BB'} = I_p$$



$$I_R = I_{RR'} - I_{BB'}$$

$$|I_R|^2 = |I_{RR'}|^2 + |I_{BB'}|^2 + 2|I_{RR'}||I_{BB'}|\cos 60^\circ$$

$$I_L^2 = I_p^2 + I_p^2 + 2 \times I_p \times I_p$$

$$I_L^2 = 3 I_p^2$$

$$I_L = \sqrt{3} I_p$$

$$V_L = V_p$$

Power:-

$$P = 3 V_p I_p \cos \theta = \sqrt{3} \times \sqrt{3} I_p V_p \cos \theta$$

$$P = \sqrt{3} V_L I_L \cos \theta$$

$$Q = \sqrt{3} V_L I_L \sin \theta$$

$$S = \sqrt{3} V_L I_L$$

Q. 3 coil of equal impedances of $4+j3\Omega$ are connected in Delta across 400 Volt, 50Hz supply. Find the phase current of each coil, line current, active, reactive &

apparent power?

$$\text{Qy- } V_L = 400 \text{ volt} = V_p, \quad f = 50 \text{ Hz}$$

$$Z = 4 + j3$$

$$Z = 5$$

$$\theta = \tan^{-1}\left(\frac{3}{4}\right) = 36.86^\circ$$

$$I_p = \frac{V_p \angle 0^\circ}{Z \angle 36.86^\circ}$$

$$= \frac{400}{5 \angle 36.86^\circ} = 80 \angle -36.86^\circ$$

$$I_L = \sqrt{3} \times 80 \angle -36.86^\circ$$

$$I_L = 138.56 \angle -36.86^\circ$$

$$P = \sqrt{3} V_L I_L \cos \theta$$

$$P = \sqrt{3} \times 400 \times 138.56 \times 0.8$$

$$P = 76.81 \text{ kWatt}$$

$$Q = \sqrt{3} \times 400 \times 138.56 \times \sin(-36.86^\circ)$$

$$Q = -\sqrt{3} \times 400 \times 138.56 \times 0.59$$

$$Q = -57.80 \text{ KVAR}$$

$$S = \sqrt{3} \times 400 \times 138.56$$

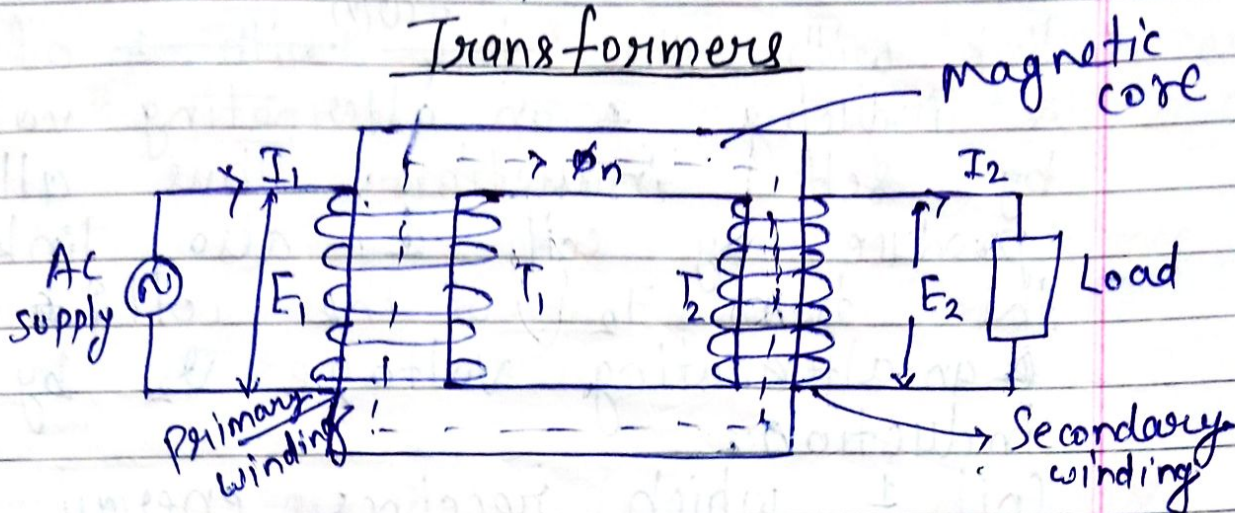
$$S = 96 \text{ KVA}$$

$$I_{RA'} = 80 \angle -36.86^\circ$$

$$I_{RB'} = 80 \angle -156.86^\circ$$

$$I_{BB'} = 80 \angle -276.86^\circ$$

Transformers

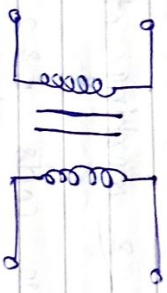


Transformer $\Rightarrow$ A transformer is a static device which consists of two or more stationary electric circuit interlinked by a common magnetic circuit for the purpose of transferring electric energy b/w them. The transfer of energy from one circuit to another takes place without a change in the frequency.

Construction & working of transformer $\Rightarrow$

Consider two coils 1 & 2 wound ^(एकसाथ) over on a simple magnetic circuit as shown in the figure about. These two coils are insulated ^(अलग) from each other and there is no electrical connection b/w them. Let T_1 & T_2 be the no. of ~~turn~~ turns in coils 1 & 2. Now when a source of alternating voltage V_1 is applied to the coil 1 and alternating current I_1 flows in it. This alternating current producing an alternating flux Φ_m in the magnetic circuit. This alternating flux

link with the ~~turn~~ with 1 of coil 1 & inducing an alternating voltage V_1 by self induction. Thus all the flux produce by coil 1 also links with the turn T_2 of the coil 2 and induces an alternating voltage V_2 by mutual induction. coil 1 which receives energy from the source of AC supply is called the "primary" which is connected to the load is and delivers energy to the load is called the "secondary" or simply electric energy is transfer from coil 1 to 2 by a common magnetic circuit.



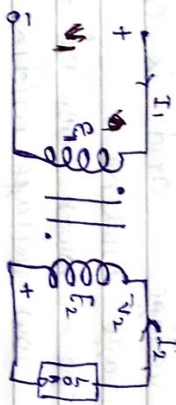
Ideal transformer $\Rightarrow$

An Ideal transformer is an imaginary transformer which has the following properties :-

- 1) Its primary winding & secondary winding resistances are negligible.
- 2) The core has infinite permeability. (given data)

3) Its leakage flux and leakage inductance are zero. So the entire flux is confined to the core and links with both the windings. Their are no losses due to resistance, hister hysteresis and eddy currents. So the efficiency is 100%.

Note - Practical Transformer has none of these properties inpite of fact that its operation is closed to the ideal transformer.



EMF Equation of transformer $\Rightarrow$

Let at any instant, the flux ϕ is.

The instantaneous emf is

$$e = -T \frac{d\phi}{dt}$$

$$e = -T \frac{d(\phi_m \sin \omega t)}{dt} = -T \phi_m \omega \cos \omega t$$

$$e = T \phi_m \omega \sin(\omega t - \frac{\pi}{2})$$

$$e = E_m \sin(\omega t - \frac{\pi}{2})$$

For sin wave $E_{rms} = \frac{E_m}{\sqrt{2}} = \frac{T \phi_m \omega}{\sqrt{2}}$

$$E_{rms} = \frac{T \phi_m \times 2\pi f}{\sqrt{2}}$$

$E = 4.44 T \phi_m f$

$E_1 = 4.44 T \phi_m f$

$E_2 = 4.44 \phi_m f T_2$

$B_m = \frac{\phi_m}{A}$

$E = 4.44 \phi_m T = \text{constant}$

Note:- The winding with higher no. of turns will have a high voltage winding. The winding with the lower no. of turns is called the low voltage winding.

Stepup transformer:- A transformer in which the output voltage is greater than its input voltage is called a "stepup transformer".

Stepdown transformer:- A transformer in which the output voltage is less than its input voltage is called a "stepdown transformer".

Note:- A transformer may receive energy at 1 voltage and deliver it at the same voltage such a transformer is called one to one transformer. For a one to one transformer

$E_1 = E_2$

$I_1 = I_2$

Voltage ratio & Turns ratio:-

$E_1 = 4.44 \phi_m f T_1$

$E_2 = 4.44 \phi_m f T_2$

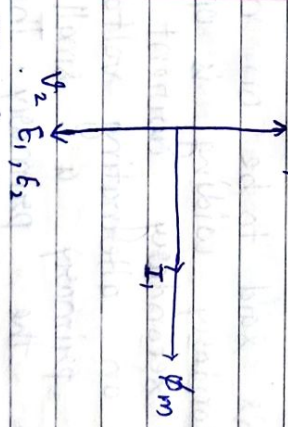
$\frac{E_1}{E_2} = \frac{T_1}{T_2} = a = \text{Transformation ratio}$

$m m f = I_1 T_1 = I_2 T_2$

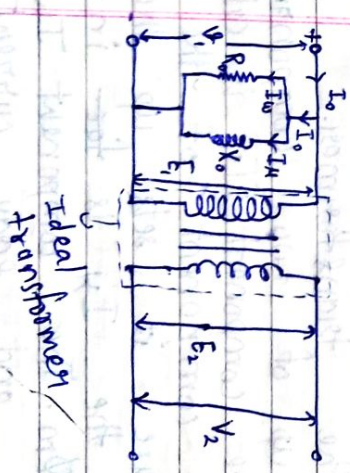
$\frac{I_1}{I_2} = \frac{T_2}{T_1}$

$\frac{E_1}{E_2} = \frac{T_2}{T_1} = \frac{I_2}{I_1}$

No-load phasor diagram of Ideal transformer



Transformer on NO-Load $\Rightarrow$

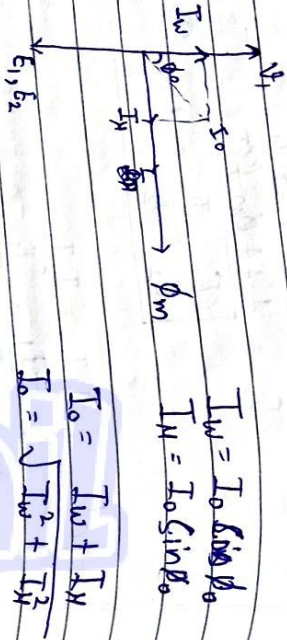


So from the equivalent circuit, $V_1 = R_0 I_w + V_2$

$I_0 = I_w + I_2$

where I_0 is the no-load current

I_W = Reactive component of the no-load current
 I_W = Active component of the no-load current



$P = V_1 I_0 \cos \phi$ (active power)
 $Q = V_1 I_0 \sin \phi$ (reactive power)

$\cos \phi_0 = \frac{I_W}{I_0}$

A transformer is said to be on no-load when the secondary winding is open circuited. So the secondary current becomes zero. When an alternating voltage V_1 is applied to the primary a small current I_0 flows in the primary. This current I_0 is called the no-load current of the transformer.

It is made up of two components

I_W & I_W . The component I_W is called the reactive component of the no-load current because it sets up the flux in core. So the I_W is in phase with the flux.

The component I_W supply the hysteresis and eddy current losses

in the core and known as active component of the no-load current. It is in phase with the applied voltage V_1 .

For a no-load transformer E_1 & E_2 are induced by the same flux ϕ_m so they will be in phase with each other and lags behind ϕ by 90° . According to Lenz law E_1 is equal & opposite to V_1 . I_W is in phase with ϕ_m and I_W is in phase with V_1 . The phase sum of I_W & I_W is I_0 . $\cos \phi_0$ is the power factor on no-load and ϕ_0 is the phase angle on no-load.

Q. Find the active & reactive component of no-load current and no-load current of a 440/220 volt single phase transformer. is the input power to the high voltage winding is 80W. The low voltage of the winding is open & power factor of the current is 0.3 lagging

$P_0 = V_1 I_0 \cos \phi$
 $\cos \phi_0 = 0.3$, $V_1 = 440V$

$\frac{240}{220} = 80 \times I_0 \times 0.3$

$\frac{2 \times 10^3}{220} = \frac{2}{24} = I_0 \Rightarrow I_0 = \frac{2}{0.083} = 0.083A$

$I_W = I_0 \cos \phi = 0.083 \times 0.3 = 0.0249$
 $P_0 = V_1 I_0 \cos \phi$

$80 = 440 \times I_0 \times 0.3$
 $I_0 = \frac{8 \times 10}{44 \times 3} = \frac{80}{132} = \frac{20}{33} = 0.60A$

$I_0 = 0.60A$

$$I_w = I_o \cos \phi = 0.60 \times 0.3 = 0.18 \text{ A}$$

$$I_w = 0.18 \text{ A}$$

$$\sqrt{I_o^2 - I_w^2} = I_H$$

$$\sqrt{0.36 - 0.032} = I_H$$

$$I_H = 0.2$$

$$\sqrt{0.928}$$

$$I_H = 0.578 \text{ A}$$

Equivalent circuit of a transformer :-

Referred value $\Rightarrow$

$$R_2 \rightarrow R_2'$$

$$\frac{I_2'^2 R_2'}{I_2^2} = \frac{I_2^2 R_2}{I_2^2}$$

$$R_2' = \left(\frac{I_2}{I_1}\right)^2 R_2$$

$$\frac{0.360}{0.032}$$

$$\frac{0.928}{0.928}$$

$$R_2' = \left(\frac{I_1}{I_2}\right)^2 R_2$$

$$R_2' = a^2 R_2$$

$$X_2' = a^2 X_2$$

$$Z_2' = a^2 Z_2$$

Let R_e , X_e and Z_e be the effective resistance reactance & impedance of the whole transformer referred to the primary

$$R_e = R_1 + R_2'$$

$$R_e = R_1 + a^2 R_2 \quad \text{--- (1)}$$

$$X_e = X_1 + a^2 X_2$$

[Signature]

$$Z_e = Z_1 + Z_2'$$

$$Z_e = Z_1 + a^2 Z_2$$

$$Z_e = \sqrt{R_e^2 + X_e^2} = R_e + jX_e$$

Equivalent values referred to the secondary $\Rightarrow$

Let R_2 , X_2 & Z_2 be the effective resistance reactance & impedance of the whole transformer referred to the secondary.

$$\left\{ \begin{aligned} I_1^2 R_1 &= I_2^2 R_1' \\ R_1' &= \left(\frac{I_1}{I_2}\right)^2 R_1 = \frac{R_1}{a^2} \end{aligned} \right\}$$

$$R_{e2} = R_2 + R_1'$$

$$R_{e2} = R_2 + \frac{R_1}{a^2} \quad \text{--- (2)}$$

$$Z_{e2} = Z_2 + \frac{Z_1}{a^2}$$

$$Z_{e2} = R_{e2} + jX_{e2}$$

$$Z_{e2} = \sqrt{R_{e2}^2 + X_{e2}^2}$$

from eqn- (1)

$$R_{e1} = R_1 + R_2 \left(\frac{I_2}{I_1}\right)^2$$

$$R_{e1} \left(\frac{I_2}{I_1}\right)^2 = R_1 \left(\frac{I_2}{I_1}\right)^2 + R_2$$

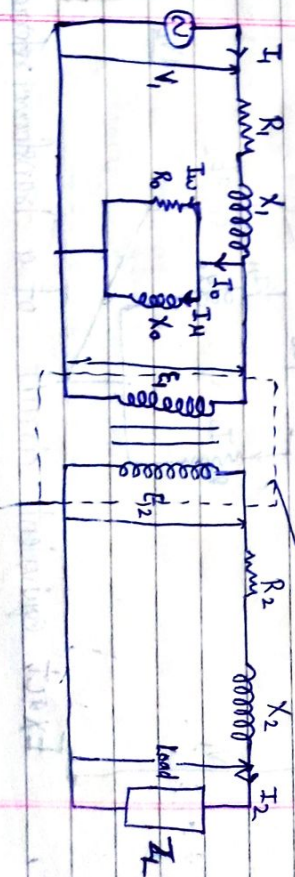
from eqn- (2)

$$\frac{R_{e1}}{a^2} = R_{e2}$$

$$X_{e2} = \frac{X_{e1}}{a^2}$$

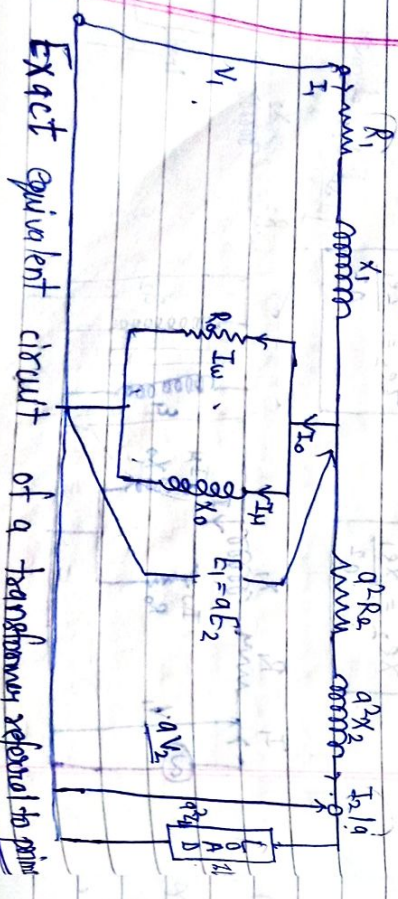
$$Z_{e2} = \frac{Z_{e1}}{a^2}$$

Ideal transformer

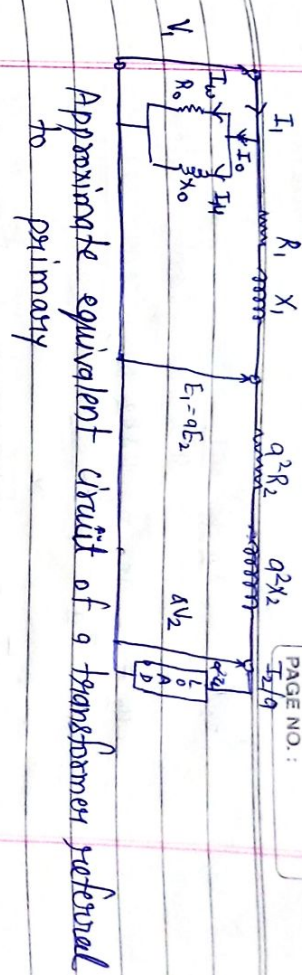


The above figure shows the complete equivalent circuit of a transformer and exact equivalent circuit referred to the primary can be deduced as follows:-

- (i) All secondary resistances, reactances & impedances are reflected to the primary as a multiple of the square of the transformation ratio. i.e. R_2, X_2 & Z_2 in the secondary becomes $a^2 R_2, a^2 X_2$ & $a^2 Z_2$ in the primary and reflected from secondary to primary directly as the product of the transformation ratio that is V_2 & E_2 in the secondary becomes aV_2 & aE_2 when referred to the primary. All the secondary currents are reflected to the primary as inversely of the transformation ratio i.e. I_2 in the secondary becomes I_2/a when referred to primary.
- 2.) All voltages are reflected from secondary to primary directly as the product of the transformation ratio that is V_2 & E_2 in the secondary becomes aV_2 & aE_2 when referred to the primary.
- 3.) All the secondary currents are reflected to the primary as inversely of the transformation ratio i.e. I_2 in the secondary becomes I_2/a when referred to primary.



Exact equivalent circuit of a transformer referred to primary



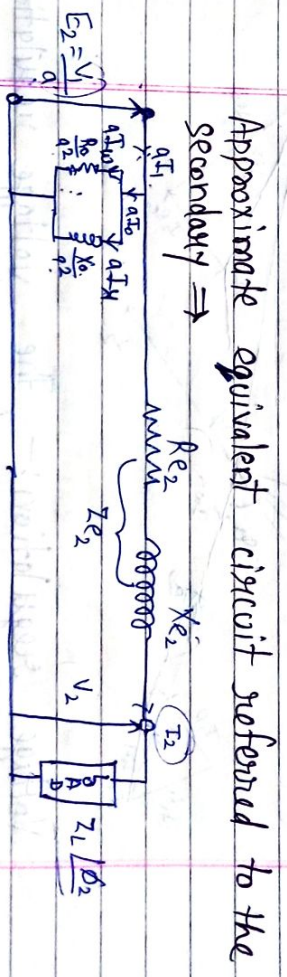
Approximate equivalent circuit of a transformer referred to primary

This approximation gives the following advantages:-

- 1) The primary & secondary impedances referred to the primary can be added easily.
- 2) The calculations for voltage regulation of the transformer become easier.

The following result are obtain from the approximate equivalent circuit.

(i) $R_e = R_1 + a^2 R_2$
 $X_e = X_1 + a^2 X_2$
 $Z_e = R_e + jX_e$
 $I_1 = \frac{V_1}{Z_e + a^2 Z_L}$ where Z_L is the load impedance.



Approximate equivalent circuit referred to the secondary

The following results are obtain from this circuit

$R_e = R_2 + \frac{R_1}{a^2}$

$$X_{eq} = X_2 + \frac{X_1}{a^2}$$

$$Z_{eq} = R_{eq} + jX_{eq}$$

$$I_2 = \frac{V_2}{Z_L}$$

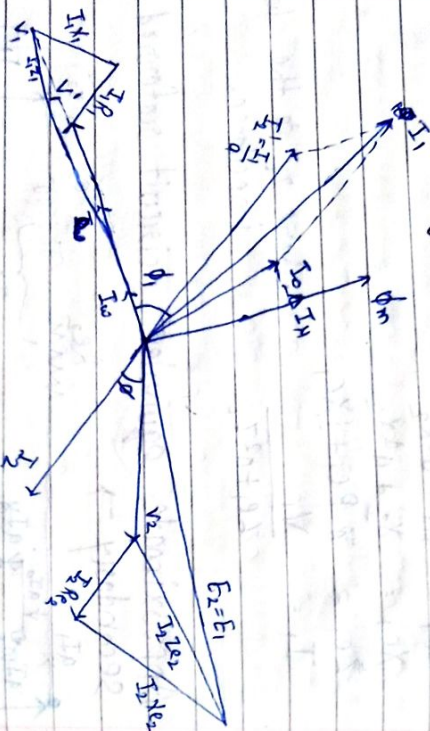
$$I_2 = \frac{E_2}{Z_{eq} + Z_L}$$

$$E_2 = \frac{V_1}{a}$$

$$E_2 = I_2 Z_{eq} + I_2 Z_L$$

$$\left[\frac{V_1}{a} = E_2 = V_2 + I_2 Z_{eq} \right]$$

Phasor Diagram $\Rightarrow$



Voltage regulation:- The voltage regulation of a transformer is defined as the percentage difference in the secondary terminal voltage b/w no load & full rated load.

At a given power factor with a same

Value of primary voltage for both no load & 'rated load'. It is expressed as either per unit or a percentage of the rated load voltage.

Per-unit voltage regulation at full load

$$= \frac{|V_{no1}| - |V_{fl1}|}{|V_{fl1}|} \quad |V_{fl1}| = \text{constant}$$

Percentage voltage regulation at full load

$$= \frac{|V_{no1}| - |V_{fl1}|}{|V_{fl1}|} \times 100$$

$|V_{fl1}| = \text{constant}$

Where V_{no1} is the no load secondary terminal voltage & V_{fl1} is the rated secondary terminal voltage.

Voltage regulation in terms of primary

$$V_{no1} = \frac{V_1}{a}$$

$$= \frac{\left| \frac{V_1}{a} \right| - |V_{fl1}|}{|V_{fl1}|}$$

Transformer efficiency $\Rightarrow$ The ratio of the output power to the input power in a transformer is known as transformer efficiency.

$$\eta = \frac{\text{Output Power}}{\text{Input Power}} \quad (\text{per unit})$$

$$\eta = \frac{\text{Output power}}{\text{Output power} + \text{Losses}}$$

$$\eta = \frac{V_2 I_2 \cos \phi_2}{V_2 I_2 \cos \phi_2 + Cu \text{ loss} + \text{iron loss}}$$

$$\eta = \frac{V_2 I_2 \cos \phi_2}{V_2 I_2 \cos \phi_2 + I_2^2 R_{e2} + P_i} \quad (\text{Kunth})$$

Losses in transformer $\Rightarrow$ The losses which occur in a transformer

- 1) ~~formation~~ core or core loss (P_i)
- 2) Copper loss or $I^2 R$ loss (P_e)

(1) Iron loss:— Iron loss occurs in the magnetic core of the transformer. This

loss is the sum of hysteresis loss P_h & eddy current loss P_e so

$$P_i = P_h + P_e$$

where $P_h = k_h f B_m^x$
 $P_e = k_e f^2 B_m^2$

k_h = constant of proportionality which depends upon the volume and quality of the core material.

k_e = constant of proportionality which depends upon the volume & resistivity of the core material.

B_m = max. flux density in the core.

f is the frequency of the alternating flux and the exponent 'x' is called 'Steinmetz constant'. Its value varies from 1.5 to 2.5 depending upon the magnetic properties of the core material. So the total core loss can be written as

$$P_i = k_h f B_m^x + k_e f^2 B_m^2$$

$$V_1 = E_1 = 4.44 \phi_m f T_1$$

$$V_1 = E_1 = 4.44 B_m A_f f T_1$$

$$B_m = \frac{V_1}{4.44 A_f f T_1}$$

$$P_h = k_h f B_m^x$$

$$P_h = k_h f \left(\frac{V_1}{4.44 A_f f T_1} \right)^x$$

$$P_h = \frac{k_h V_1^x}{4.44^x A_f^x T_1^x} f^{1-x}$$

$$P_h = \frac{k_h V_1^x f^{1-x}}{K_h (4.44 A_f T_1)^x}$$

$$K_h = k_h \left(\frac{1}{4.44 A_f T_1} \right)^x$$

This relation shows that the hysteresis loss depends upon both the applied voltage and frequency. Similarly

$$P_e = k_e f^2 B_m^2$$

$$P_e = k_e f^2 \left(\frac{V_1}{4.44 A_f f T_1} \right)^2$$

$$P_e = \frac{k_e V_1^2}{(4.44 A_f T_1)^2} f^2 \quad \left\{ \begin{aligned} K_e &= \frac{k_e}{(4.44 A_f T_1)^2} \end{aligned} \right.$$

$$P_e = K_e V_1^2$$

This relation shows that the eddy current loss depends on the square of the applied voltage and is independent of frequency.

The total core loss is

$$P_i = K_h V^\alpha f^{1-\alpha} + K_e V^2$$

2.) Copper loss $\Rightarrow$ Copper loss is the I^2R loss which takes place in the primary and secondary windings because of the winding resistances.

Total copper loss in a transformer

= primary winding copper loss + secondary winding copper loss

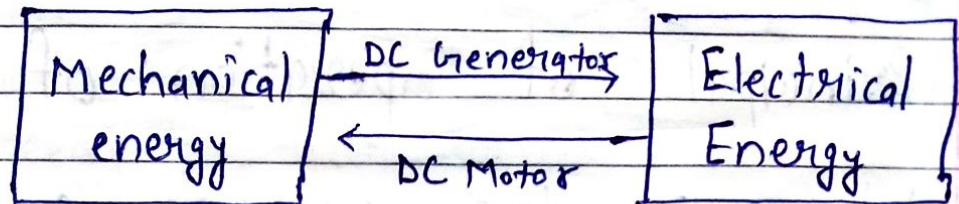
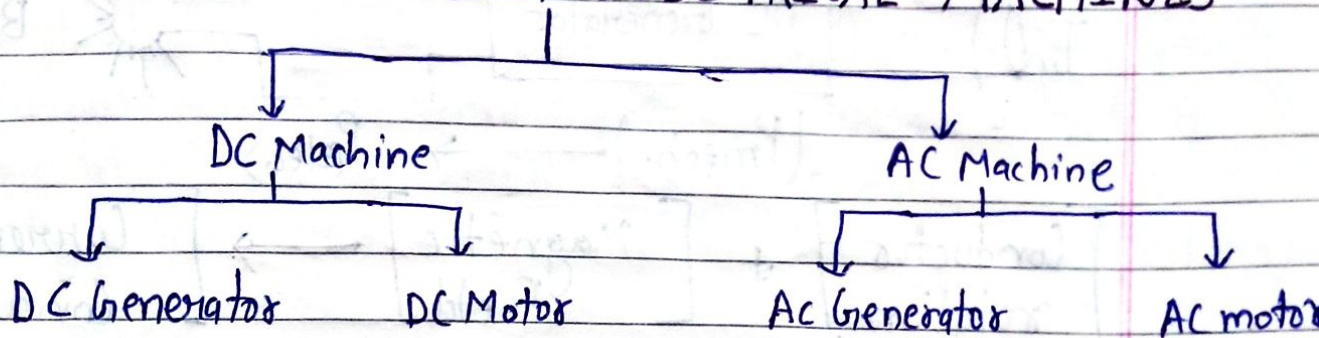
$$P_c = I_1^2 R_1 + I_2^2 R_2$$

or

$$P_c = I_1^2 R_{e1}$$

$$P_c = I_2^2 R_{e2}$$

ROTATING ELECTRICAL MACHINES



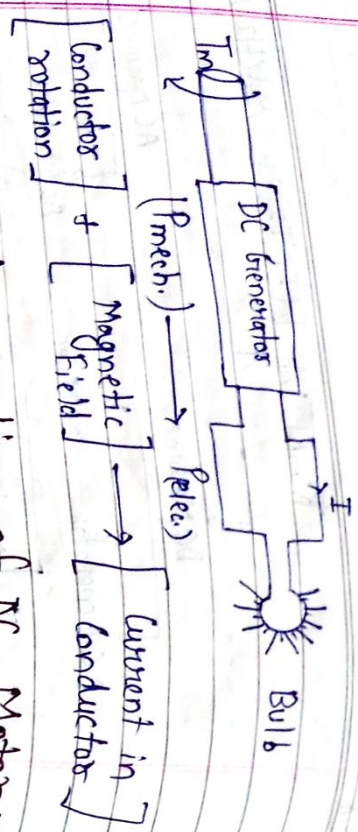
The rotating electrical machines are the electro mechanical ~~electro~~ energy conversion devices that converts mechanical energy into electrical energy and electrical energy into mechanical energy. There are two types of electrical machine—

1) DC Machine

2) AC Machine

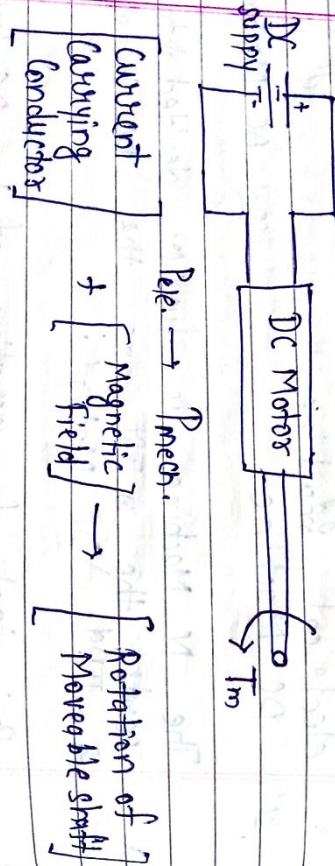
The DC Machine is further divided into two parts the 1st one the 'DC Generator' & 2nd 'DC Motor'.

Principal of operation of a DC Generator
⇒ If an externally applied force makes the conductor to move into the magnetic field. The current starts to flow in the conductor. Thus, converting electrical mechanical energy into electrical energy. The figure shows the operating principal of the DC Generator.



Principal of operation of DC Motor $\Rightarrow$

When ever the current carrying conductor placed in the magnetic field it experiences a force that tends to move it. Therefore converting the electric energy into mechanical energy.



NOTE $\Rightarrow$ DC Motor follows the Fleming's left hand.

2) DC Generator follows the Fleming's Right hand Rule.

EMF Eqⁿ of DC Machine $\Rightarrow$ In generator

the armature rotates an emf is generated across its conductor and

this emf is called generating emf. (E_g). Similarly Motor due to the armature rotation and emf is induced which is known as back emf. (E_b)

Let us consider ϕ = Flux per pole (Wobers)
P = No. of poles
Z = Total no. of armature conductor,
A = No. of parallel paths through armature
N = Speed of armature (rpm)

During one revolution of armature in a P pole generator each armature conductor cuts the magnetic flux P-times. So the flux cut by one conductor in one revolution is equal to $P\phi$.

Now the No. of revolution made by the armature per second is $N/60$. So the average induced emf in one conduction is equal to

$$E = \frac{P\phi N}{60} \text{ Volts}$$

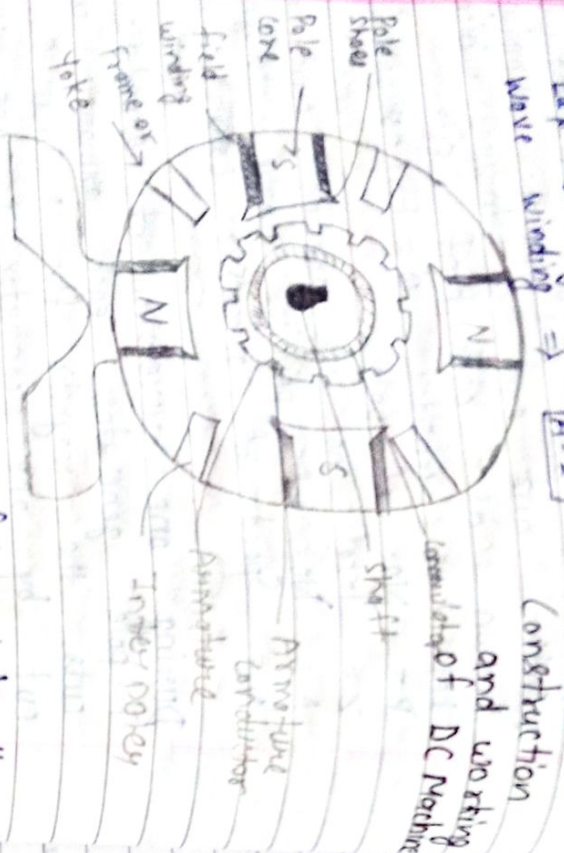
If the number of armature conductor per parallel path is equal to $\frac{Z}{A}$. so the total

emf generated b/w the terminals is equal to $E = (\text{Avg. induced emf in one conductor})$

$\times$ (No. of conductor per parallel path)

$$E = \phi \frac{P N}{60} \cdot \frac{Z}{A} \text{ Volt}$$

Lap winding $\Rightarrow \left[\frac{P=A}{A=2} \right]$
Wave winding $\Rightarrow$



Construction and working of DC Machine

A DC machine consists of mainly three parts:-
(i) Magnetic field system
(ii) Armature
(iii) Commutator & brush gear

(1) Magnetic field system $\Rightarrow$ The magnetic field system is the stationary part of the machine. The magnetic field system consists of different parts:-

(a) Frame or yoke :- The outer frame or yoke is a hollow cylinder made up of cast steel. The yoke serves the following two purposes

(i) It supports the pole core and actual the protecting cover for machine.

(2) Its pole forms a part of magnetic circuit.

(b) Pole core :- An even no. of pole cores are bolted to the yoke. Since the poles project inwards they are called Salient poles.

(c) Pole shoes $\Rightarrow$ Each pole core has a pole shoe having concave surface and serves the following two purposes :-
(i) It supports the field windings
(ii) It increases the cross sectional area of the magnetic circuit and reduces its reluctance.

(d) Field windings $\Rightarrow$ Each pole core has one or more field windings placed over it to produce the magnetic field in the machine.

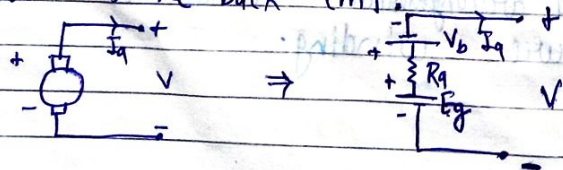
(2) Armature $\Rightarrow$ The rotating part of the DC Machine is called armature. The armature consists of a shaft upon which a laminated cylinder called armature core is mounted. The armature core has slots on its outer surface. The insulated conductors are put in the slots of the armature core. And these conductors are suitably connected. This connected arrangement of conductors is called armature winding.

3.) Commutator and brush gear $\Rightarrow$ Alternating voltage is produced in the windings rotating in a magnetic field. To obtain the direct current in the external circuit a commutator is needed. This commutator is required in the system to convert the AC into DC & DC into AC supply. It is made up of no. of waved shape, copper brass insulator from each other and from the shaft.

Brushes:- Current is collected from the armature winding by means of two or more carbon brushes mounted on the commutator. Each brush is supported in a metal box called the brush box or brush holder.

Equivalent circuit of a DC Machine:- The armature of a DC Machine can be represented by an equivalent electric circuit. It can be represented by 3 series connected elements E_g , R_a & V_b . The element ' e ' is generated voltage. R_a is armature resistance & V_b is the brush contact voltage drop. The equivalent circuit of the armature of a generator & of a DC motor is shown by the figure below. In case of the DC motor e is the back emf.

DC Generator



DC motor:-

In case of generator, current flows from armature to line
 $V = E_g - I_a R_a - V_b \Rightarrow V = E_g - I_a R_a$ for fundamental generator

In case of motor, current flows from line to armature
 $V = E_g + I_a R_a + V_b \Rightarrow V = E_g + I_a R_a$ for fundamental motor

Back emf:- When the motor armature rotate its conductor cut the magnetic flux there fore the emf of rotation E_b is induced in them and in case of the motor this emf of rotation is known as "back emf" or "counter emf" because it opposes the apply voltage.

Types of DC Machine $\Rightarrow$
 DC Machine

Separately excited

Self excited

Shunt wound

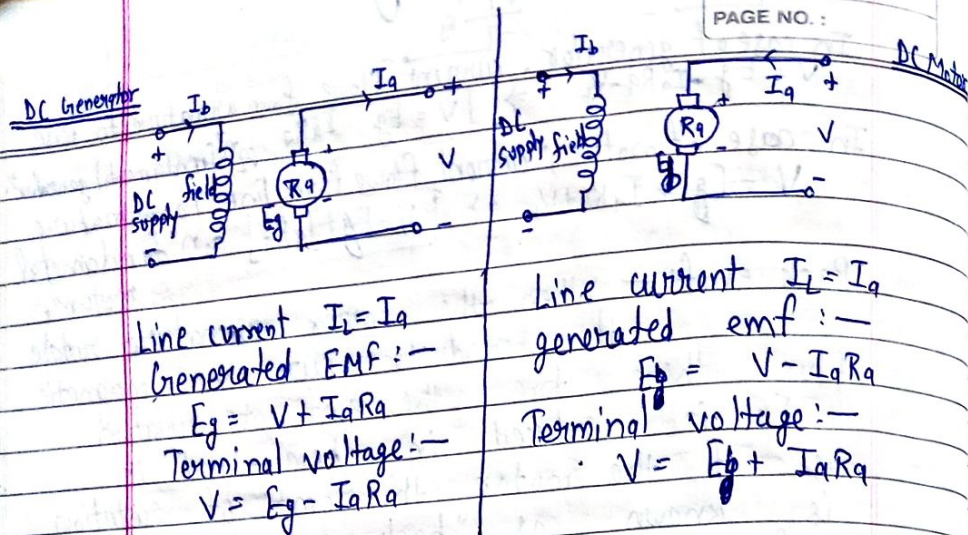
Series wound

Compound wound

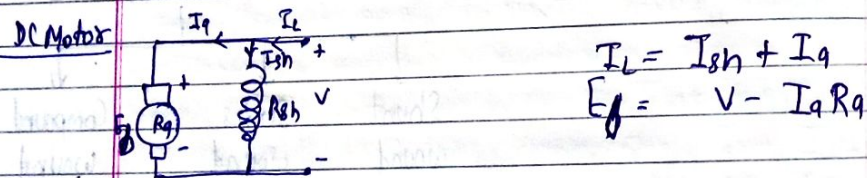
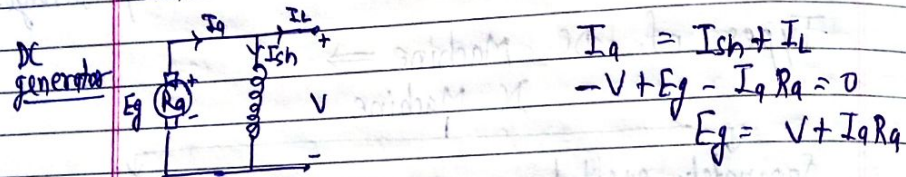
Short Shunt

Long Shunt

1.) Separately excited DC Machine $\Rightarrow$ When the field windings are energized by a separate DC source the connection is called separately excited DC Machine.

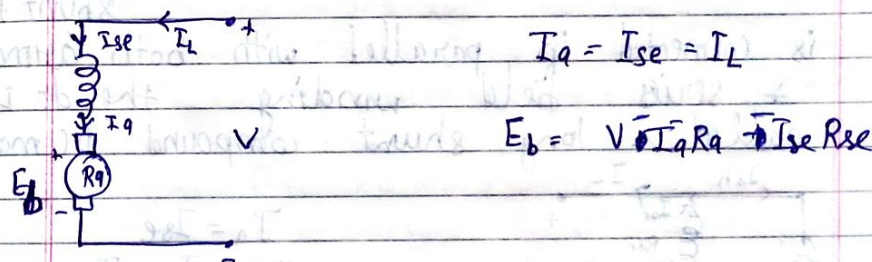
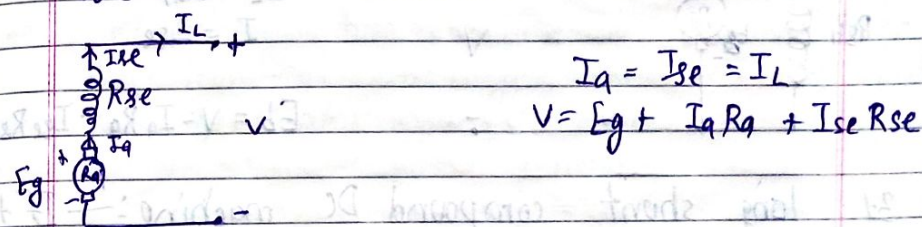


Shunt (parallel) Wound DC Machine:-



A machine in which the field coils are connected in parallel with armature called a shunt wound dc machine. Since the shunt field receives the full output voltage of a generator or the supply voltage of a motor it is generally made up of large no. of turns of a fine wire.

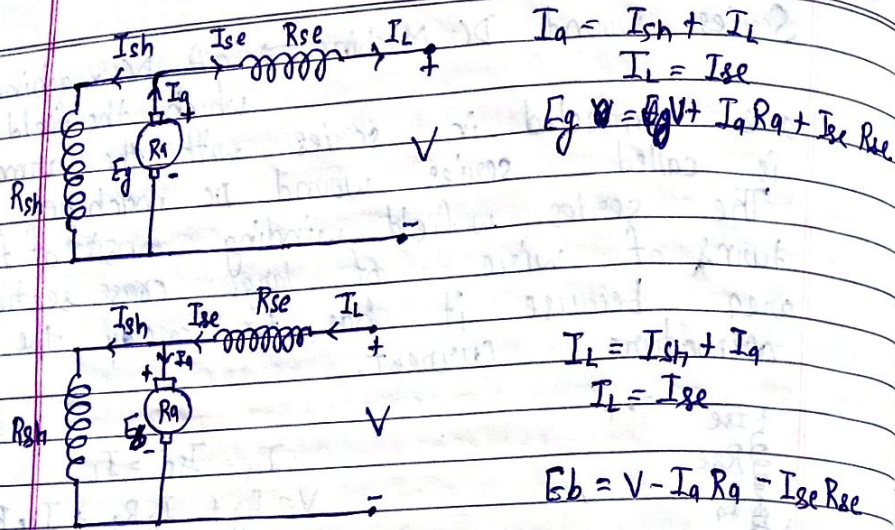
Series wound DC Machine $\Rightarrow$ A DC machine in which the field coils are connected in series with the armature is called series wound DC machine. The series field winding consists of few turns of wire of large cross sectional area because it has to carry the large armature current.



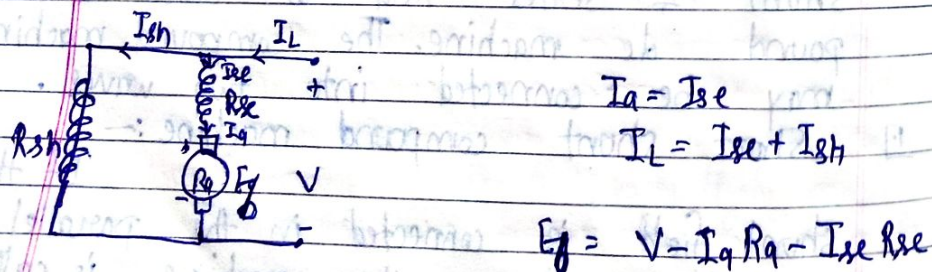
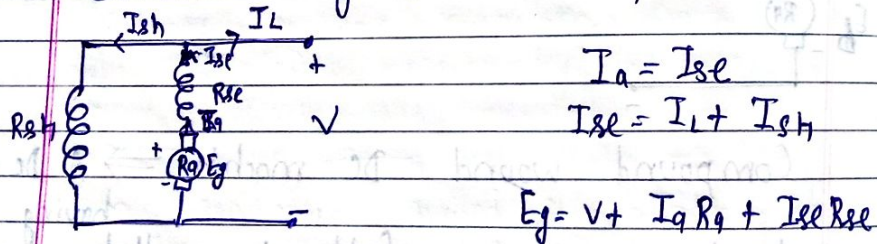
Compound wound DC machine $\Rightarrow$ A DC machine having both shunt & series field is called a compound dc machine. The compound machine may be connected into two ways.

1) Short shunt compound machine:-

If the shunt field is connected in the parallel with armature along the machine is called short shunt compound machine.



2.1 long shunt compound DC machine:— If the shunt field is connected in parallel with both armature & series field winding the dc is called long shunt compound DC machine.



Q. A 4 pole 900 rpm dc machine has a terminal voltage of 220 Volt & induced voltage of 240 Volt. At rated speed the armature resistance is 0.2 Ω . Is the machine operating as a generator or a motor? Compute the armature current & the no. of armature conductor if the flux per pole is 10 mWb & the armature wave wound.

Note— When induced voltage is greater than the terminal voltage the machine working as generator and vice versa.

Ans— $V_t < V_i$ so generator

$$E_g = V + I_a R_a$$

$$240 = 220 + I_a (0.2)$$

$$\frac{20}{0.2} = I_a \Rightarrow \boxed{I_a = 100 \text{ A}}$$

$$E = \frac{P \Phi N}{60} \times \frac{Z}{A}$$

$$240 = \frac{4 \times 10 \times 10^{-3} \times 900}{60} \times \frac{Z}{2} \quad (\because \text{Wave winding } A=2)$$

$$\frac{24 \times 8 \times 100 \times Z}{2 \times 8 \times 100 \times 10^7} = Z$$

$$\boxed{Z = 8 \times 10^{-7}}$$

Speed of a DC Machine $\Rightarrow$ The emf equation of a DC machine is given by

$$E = \frac{NP\Phi Z}{60 \times A}$$

Now, solving of $N = \frac{E \times 60}{P Z \phi}$

$$N = \frac{E}{k \phi}$$

where

$$k = \frac{P Z}{60 A}$$

k = machine const.

E = induced emf

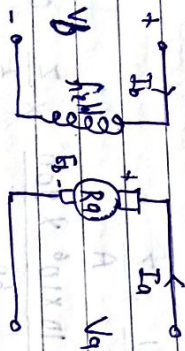
ϕ = flux per pole & field flux

In case of generator

$$N = \frac{V + I_a R_a}{k \phi}$$

motor: $N = \frac{V - I_a R_a}{k \phi}$

Torque speed characteristic of a DC Motor



In separately excited DC motor the field windings are excited from a separate DC source. So the motor back emf or speed voltage is given by E_b & by applying KVL, we get $E_b = V_a - I_a R_a$

And the armature induced voltage with motor rotating at a speed ω is given by $E_b = k \phi \omega$

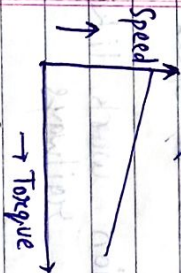
where ϕ is the field flux & k is machine constant. Similarly the motor torque is given by $T = k \phi I_a$

So from the above two equations the expression for speed ω in terms of torque T can be obtained as

$$\omega = \frac{V_a}{k \phi} - \frac{R_a}{(k \phi)^2} T$$

Now if we plot the torque speed characteristic we found that as the torque developed is increased the speed will fall with the gradient of $\frac{R_a}{(k \phi)^2}$.

The torque speed characteristic is shown below:-



Q. A 4 pole generator having wave wound armature winding has 51 slots and each slot containing 20 conductor. What will be the generated voltage when the machine is driven at 1500 rpm & flux per pole is 7 milli weber.

Ans-

$$E = \frac{NP \phi}{60} \times \frac{Z}{A}$$

$$E = \frac{500 \times 4 \times 7 \times 10^{-3}}{60} \times \frac{1020}{3570} = 3570 \text{ V}$$

Application of DC machine:-

DC generator :- 1) DC generators are used for light and power supply in hospitals.

2) It is used for charging batteries.

3) It is used for arc lighting & the belting.

4) It is used for dc locomotive for regenerative braking.

DC Motor:- 1) It is used for the production of high starting torque.

2) It is used in steel rolling mills, paper mills, etc.

3) It is used where very accurate speed is required.

4) It is used for traction work like lifts, trains & in electric railways.

Speed control of a DC Motor $\Rightarrow$ The speed of a DC

motor is given by the relationship

$$N = \frac{V - I_a R_a}{k \phi}$$

The above eqⁿ shows that the speed is depended upon the supply voltage 'V', the armature resistance 'R_a' & the field flux ϕ in practice the variation of these 3 factors is used for speed control. Thus, there are 3

general method of speed control of DC motor :- (i) Variation of resistance in the armature circuit $\Rightarrow$ This method is called armature resistance control.

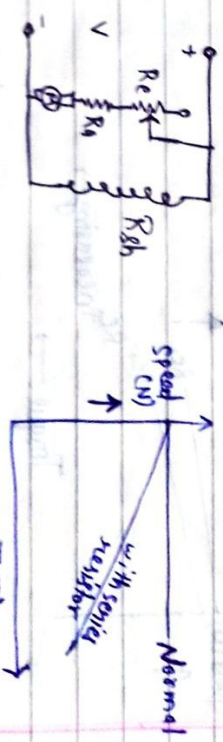
(ii) Variation of field flux $\phi \Rightarrow$ This method is called field flux control.

(iii) Variation of applied voltage $\Rightarrow$ This method is called armature voltage control.

(i) Armature resistance control :- In this method a variable series resistor R_e is put in the armature circuit.

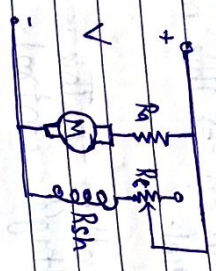
In this case the field is directly connected across the supply and flux ϕ is not affected by variation of R_e.

The voltage drop in R_e reduces the voltage applied to the armature & therefore the speed is reduced. This method is only used for small motors.



(ii) Field flux control $\Rightarrow$ Since the flux is produced by the field current the control of speed by this method is obtained by controlling the field current. In this

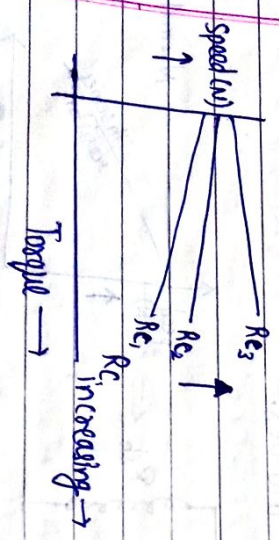
Shunt motor this is done by connecting a variable resistor R_c in series with the shunt field winding. The resistor R_c is called the shunt field regulator.



The shunt field current is given by

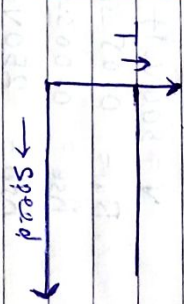
$$I_{sh} = \frac{V}{R_{sh} + R_c}$$

The connection of R_c in the field reduces the field current and hence the flux ϕ is also reduced. The reduction in the flux will result in an increase in the speed. Consequently the motor runs at a speed higher than the normal speed. For this reason the method of speed control is used to give motor speed above the normal.



(iii) Armature voltage control $\Rightarrow$ Speed control of DC motor can also be obtained by varying the applied voltage to the armature. This method

of speed control is based on the principle of Ward / Leonard system. Since the speed control is carried out with the excited current I_a & with constant motor field flux ϕ_m . A constant torque upto the base speed is obtained. Hence, with the armature voltage control method constant torque is obtained from the speed below the base speed.



Q. DC shunt machine connected to 250 volt supply has an armature resistor 0.2- Ω & the resistance of the field circuit is 100- Ω . Find the ratio of speed as a generator to the speed as a motor, the line current in each case is 80A.

Sol- $N_1 = \frac{V + I_a R_a}{K \phi} = \frac{250 + 80 \times 0.2 + 100}{K \phi}$

$N_2 = \frac{V - I_a R_a}{K \phi} = \frac{250 - 80 \times 0.2}{K \phi}$

$\frac{N_1}{N_2} = \frac{V + I_a R_a}{V - I_a R_a}$

$$I_{sh} = \frac{V}{R_{sh} + R_c} = \frac{250}{0.12 + 100} = \frac{25000}{100.12}$$

Method $I_a = I_L = I_{sh} = 80 - 2.5 = 77.5A$

generator $I_a = I_L + I_{sh}$

8. A long shunt compound generator delivers a load current of 50 A at 500 V. It has a series field of 0.003 Ω and shunt field of 250 Ω. Calculate the generated voltage & the armature current.

Sol.

And have armature resistance of 0.05 Ω
 series field of 0.003 Ω. calculate the generated voltage & the armature current.

$$I_a = I_L + I_{sh}$$

$$I_a = 50 + 2 = 52 \text{ A}$$

$$V = 500 \text{ V}$$

$$R_a = 0.05 \Omega$$

$$R_{se} = 0.003 \Omega$$

$$R_{sh} = 250 \text{ V}$$

$$E_g = V + I_a R_a + I_a R_{se}$$

$$I_{sh} = \frac{V}{R_{sh}} = \frac{500}{250} = 2 \text{ A}$$

$$E_g = 500 + 52 \times 0.05 + 52 \times 0.003$$

$$500 + 2.6 + 0.156$$

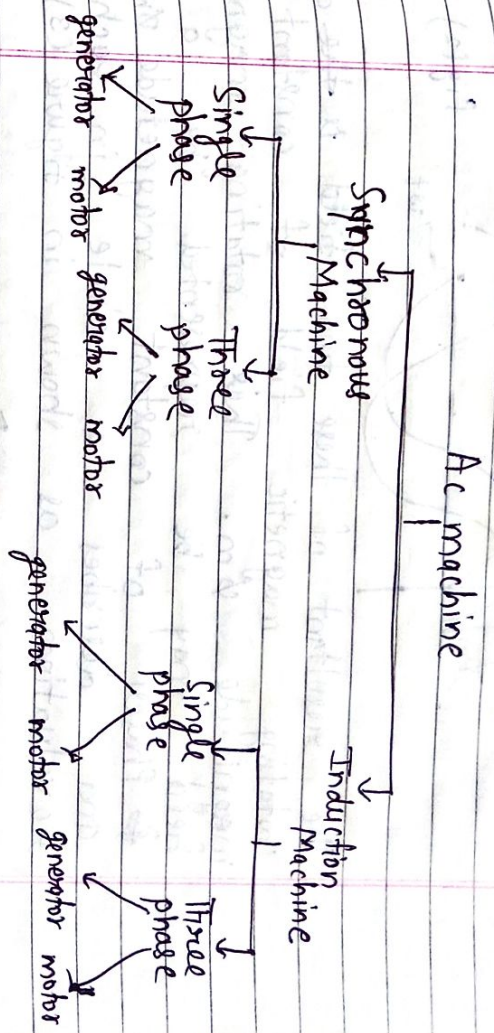
$$E_g = 502.756 \text{ V}$$

260
 53
 156

Speed regulation $\Rightarrow$ It is defined as the change in speed from full load to the no-load and it is expressed as a percentage of the full load speed

$$\% \text{ Speed regulation} = \frac{N_0 - N_f}{N_f} \times 100$$

where N_0 = no load speed
 N_f = full load speed of the motor



(i) Single phase induction motor $\Rightarrow$

(9) Production of rotating field: — Consider two windings A & B and they are displaced by 90° in space. Suppose that these windings produce magnetic field equal in magnitude & 90° apart in time as shown in figure (i). So

$$\phi_A = \phi_m \sin \omega t$$

$$\phi_B = \phi_m \sin (\omega t + \omega t)$$

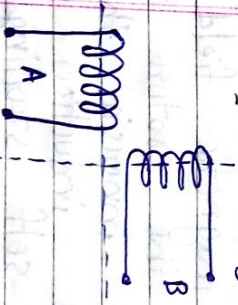


Figure (1)

The waveforms of the field are shown in the figure (2).



The resultant of these two field is a rotating magnetic field. This rotating magnetic field may be represented by a vector of constant magnitude ϕ_m and describes a circle in each revolution as shown in figure (3).



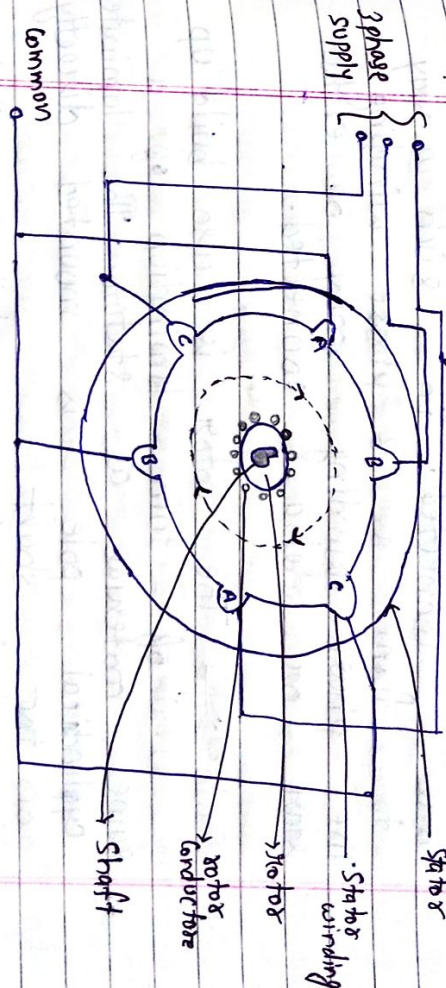
single phase induction motor principles

A single phase induction motor consists of a single phase winding mounted on the stator and a cage winding on the rotor. When a single phase supply is connected to the stator winding a pulsating magnetic field is produced. Under these conditions the rotor doesn't rotate due to the inertia.

For a single phase induction motor is inherently not self starting and requires some special starting means. If however the single phase stator winding is

excited and the rotor of the motor is started by an auxiliary winding. When the starting device is removed, the motor continues to rotate in the direction in which it is started.

Construction of a 3 phase induction motor:-



3 phase induction motor is the most popular type of AC motor. An induction motor doesn't require any starting device. So it is called self started. An induction motor is very commonly used for industrial drive since it is cheap, efficient, reliable and robust. It has good speed regulation, high starting torque and requires little maintenance. A 3 phase induction motor essentially consists of 2 parts. The stator & the rotor.

1) Stator :- The stator is the stationary part build up of high grade alloy steel laminations. The laminations are slotted on the inner periphery and are insulated from each other. The insulated stator conductors are placed in those slots and are connected in such a way to form a 3 phase winding. The phase winding may be either star or delta connected.

2) Rotor :- The rotor is also build up of thin lamination of a same material as stator. The laminated cylindrical pole is mounted directly on the shaft. These laminations are slotted on their outer periphery to receive the rotor conductors. There are two type of induction motor rotors.

- 1) Squirrel cage rotor
- 2) Phase wound or slip ring rotor

1) Cage rotor :- It consists of a cylindrical laminated core with slots parallel to the shaft axis. Each slot contain bars of aluminium or copper at each end of

rotor. The rotor bar conductors are short circuited by heavy end rings to form a cage of the rotor.

2) Slip ring rotor :- The wound rotor consists of a slotted armature insulated conductors. are placed in the slots and connected to form a 3-phase double layer distributed winding similar to the stator winding. The rotor winding are connected in star. And brought outside of the rotor & connected to the 3 insulated slip rings. which are mounted on the shaft with brushes resting on them.

Synchronous speed $\Rightarrow$ It is the speed of field in the stator winding of the magnetic motor at which emf is produced by the At machine. And it is given by

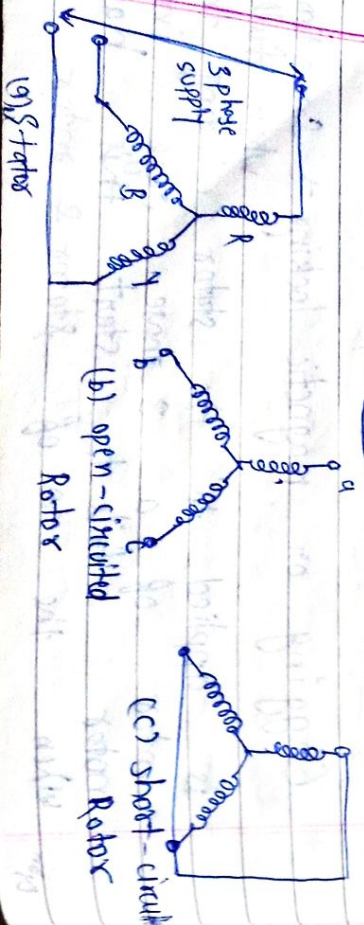
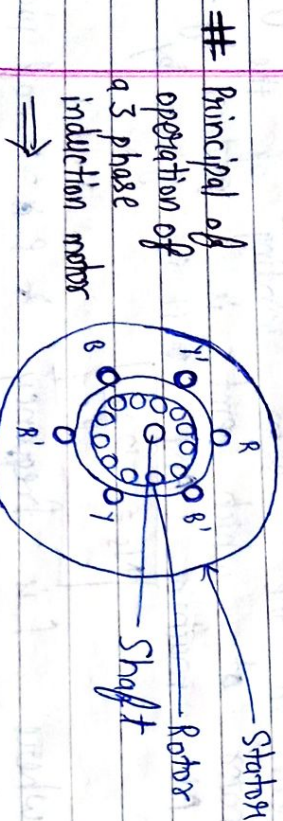
$$N_s = \frac{120f}{P}$$

where f is frequency & $P = \text{No. of pole}$

Cogging or magnetic locking $\Rightarrow$ Some time even is applied to the stator winding the rotor of a 3 phase cage induction motor fails to start this happens when the no. of stator & rotor slots are

either equal or have an integral ratio. With the no. of stator slots equal or an integral multiple of rotor slots, strong aligning forces are produced b/w stator or rotor at the instant of starting. These forces make an element of torque greater than the accelerating torque with the failure of the motor to start. This phenomenon of magnetic ~~locking~~ cogging is called cogging.

Crawling $\Rightarrow$ Induction motor has the tendency to run at a very low speed compared to its synchronous speed that is up to $\frac{1}{7}^{th}$ of the synchronous speed, this phenomena is called crawling.



Let us consider a 3 phase induction motor as shown about with stator & rotor wound for 3 phase and having an identical no. of poles. Initially assume that the stator winding is open circuit and the stator is connected to a 3 phase supply. As we know that when the stator winding carries the 3 phase current a rotating magnetic field is produced in the air gap b/w the stator and the rotor which runs at the synchronous speed.

This rotating magnetic field induces an emf in the stator winding which balances the supply voltage and at the same time this emf (back emf) also induces the emf in the rotor winding. But as the rotor is open circuited so no rotor current flows and the rotor remains stationary. Now if the rotor winding is short circuited so the rotor winding will carry a 3 phase current due to the induced emf in rotor.

Now the rotor behaves like a current carrying conductor placed in the rotating magnetic field of the stator and experiences a torque. Hence the motor starts rotating.

Principle of operation of a 3 phase synchronous generator $\Rightarrow$

The working principle of these machines are governed by two laws

Laws:-

1) Faraday's law of induction

2) Law of interaction
1) Faraday's law of EMI $\Rightarrow$ According to this law is used as a generator when the machine is induced in the armature and emf is induced in the armature conductors when ever cuts its the magnetic field.

2) Law of interaction $\Rightarrow$ This law relates to the production of force or torque that is when ever a current carrying conductor is placed in the magnetic field produced by the current carrying conductor and the main field force is exerted on the conductor producing torque. So the mechanical power spent in maintaining the relative motion of the conductor is converted into electrical energy.

EMF equation of AC Machines:-

$$N_s = \frac{120f}{P}$$

$$E = \frac{P \phi N}{60} \times Z_{ph}$$

$$I_{ph} = \frac{Z_{ph}}{2}$$

$$E = \frac{P \phi N}{60} \times 2 I_{ph} \times \frac{1}{2}$$

$$E = \frac{4 P \phi N I_{ph}}{120} = 4 I_{ph} \phi \times \frac{NP}{120}$$

$$E = 4 \phi I_{ph} f$$

$$(E_{ph})_{rms} = \text{Avg} \times \text{form factor}$$

$$= 4 \phi I_{ph} f \times 1.11$$

$$E_{ph} = 4.44 \phi I_{ph} f$$

Q. A 3 phase star connected 2 pole alternator run at 3600 rpm and flux per pole is 0.1 weber determine the frequency & magnitude of the generated emf.

Ans:-

$$N_s = \frac{120 \times f}{P}$$

$$\frac{3600 \times 2}{60} = f \quad [f = 60 \text{ Hz}]$$

$$E_{ph} = 4 \times 0.1 \times 500 \times 60 \times 1.11$$

$$E_{ph} = 120 \times 100 = 12000 \times 1.11 = \frac{13200}{2} = 6600$$

$$\frac{V_L}{\sqrt{3}} = V_p$$

$$E_L = \sqrt{3} \times 6600 = 1.73 \times 6600 = 11418$$

* Armature reaction $\Rightarrow$ In DC machine the main field is produced by field windings. In both generating and motoring mode the armature carries the current and a magnetic field is produced which is called the armature flux. So the effect of armature flux on the distribution of the main field flux is called the 'armature reaction'. It demagnetizes and cross magnetizes the main field, decrease the value of flux per pole and increase the iron losses in the machine.

Torque Slip characteristics :-

we have
$$T = \frac{K S R_2 E_{20}^2}{R_2^2 + (S X_{20})^2}$$

where $K = \text{constant}$, $s = \text{Slip speed}$

$R_2 = \text{Resistance / Phase of rotor}$

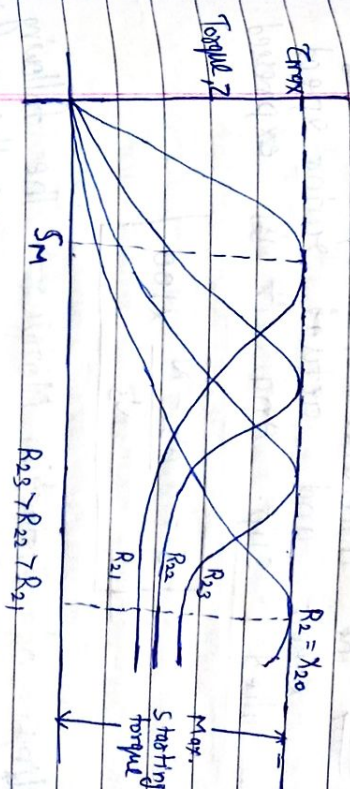
$E_{20} = \text{Emf induced / Phase of rotor at standstill}$

$X_{20} = \text{Reactance / Phase of rotor at standstill}$

if R_2 and E_{20} are constant then the torque depend upon the slip. So the torque slip characteristic curve can be divided into 3 regions :-

1) Low slip region :-

At synchronous speed $s=0$ therefore the torque is also zero. When the speed is very near to synchronous speed the slip is very low and $(s X_{20})^2$ is negligible in comparison with R_2 . Therefore if R_2 is constant the torque is directly proportional to the slip. Hence the torque slip curve is a straight line in this region.



2) Medium slip region :- As the slip increase the term $(s X_{20})^2$ becomes large

so that $(R_2)^2$ may be neglected therefore torque = $\frac{(s X_{20})^2}{K_3 R_2}$

Thus the torque is inversely proportional to the slip. The torque

slip characteristic is represented by a rectangular hyperbola in this region. In doing so it passes through the point of maximum torque where $R_2 = X_{20}$. The maximum torque developed in an induction motor is called pull out & breakdown torque.

3.]

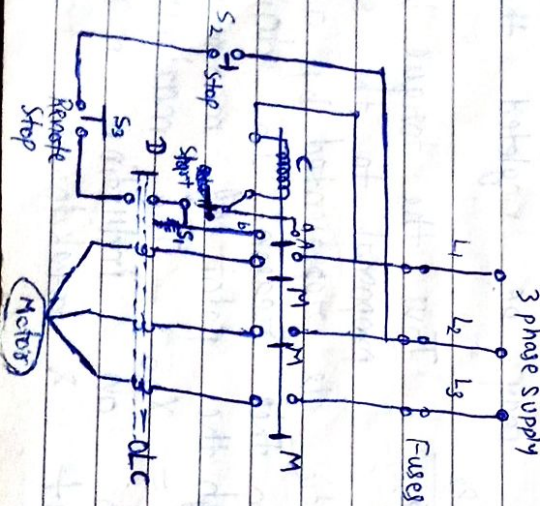
High Slip reason:- The torque decreases beyond the point of max. torque as a result the motor slows down and eventually stop in this reason. At this stage the over load protection must immediately disconnect the motor from the supply to prevent damage due to over heating.

Slip $\Rightarrow$ The difference b/w synchronous speed and actual rotor speed is called slip. And it is expressed as

$$S = \frac{N_s - N}{N_s} \times 100$$

Starting of Induction Motor $\Rightarrow$ The following are the common used starters for induction motor :-

(i) Direct - on - line starter :-

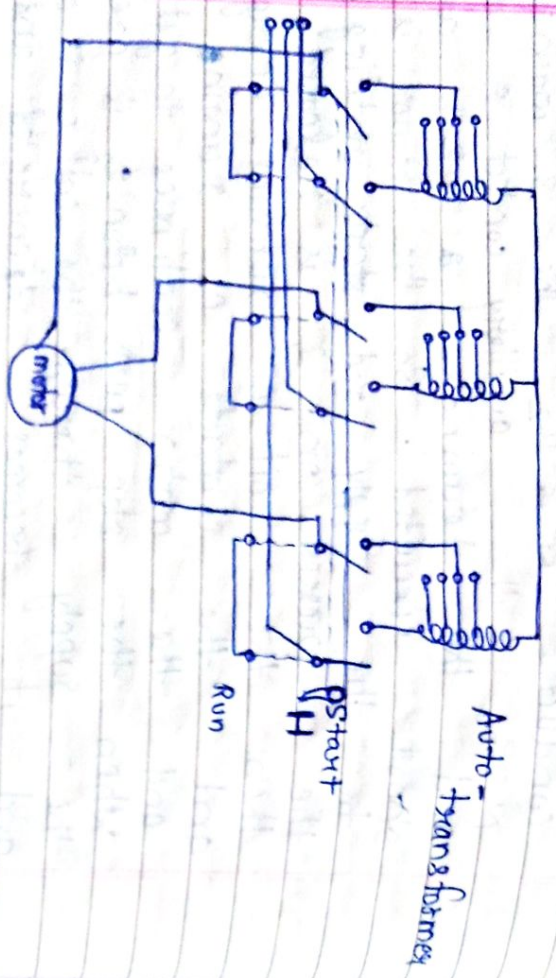


In this method starting the motor is controlled by means of a direct-on-line starter. It consists of a coil operated contactor C controlled by start and stop push buttons. On pressing the start button S₁, the coil C is energised from the 2 line conductors L₁ & L₂. The 3 main contacts M & the auxiliary contact A closed and the terminals a & b are short circuited. So the motor is connected to the supply and starts rotating. When the pressure on S₁ is released, even then the coil C remains energised and the contacts M remain closed. When the stop push button S₂ is pressed the supply through the coil C is disconnected and it becomes deenergised and the main contact M and auxiliary contact A become open. So the supply to the motor is disconnected and the motor stops.

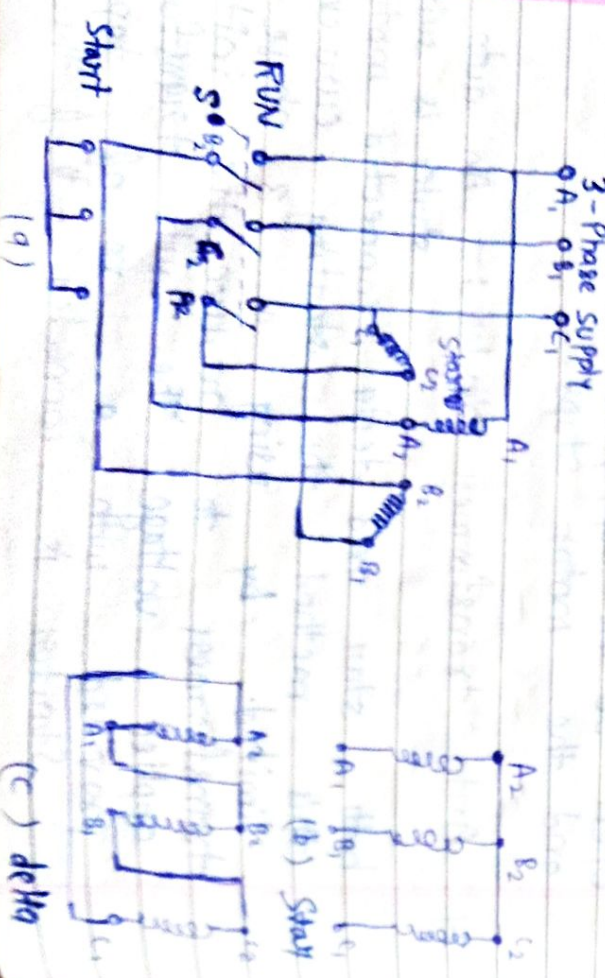
2.] Auto transformer starter :- An auto transformer

starter is suitable for both star and delta connected motors. In this method the starting current is limited by using a 3 phase auto transformer to reduce the initial starting applied voltage. The auto transformer is provided with a no. of tapping. The starter is connected on 1 of the tapping.

to obtain the most suitable starting voltage. A double throw switch is used to connect to auto transformer with the in the circuit for with the help of the handle H as shown in the figure below



3.1 Star-delta Starter, —



A Star-delta starter is used for a cage motor to run normally on delta connected stator winding. The figure 'a' shows the connection of a 3-phase induction motor with a star-delta starter. When the switch 'S' is in the start position. The stator windings are connected in star as shown in figure 'b'. When the motor picks up the speed the change over switch 'S' automatically connects the stator winding in delta as shown in figure 'c'. By connecting the stator winding first in star and then in delta the line current drawn by the motor at starting is reduced by one third ($1/3$)rd. So in this method the starting torque is reduced to ($1/3$)rd that obtained by direct delta starting.

Speed control of induction motor $\Rightarrow$ The

speed of an induction motor is given by

$$N_s = (1-s)N_s$$

$$N_s = \frac{120f}{P}$$

$$N_{s1} = \frac{120f}{P} (1-s)$$

from the above eqⁿ it is seen that the motor speed can be changed by a change in frequency f , no. of pole P and

slip 's'. The main method used for speed control of induction motor are as follows :-

- (i) By changing the no. of stator pole.
- (ii) Stator voltage control
- (iii) Supply frequency control
- (iv) Rotor resistance control
- (v) Slip energy recovery

(1) By Pole changing method :- The method of pole changing are suitable for cage motors only because the cage rotor automatically develops no. of poles equal to the poles of the stator winding. So by changing the no. of stator poles we can control the speed of the motor. The no. of stator poles can be changed by 3 method

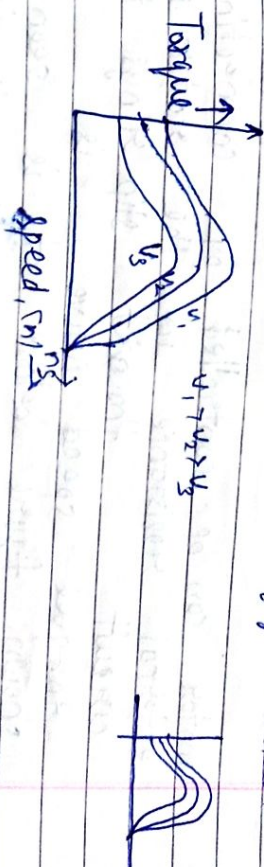
- (i) Multiple stator winding
- (ii) Method of consiguant pole
- (iii) ~~Reverse~~ Pole amplitude modulation

(ii) Stator voltage control :- The speed of the 3-phase induction motor can be varied by varying the supply voltage we know that

$$T = \frac{K_s E_{20}^2}{R_2^2 + (sX_{20})^2}$$

From this relation torque developed is directly proportional to the square of the supply voltage. So the speed control obtained by varying the supply voltage

until the torque required by the load is developed at the desired speed. There fore as the voltage is reduced to reduced the speed the torque developed by the motor also reduces. The torque speed characteristic of a 3 phase induction motor for varying supply voltage is shown in the figure below :-



(iii) Variable frequency control $\Rightarrow$ The synchronous induction motor is given by $N_s = \frac{120f}{p}$ Speed of an

therefore the synchronous speed and the speed of the motor can be controlled by the varying the supply frequency. The emf induced in the stator of the induction motor is given by $E = 4.44 T \phi_m f$ therefore if the supply frequency is changed

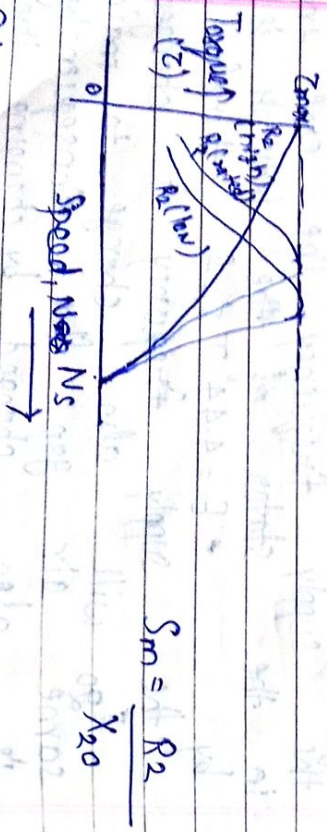
E_{20} will also change to maintain the same air gap flux. Consequently the torque is also changed by changing the supply voltage. Thus the speed control of an induction motor can be done by using variable frequency supply. which requires a variable voltage source. The variable frequency

supply is generally obtained by the following

- (i) Voltage source inverter
- (ii) Current source inverter
- (iii) Cyclo converter

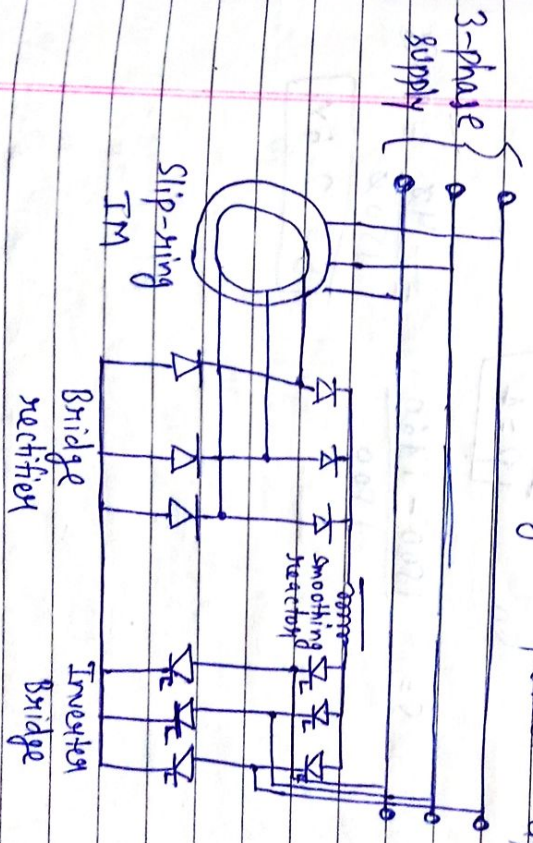
④ Rotor resistance control $\Rightarrow$ The speed of wound induction

motor can be controlled by connecting external resistance in the rotor circuit through slip rings the torque speed curve is shown figure below it is seen that when the value of R_2 is increased as a result the slip will also increase and consequently the speed pull out ~~set~~ speed decreases but maximum torque remain constant. Therefore by this method control is provided to rated speed to lower speed.



⑤ Slip energy recovery $\Rightarrow$ In the rotor the slip power is wasted as $I^2 R$ loss during the low speed operation.

So the efficiency of the drive system by this method of speed control is reduced. So the slip power from the rotor circuit can be recovered and fed back to the AC source thus the over all efficiency can be increased. This method for recovering the slip power is known as static Scherbius drive. In this method a portion of the rotor AC power is converted into DC. By a diode rectifier. The rectified current is smoothed by the smoothing reactor then the output of the rectifier is fed back to the AC source. This method which inverts into AC power as fed back to the AC source. This method is used large power application.



Q. An induction motor runs at 1460 rpm under fullload condition when supplied by 50Hz supply find the no. of pole & slip of the motor?

Ans-

$$N = 1460 \text{ rpm}$$

$$f = 50 \text{ Hz}$$

$$N_s = \frac{120f}{P}$$

$$P = \frac{120 \times 50}{1460} = \frac{600}{146} \approx \frac{300}{73}$$

$$N_s = \frac{120f}{P} \quad \text{if } P=2$$

$$N_s = \frac{6 \times 120 \times 50}{2} = 3000$$

$$\text{if } P=4 \quad N_s = \frac{120 \times 50}{4} = 1500$$

$$S_0 = \boxed{P=4}$$

$$S = \frac{1500 - 1460}{1500} = \frac{40}{1500} = 0.026$$

$$\boxed{S = 2.6\%}$$

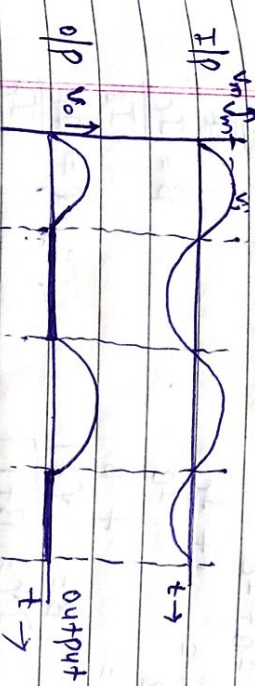
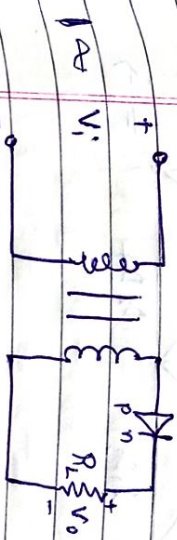
conductivity of semiconductor :-

$$\sigma = \frac{J}{E}$$

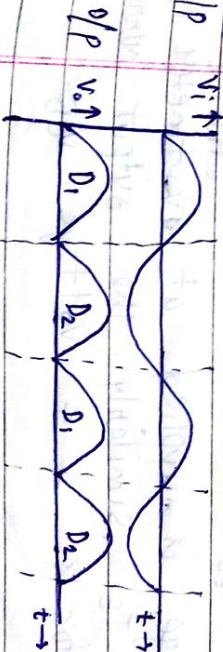
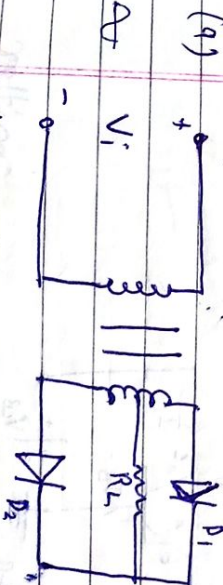
$$\boxed{J = \sigma E}$$

Semiconductor

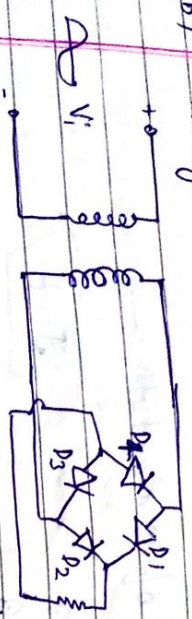
(i) Single phase half wave rectifier with R-load :-



(ii) Single phase full wave rectifier with R-load :-



(b) Bridge rectifier in full wave rectifier



$$I_E = I_B + I_C$$

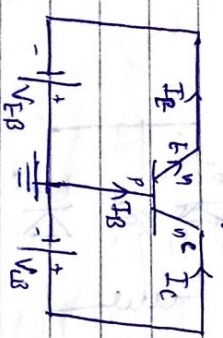
$$\frac{I_E}{I_B} = 1 + \frac{I_C}{I_B}$$

$$V = 1 + \beta$$

$$\beta = \frac{V-1}{V}$$

$$\alpha = \frac{I_C}{I_E}$$

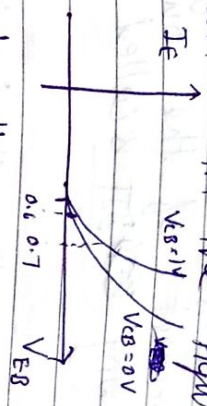
(i) Transistor configuration: —
CB



The input voltage is applied at a wattmeter V_{EB} & current I_E . Similarly the output current at collector is I_C & voltage V_{CB} .

input characteristic: —

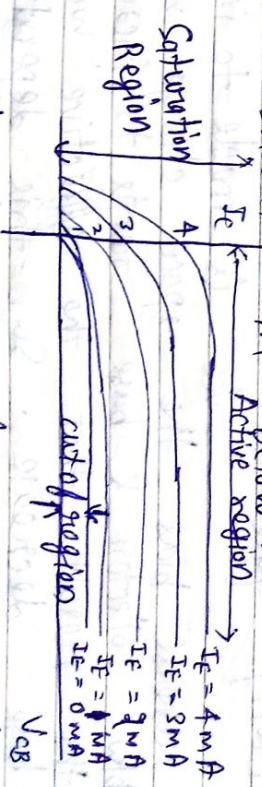
A curve is drawn b/w current I_E & V_{EB} at const. output voltage that is V_{CB} . And the input characteristic shown in the figure below: —



Due to the increase in output voltage V_{CB} the I_E increases for a small change in V_{EB} because increased reverse bias on BC junction increases the depletion region width therefore reduces the base width for the recombination and increases the concentration of majority carriers. So by changing the output voltage V_{CB} increase the I_E .

Output characteristic: — The curve of output current

V_{CB} are plotted for the constant value of input current I_E . And the output char. is shown in below: —



The three region of operation of a transistor can be seen on the output characteristic curve.

(i) Active region:— In this region the transistor work as a amplifier. When the input junction is forward bias & output junction is reverse bias. The output current I_c is almost constant for a given input current independent of the output voltage V_{ce} . That is why ~~BJT~~ BJT is called a current control device.

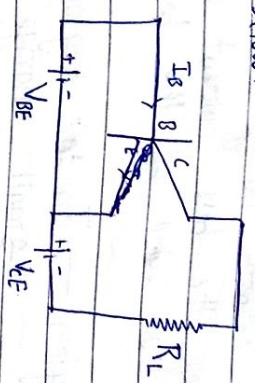
(ii) Cut off region:— When the input junction is in reverse bias and the device is set to be off. This region lie below $I_E = 0$. However a small character current flow due to minority carrier.

(iii) Saturation region:— The output junction is forward biased and the transistor is said to be saturated and the region lie at the left of $V_{ce} = 0$. In this region a small change in V_{ce} will rapidly increase the I_c .

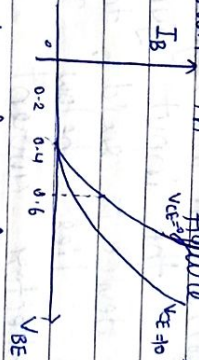
Note— Early effect $\rightarrow$ As the collector voltage V_{ce} is made to increase the reverse bias the space charge width b/w collector & base tends to increase with the result that the effective width of the base decrease. So this dependence of base width on collector to emitter voltage is known as the early effect.

(ii) Common emitter configuration:— Transistor in common emitter ~~mode~~ is shown in

Figure below.



(a) Input characteristics:— To determine the input char. the collector to emitter voltage V_{ce} is kept constant and the base current is increased from zero in equal steps by increasing V_{be} as shown in figure below.



In input char. for a fixed value of V_{ce} base current reduces for an increase in voltage V_{ce} due to the early effect.

(b) Output characteristics:— The curve of the output current I_c vs output voltage V_{ce} are plotted for different constant values of input current I_B . The output char. curve drawn for the 3 region of a operation of a transistor as shown in figure below I_c .



(i) Active region:— In this region transistor works as an amplifier. In this region the effect of the collector voltage is very small for low value of the base current but the effect increases with the value of the base current.

(ii) Saturation Region:— In this region of operation a small change in V_{CE} forward bias the output junction and the transistor is said to be saturated. In this region a small change in V_{CE} will rapidly increase the I_C .

(iii) Cut off region:— In this region of operation the input junction is reverse bias so the device is said to be off. This region lies below $I_B = 0$. However a small collector current flows due to minority carriers.

(iii) Stray loss:— Leakage flux in a transformer produces eddy current in the conductor, transformer oil etc. And these eddy currents are responsible for straight losses.

(iv) ~~Die~~ Dielectric loss:— It occurs in the insulating material such as in the transformer oil and in solid insulation of transformer. The stray loss & dielectric loss are usually very small so there existence will be negligible in an transformer.

$$15 - 8I_1 - 3I_1 + 3I_2 + 8I_3 = 0$$

$$15 = 11I_1 - 3I_2 - 8I_3 \quad \text{--- (1)}$$

$$5I_3 + 3I_1 = 10I_2 \quad \text{--- (2)}$$

$$5I_2 + 8I_1 = 23I_3 \quad \text{--- (3)}$$

Unit # 4

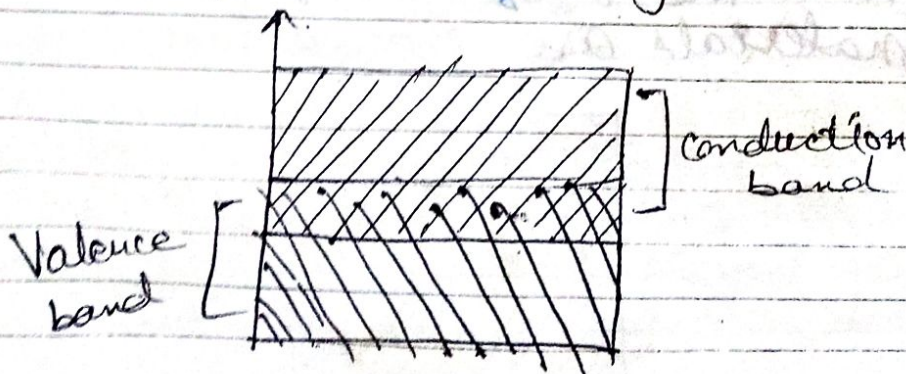
BASIC ELECTRONICS

★ Introduction:

Conductivity:

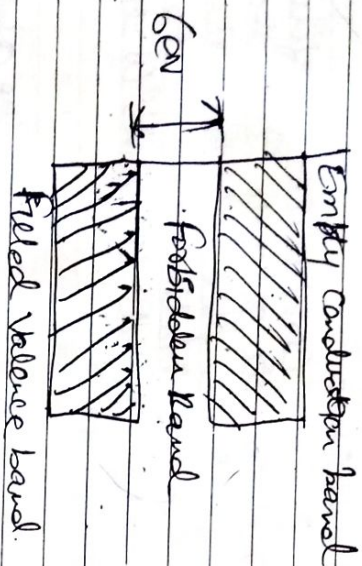
Conductivity refers to the ability of a material to generate free electrons to conduct current. These free electrons are generated by providing excitation energy to the electrons in valence band to move to the conduction band.

Conductors (Metals): In the absence of the energy gap, a large number of electrons are available for conduction at room temp. So without conduction even at room temp. such materials conduct easily and hence called as conductor. eg copper, iron etc.



(b) Insulator :

The materials which have large forbidden gap between the energy bands approx. 6 eV . So it becomes impossible for an electron to jump from valence to conduction band. So such materials do not conduct called an insulator. eg Diamond.



(c) Semi-conductor :

The material for which the forbidden energy gap is relatively very small is called a semi-conductor. approx 1 eV . The most important semi-conductor materials are Germanium and Silicon.



★ Semiconductors :

Semiconductors have resistivity in between that of conductors and insulators. They have negative temp. coefficient of resistance and when suitable metallic impurities are added, the conductivity of semiconductor change.

Based on the absence and presence of these extended impurities, semiconductors are classified as intrinsic and extrinsic semiconductors.

Intrinsic (pure) semiconductors :

A piece of semiconductor in its purest form is known as Intrinsic semiconductor.

At 0°temp , no free electron is available as the electrons in the valence shell makes bond and share the valence electrons of other neighbouring atoms to attain 8 electron position in valence shell and get completely filled and stable. So they behave as perfect insulator.

Now, at room temperature a no. of electrons in valence band absorb the thermal energy from surrounding, due to this they break covalent bond and shift to conduction band.

So each electron shifted to conduction band, one hole is created in valence band. Thus, free electrons and holes are generated in the intrinsic semiconductor.

* Extrinsic Semiconductor:

When some impurity generally referred to as dopant is added to intrinsic semiconductor, its conductivity can be increased. These types of semiconductor are called extrinsic semiconductor.

The process of adding impurity is called doping. The dopant added is generally of two types, either trivalent or pentavalent atoms.

When an intrinsic semiconductor is doped with pentavalent atom (i.e. atom having five electrons in its valence shell, it is called donor doping. This type of doping creates more number of free electrons and results into N-type of semiconductor.

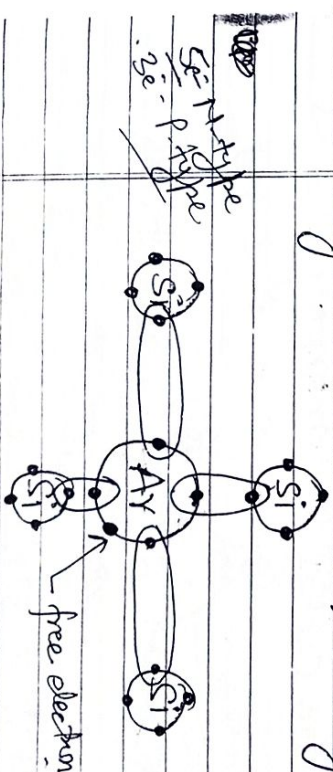
When an intrinsic semiconductor is doped with trivalent atom (i.e. atom consisting of three electrons in valence shell, it is called acceptor doping). This type of doping creates more number of holes and results

extrinsic semiconductor is called as P-type semiconductor.

* N-type semiconductor:

To form N-type semiconductor, pentavalent impurity such as Antimony, phosphorus and arsenic is added to intrinsic semiconductor. Each atom of dopant donates one free electron. After donating one electron, each dopant atom acquires a positive charge.

Current conduction in N-type is due to free electrons so they are called majority carriers and the holes are small in number they are called as minority carriers.



So, as the donor impurity is added, energy level also introduced just below the conduction band. The donor energy level is 0.05 eV distance below the conduction band.

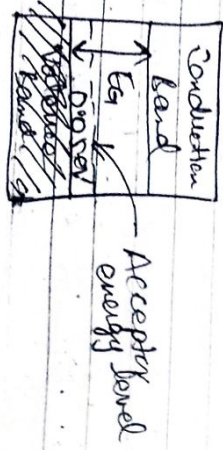
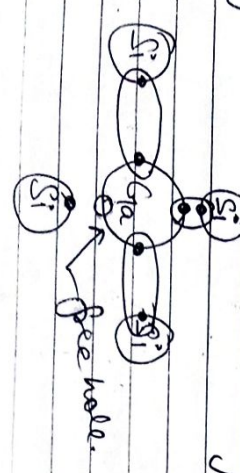
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(b). P-type semiconductor:

When a small amount of trivalent impurity such as Boron, Gallium or Indium is added to an intrinsic or pure, semiconductor it is called a p-type semiconductor.

In this, the three valence electron of the dopant makes covalent bond with the three adjacent silicon atom. Due to shortage of one electron, the fourth covalent bond remains incomplete. This results in vacant space, called as hole.

~~Now~~ In p-type semiconductor, the carrier concentration is due to by holes, hence holes are majority carriers and the free electrons are small in number. They are called as minority carriers.



4

* Conductivity of semiconductor:

In a pure semiconductor, the no. of holes are equal to the no. of free electrons. With each electron-hole pair created, two charge-carrying particles are formed. One is negative, which is a free electron with mobility μ_n and other is positive i.e. the hole with mobility μ_p .

The mechanism by which hole contributes to conductivity is that when hole exists, the valence electron can jump into it and thus hole leaves its original position. Hole moves in the opposite direction to the electron.

(a) ○ ○ ○ ○ ○ ○ ○ ○ ○ ○

(b) ⊙ ⊙ ⊙ ⊙ ⊙ ⊙ ⊙ ⊙

q , the total current density J is

$$J = J_n + J_p$$

$$J = q \cdot n \cdot \mu_n E + q \cdot p \cdot \mu_p E$$

$$J = q [\mu_n n + \mu_p p] E$$

$$J = \sigma E$$

* For pure semiconductor, $n = p = n_i$
where n_i is referred to as intrinsic
Concentration.

$$n_i = n_i (e^{E_n + E_g} + e^{E_p - E_g})$$

$$\text{where } [n_i = q n_i (e^{E_n + E_g} + e^{E_p - E_g})]$$

* For N-type semiconductor, $n \gg p$.

$$n_i = q n_i$$

Unit of conductivity is S/m .

* For P-type semiconductor, $p \gg n$,

$$n_i = q n_i$$

diode is a one-way device which allows the current to flow in one direction.

* Conduction in Semiconductor diode:

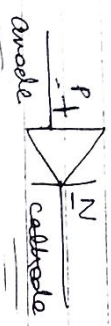
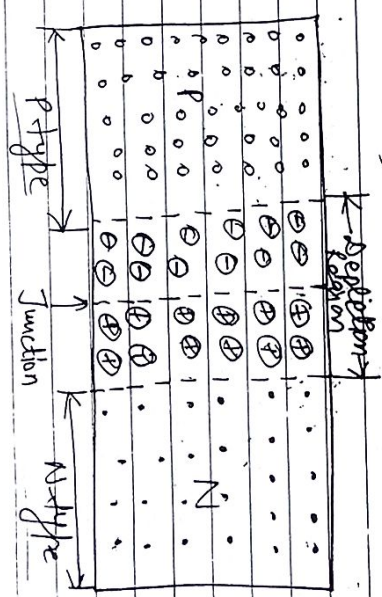
If the donor impurities are introduced into one side and acceptor into the other side of a single crystal of a semiconductor, a p-n junction is formed. V-type material and P-type materials have higher concentrations of electrons and holes resp. So at the junction there is a tendency for the free electrons on N-type material to diffuse.

Depletion region :- It is an insulating region with in a conductive material. (5)
Although in a conductive material, where free mobile charge carriers have been diffused away, leaving behind the impurities.

In the P-side and holes from P-type to the N-side. This process is called diffusion.

When the electrons diffuse to P-type region and holes to N-region, there is a region of charged ions formed at the junction. In the region of the junction is the depleted mobile charges. It is called depletion region.

And as a result, a potential difference is established between the N-type and P-type regions which is called potential barrier. (V).



$$V_p = V_T \ln \frac{N_D N_A}{n_i^2}, \text{ where } V_T = \frac{T}{11,600}$$

* Biasing of a PN Junction:

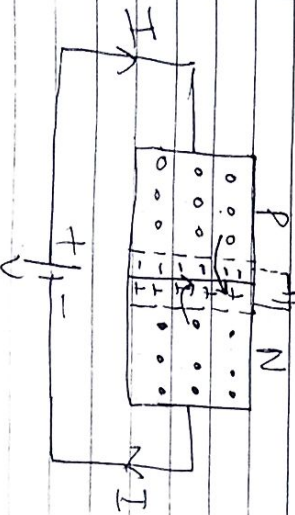
At equilibrium, under open circuit, the current through the junction is zero. In order to break circuit from through the junction, we need to apply some potential to the PN diode.

The application of DC voltage across the diode is called biasing.

* There are two type of biasing:

(i) Forward biasing:

A diode is said to be forward biased, when an external voltage source is connected with its positive towards P-side of the diode and negative towards the N-side of the diode.



When the applied voltage is greater than the potential barrier (V_0), then

the holes in the P-side are repelled by the positive terminal of the battery and forced to cross the junction.

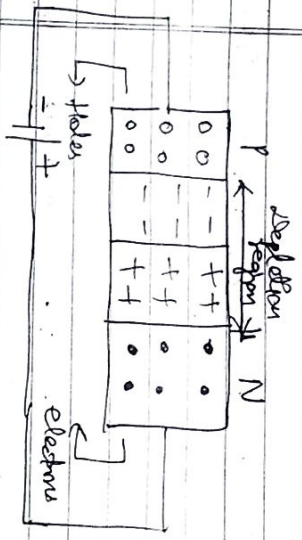
Similarly, the electrons in the N-side are also repelled by the negative of the battery and forced to cross the junction. So this results in the reduction of the potential barrier and allows large forward current to flow.

(ii) Reverse bias:

When the negative terminal of the battery is connected to the P-side and positive terminal of the battery is connected to the N-side of the PN junction, the bias applied is known as reverse bias.

Due to this, the holes of the P-side moves or attracted towards the negative terminal of the battery and electrons of the N-side are to the positive terminal of the battery so as a result the width of the depletion region increases, the potential barrier is increased which prevents the flow of majority carriers. (7th 11)

②



Variation of diode current is given by,

$$I = I_0 \left(e^{\frac{V}{nV_T}} - 1 \right)$$

where $V_T = \frac{T}{11600}$, T is temp.

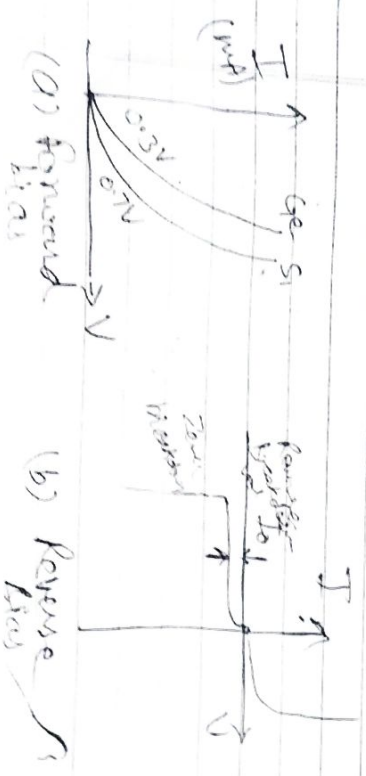
$n = 1$ for Ge, 2 for Si

$I =$ Current through diode

$V =$ Applied voltage

$I_0 =$ Reverse saturation current

V-I characteristics:



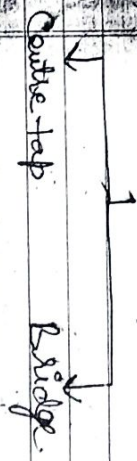
⑦

Rectifiers:

Rectifier is defined as an electronic device used for converting AC voltage into unidirectional voltage or DC voltage. It uses the PN junction diode.

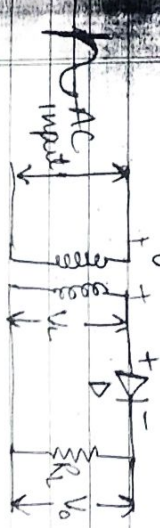
They are classified depending upon the period of conduction:

- i) Half wave rectifier
- ii) Full wave rectifier



Half wave rectifier:

The basic circuit and waveforms of a half wave rectifier is shown below:



It uses a single diode which conducts in one half of the input AC cycle, thus, pulsating DC voltage is obtained for half cycle of input AC. Hence, the circuit is called half wave rectifier.

①

Let the V_i be the voltage to the primary of the transformer and given by the eqⁿ

$$V_i = V_m \sin \omega t$$

Case 1: During the positive half cycle of the input signal, the p-side of diode becomes more positive with respect to the cathode and hence, diode D conducts. So the whole input voltage will appear across the load resistance R_L .

Case 2: During negative half cycle of the input signal, the p-side of the diode becomes negative with respect to the cathode and hence, diode is reversed biased and so does not conduct. So the voltage across R_L is zero.

(i) Avg. value:

$$V_{avg} = \frac{1}{2\pi} \left[\int_0^{\pi} V_m \sin \omega t d(\omega t) + \int_{\pi}^{2\pi} 0 d(\omega t) \right]$$

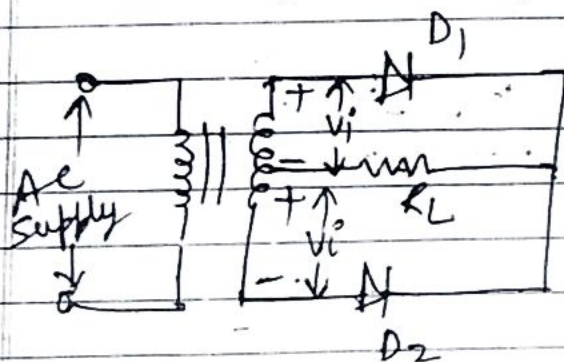
$$V_{avg} = \frac{V_m}{2\pi} [-\cos \omega t]_0^{\pi}$$

$$V_{avg} = \frac{V_m}{\pi} = 0.318 V_m$$

(ii) Full wave rectifier :

(i) Centre Tap rectifier :

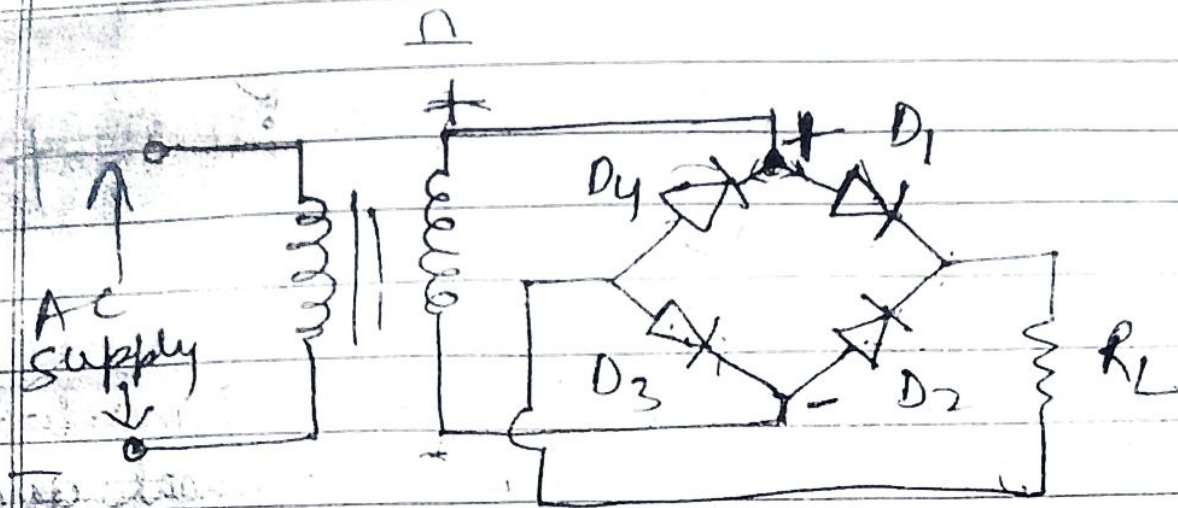
full wave rectifier is an electronic device that converts an A.C voltage into a pulsating D.C voltage using both half cycles of the applied A.C voltage. It uses two diodes of which one conducts during one half cycle while the other diode conducts during the other half cycle of the A.C voltage.



Case 1: During positive half ~~cycle~~ cycle, the p side of D_1 is become positive and at the same time, the anode of Diode D_2 is negative so D_1 conducts and D_2 does not conduct.

Case 2: During negative half cycle of the D_2 becomes forward biased and D_1 becomes reversed biased so D_1 does not conduct and D_2 conducts.

(10)



Case I: For positive half cycle of the input AC voltage, diodes D_1 and D_3 conduct whereas D_2 and D_4 do not conduct and load current flows through R_L .

Case II: For negative half cycle of the input AC voltage, diodes D_2 and D_4 conduct whereas D_1 and D_3 do not conduct and load current flows through R_L .

* Bipolar Junction Transistor (BJT) :

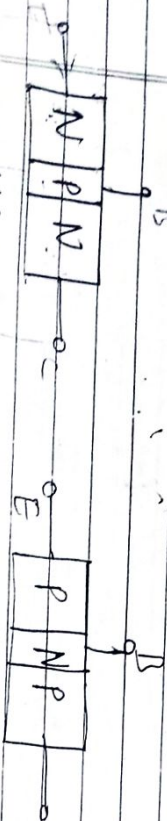
A BJT is a three terminal semiconductor device in which the operation depends on the interaction of both majority and minority carriers and hence the name bipolar. With three terminals, two junctions are present, hence it is a bi-junction device.

Transistor refers to as 'Transfer of resistor' because it transfers the signal from low resistance input junction to high resistance output junction.

Construction:

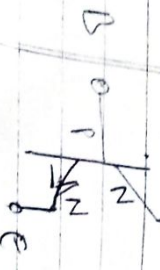
The BJT consists of a silicon crystal, in which a thin layer of N-type silicon is sandwiched between two layers of P-type silicon. This transistor is referred to as NPN.

And when a thin layer of P-type is sandwiched b/w the two layers of N-type silicon. This transistor is referred to as PNP. And the three terminals are emitter, base and collector.

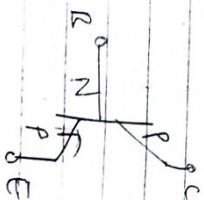


(a) NPN

(b) PNP



(a) NPN Symbol



(b) PNP Symbol

(i) Emitter: The main function of this region is to supply majority carriers to the base and it is most heavily doped.

(ii) Base: This is the middle terminal of the transistor which is thinnest of the three and it is very less doped as it is to have the most of injected carriers.

The main purpose of this terminal is to collect the carriers through base and having large area toward the other two terminals.

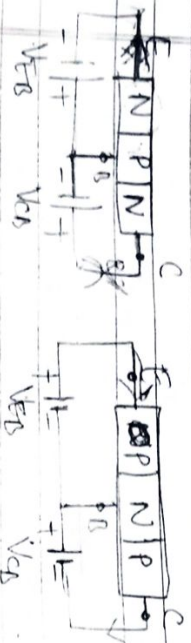
Transistor Biasing:

The application of external voltage is referred to as biasing. After biased, the junction of BJT can also be forward or reverse biased.

So, depending on the nature of biasing, transistors work in either of the three regions:

(1) Active region or (mode):

When emitter-base junction is forward biased and collector-base junction is reverse biased.



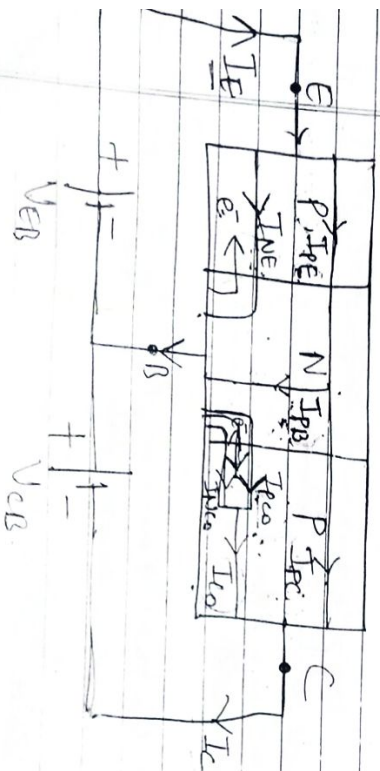
(a) NPN

(b) PNP

(b) Cut-off region: When both the junctions are reversed biased.

(c) Saturation region: - When both the junctions are forward biased.

Transistor current components:



(i) Emitter current:

$$I_E = I_{PE} + I_{NE}$$

As $I_{PE} \gg I_{NE}$

$$I_E \approx I_{PE}$$

Collector current:

$$I_C = I_{PC} + I_{CO}$$

And $I_{CO} = I_{PCO} + I_{NCO}$

I_{PCO} is due to holes
 I_{NCO} is due to e^- .

So, net collector current:

$$I_C = \alpha I_E + I_{CO}$$

Where $\alpha > I_{PC}$ is the large current gain
 I_E

and I_{CO} is the reverse saturation current

(iii) Base current:

$$I_B = I_{NE} + I_{PB} - I_{CO}$$

* Current Amplification factor:

(i) The alpha factor (α): Ratio of I_C to I_E .

$$\alpha = \frac{I_C}{I_E} \quad \begin{array}{l} \text{Collector current} \\ \text{Emitter "} \end{array}$$

(ii) Beta factor (β): Ratio of collector current (I_C) to I_B .

$$\beta = \frac{I_C}{I_B}$$

(iii) Gamma factor (γ): Ratio of I_E to I_B .

$$\gamma = \frac{I_E}{I_B}$$

* Relationship between α and β and γ .

We know, $I_E = I_B + I_C$

Now, incremental change in emitter current

$$\Delta I_E = \Delta I_B + \Delta I_C$$

$$\Delta I_B = \Delta I_E - \Delta I_C$$

$$\beta = \frac{I_C}{I_B} > \frac{\Delta I_C}{\Delta I_E - \Delta I_C}$$

Considering numerator and denominator by ΔI_E (we get),

$$\beta = \frac{\Delta I_C / \Delta I_E}{\Delta I_B / \Delta I_E}$$

$$\boxed{\beta = \frac{\alpha}{1 - \alpha}}$$

$$\text{or } \beta(1 - \alpha) = \alpha$$

$$\beta - \beta\alpha = \alpha$$

$$\beta = \alpha + \beta\alpha$$

$$\beta = \alpha(1 + \beta)$$

$$\boxed{\alpha = \frac{\beta}{1 + \beta}}$$

$$\text{Also, } \gamma = \frac{\Delta I_E}{\Delta I_B}$$

$$\gamma = \frac{\Delta I_E}{\Delta I_E - \Delta I_C}$$

$$\gamma = \frac{1}{1 - \frac{\Delta I_C}{\Delta I_E}} = \boxed{\frac{1}{1 - \alpha}}$$

We know that,

$$\alpha = \frac{\beta}{1 + \beta}$$

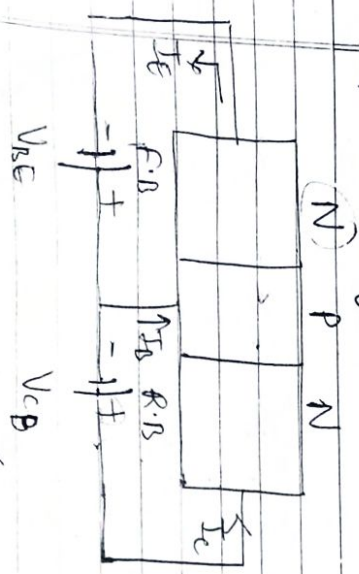
$$\gamma = \frac{1}{1 - \alpha}$$

$$\gamma = \frac{1}{1 - \left(\frac{\beta}{1 + \beta} \right)}$$

$$\gamma = \frac{1 + \beta}{1 + \beta - \beta}$$

$$\gamma = 1 + \beta$$

* Operation of Transistor:



Transistor as a power amplifier circuit used to amplify and switch electronic signals and electronic power.

As shown in the fig; the forward bias applied to the emitter base junction of an NPN transistor causes a lot of electrons from the emitter region to cross over to the base region. And ~~also~~ as the base is lightly doped with P-type impurity, the no. of holes in the base region is very small. And the no. of electrons that combine with the holes in the P-type base region is also very small. Hence a few electrons combine with holes to constitute a base current I_B . The remaining electrons cross over into the collector region and move towards the positive terminal of the supply V_{CC} .

So, the magnitude of the emitter current I_E is given by,

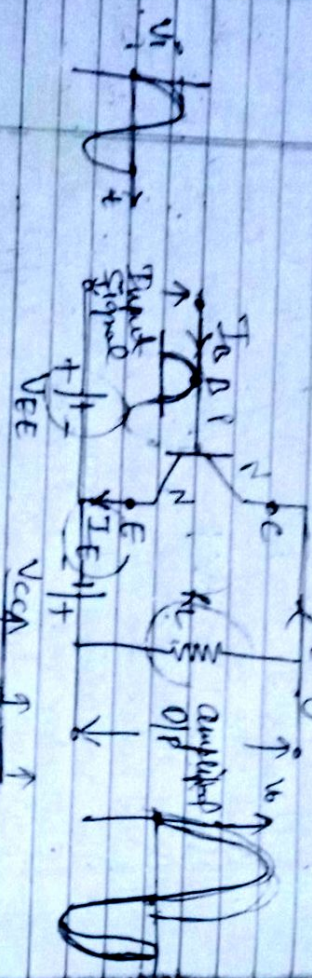
$$I_E = I_C + I_B$$

* Transistor as an Amplifier:

The very first requirement for a transistor to work as an amplifier is that its emitter junction is forward biased and collector junction is reverse biased and transistor operates in its active region.

we can use common emitter circuit to amplify the input signal without a delay of π significantly.

The fig. below shows the basic circuit of a common-emitter transistor amplifier. In this, the E junction is forward biased by the battery V_{BE} and C is reverse biased by the battery V_{CE} .



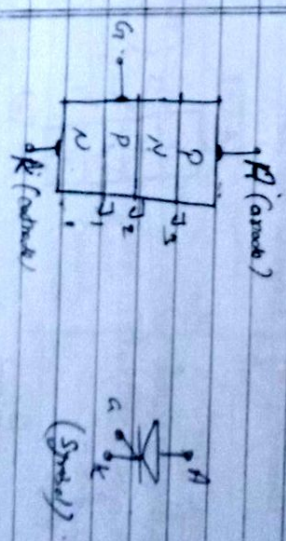
★ Operation :

When a weak signal to be amplified is applied b/w emitter base junction during the half cycle, the forward bias across the junction increases and electron flow increases through base and hence increases I_C . And this increase in I_C causes more voltage drop across the load resistor R_L .

During (-ve) half cycle, the forward biased voltage across the $E-B$ decreases and this decreases the I_C and also the voltage drop across R_L . So an amplified signal appears across R_L .

Thyristor (SCR)

- It is a solid state device like Transistor but their character is quite different.
- It is a 4-terminal (Anode, Cathode, Gate, 4 layer P-N-P-N) and three junction (J_1, J_2, J_3) device.
- Gate terminal is connected near to the cathode terminal.
- SCR is a uni-directional device.
- SCR is a semi-controlled device.
- SCR is a charge controlled device.
- SCR is a uni-directional device, which lets the current flow cathode to anode, but unlike diode it also block current flow from anode to cathode.
- Inner two layer of SCR are lightly doped so that strength of junction is more than strength of junction J_1, J_3 .

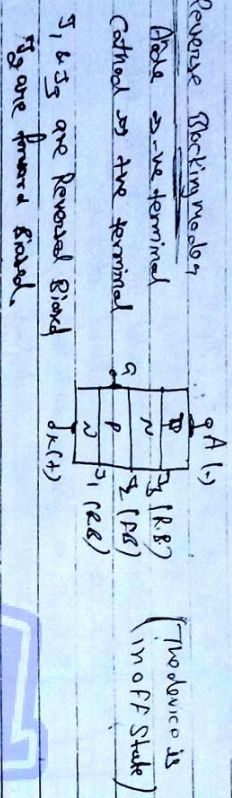


★ Characteristics of SCR

- (1) V-I characteristics
- (2) Gate characteristics

Static V-I Characteristics
It has 3-Mode of operation

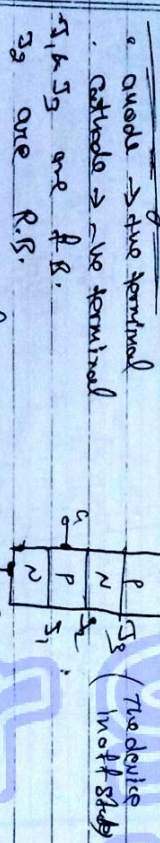
1. Reverse Blocking Mode



In this section only small amount of reverse leakage current is flow (due to minority carrier) in this section.

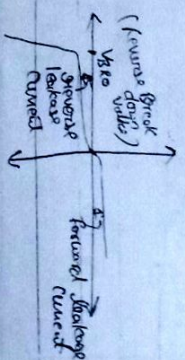
When the reverse voltage become more than avalanche breakdown occurs at reverse bias junction J_1, J_3 and device start conducting at reverse direction.

2. Forward Blocking Mode

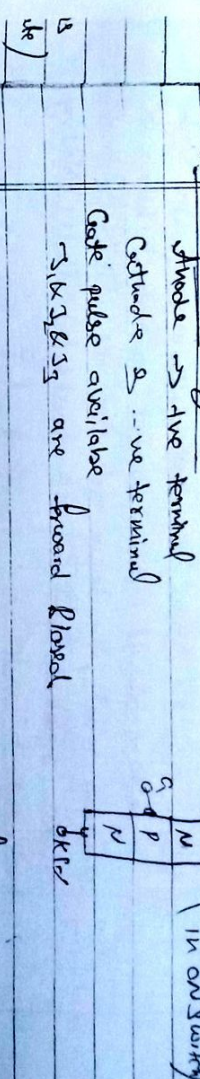


Only small amount of forward current flow in this section.

The device remains in this mode until the anode voltage reaches a critical value called forward Breakover Voltage (V_{BO})

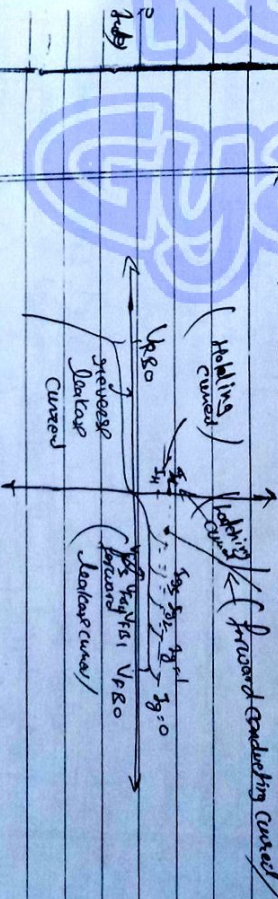


3. Forward Conducting Mode



When anode voltage is greater than the forward Breakover voltage then Avalanche Breakdown occurs at J_2 and device start conducting.
SCR start conducting only if current through the device greater than a minimum current called latching current during turn ON process.

Some important terms in V-I characteristics



4. Latching current - It is the minimum anode current required to turn ON the SCR during turning ON process. When I_A Breakdown then large number of carriers are generated so, large current starts flowing. This current is latching current.

5) Holding Current is when the device is already in Conducting state and if anode current is reduced below a minimum level, the device gets turn off. This minimum current required to turn off device is called Holding current.

2) Forward Breakover Voltage is

It is the critical or minimum voltage required for Breakover of junction T_2 under forward biased condition for turn on device.

• When anode voltage equal to forward breakover voltage the junction T_2 breakdown due to avalanche breakdown.

• When the gate current increases the forward breakover voltage decreases.

3) Turner Break down is

It depends on voltage and doping level, both layer must be heavily doped, small voltage required to breakdown. It is electrostatic breakdown.

3) Avalanche breakdown is

It depends on occurs due to thermal effect or temperature. It occurs in lightly doped semi-conductor junctions.

Single-Phase Inverter :-

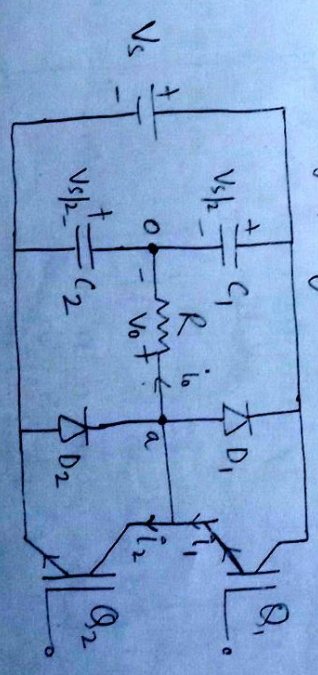
An Inverter is a circuit which converts a DC power into an AC power at desired output voltage and frequency. The AC output voltage could be fixed or variable voltage and frequency. The output voltage waveform of an ideal inverter should be sinusoidal.

There are two types of single-phase inverter :-

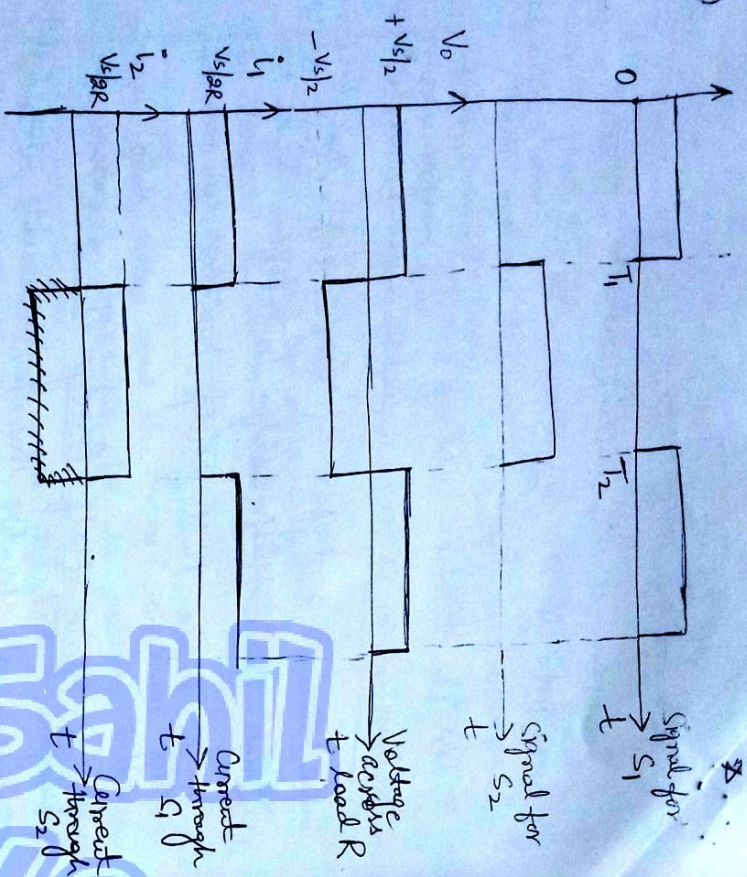
- 1) Single-Phase Half-Bridge Inverter.
- 2) Single-Phase Full-Bridge Inverter.

1) Single-Phase Half-Bridge Inverter :-

For, the half-bridge inverter circuit, the centre-tap of the DC supply is used as one of the load terminals. The centre-tap is created by the two series-connected equal-valued capacitors across the DC supply. The two equal-valued capacitors across the DC supply. The two equal-valued capacitors, C_1 and C_2 are switched alternately, at the desired output frequency, for.



(15)



Working :-

- 1) The current entering node a is considered to be positive.
- 2) The switches S_1 and S_2 are unidirectional i.e. they conduct current in one direction.
- 3) The switching sequence is so design that switch S_1 is on for the time duration $0 \leq t \leq T_1$ and the switch S_2 is on for the time duration $T_1 \leq t \leq T_2$.
- 4) When switch S_1 is turned on, the instantaneous voltage across the load is $V_o = +V_s/2$.
- 5) When the switch S_2 is only turned on, the voltage across load is $V_o = -V_s/2$.

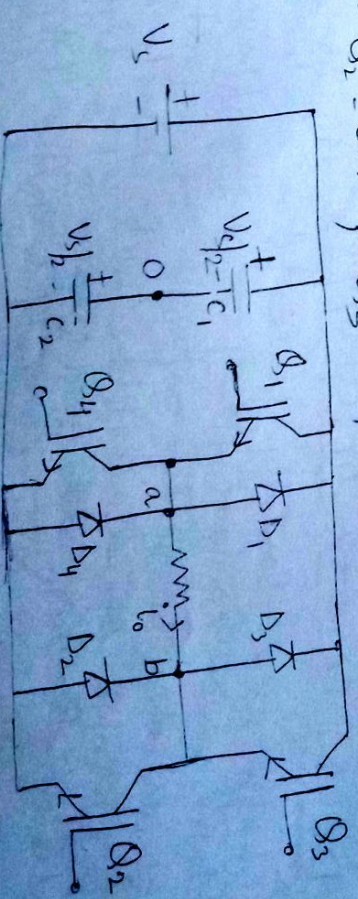
Single Phase Full Bridge Inverter :-

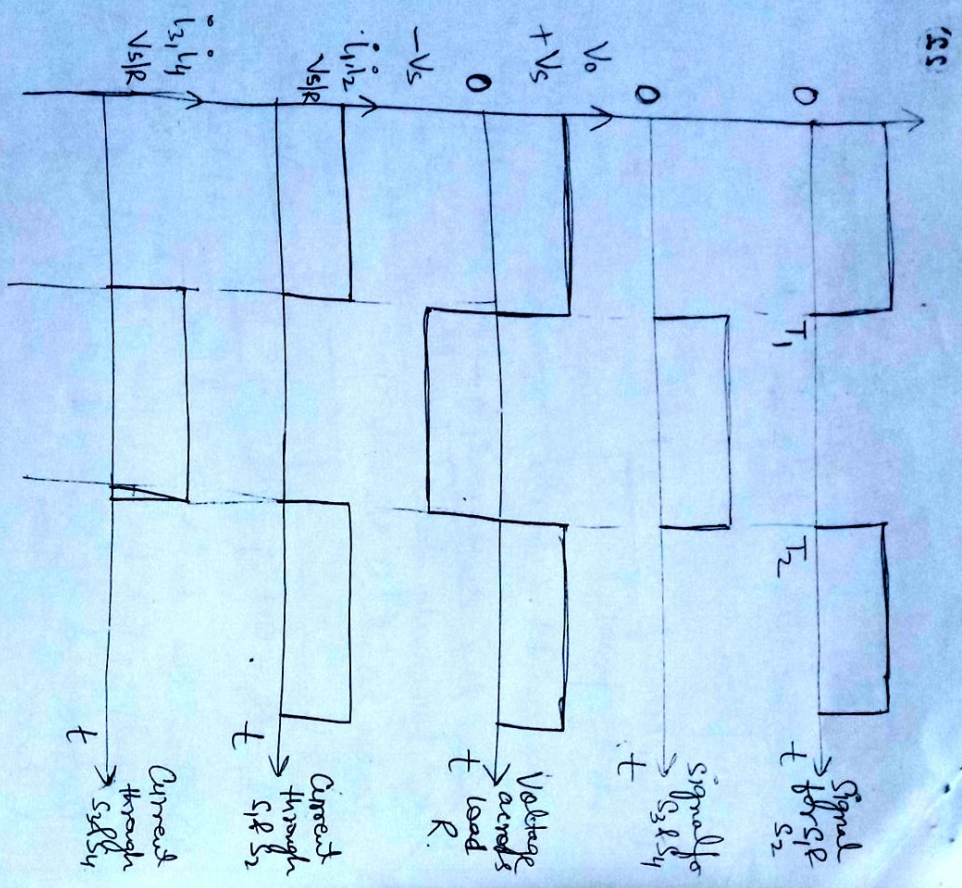
(22)

A single phase bridge DC-AC inverter is shown in figure below. The analysis of the single phase DC-AC inverter is done by taking the following assumptions into account as :-

- 1) The current entering node a in figure is considered to be positive.
- 2) The switches S_1, S_2, S_3 & S_4 are unidirectional.
- 3) When the switches S_1 and S_2 are turned on simultaneously for a duration $0 \leq t \leq T_1$, the voltage across load is $V_o = V_s$ and the current flows from a to b.
- 4) When the switches S_3 & S_4 are turned on for a duration $T_1 \leq t \leq T_2$, the voltage across the load is reversed i.e. $V_o = -V_s$ and current flows from b to a.

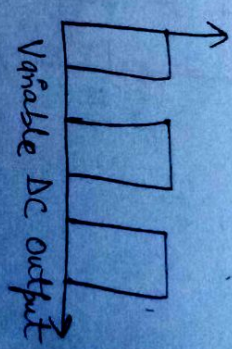
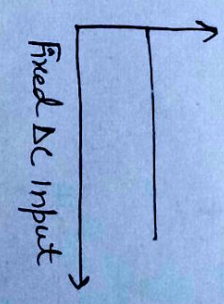
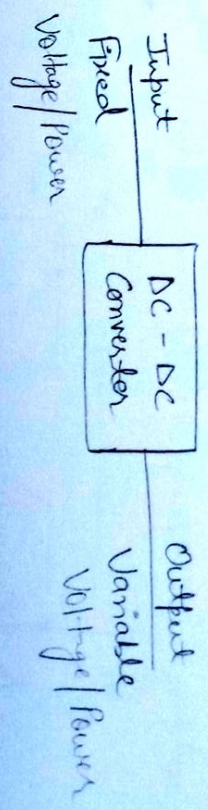
$\phi_1 - \phi_2 = \text{OFF} ; \phi_3 - \phi_4 = \text{ON} \Rightarrow V_o = -V_s$





DC-DC Converter or Chopper :-

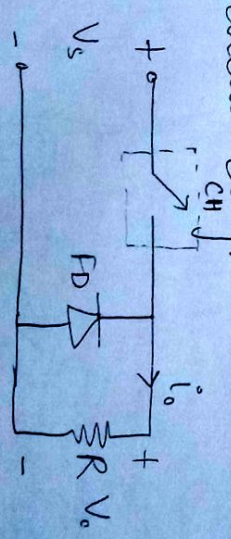
Chopper is a basically static power electronics device which converts fixed DC voltage / power to variable DC voltage or power. It is nothing but a high speed switch which converts AC to DC. The load from source at a high rate to get variable or chopped voltage at the output.



We can control and vary a constant DC voltage with the help of a chopper. Chopper can increase or decrease the DC voltage level at its opposite side. So, it is also known as DC transformer.

1) Step down chopper or Buck converter :-

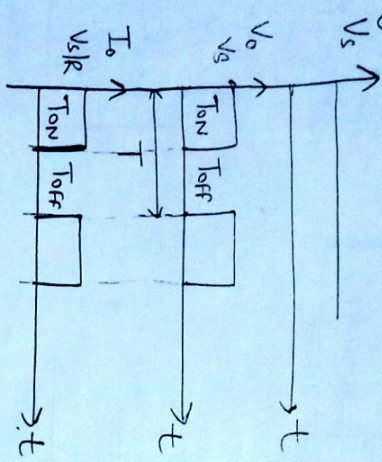
Step down chopper is used to reduce the input voltage level at the output side as shown in the circuit diagram below:-



When CH is turned ON, V_s directly appears across the load i.e. $V_o = V_s$.

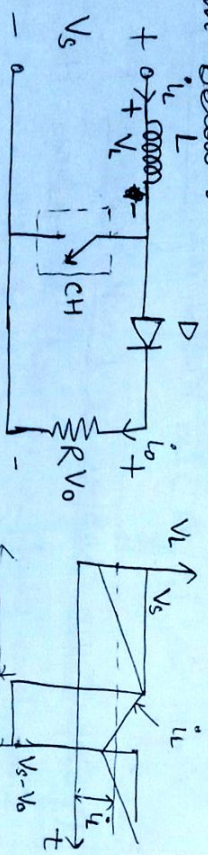
When CH is turned OFF, V_s is disconnected from the load i.e. $V_o > 0$.

The voltage waveform is shown below:-



2) Step up chopper or Boost Converter :-

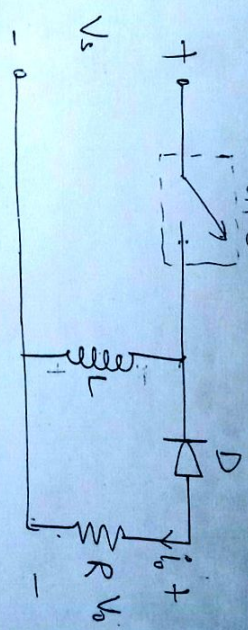
Step up chopper is used to increase the input voltage level of its output side as shown in the circuit diagram below :-



When the switch CH is closed for time duration T_1 , the inductor current rises and the energy is stored in the inductor. If the switch CH is opened for time duration T_2 , the energy stored in the inductor is transferred to the load via the diode D and the inductor current falls.

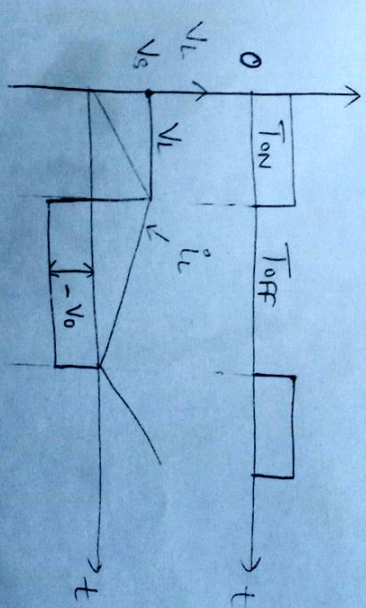
Buck-Boost Converter :-

With the help of Buck-Boost converter, we can increase or decrease the input voltage level at its output side as per our requirement as shown in the circuit diagram below :-



When CH is ON, the source voltage V_s is applied across the inductor L and it will be charged and stored energy in the inductor so, $V_L = V_s$.

When chopper is OFF, inductor L reverses its polarity and discharges through load and diode D. So, $V_o = -V_L$



★ Power Measurement :-

Power is the rate at which work is done.
In a purely resistive circuit, the voltage & current change direction at the same time.
So, the formula is :

$$P = V_{rms} I_{rms}$$

Power meter :-

Power meter is a multi-function device that measures precisely continuous current, alternative current, intensity of DC or AC and power in Watts.

★ Wattmeter :-

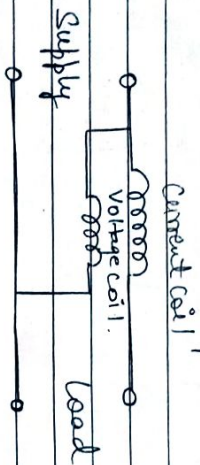
Wattmeter is an instrument used to measure the electrical power. The traditional form of wattmeter is called as dynamometer wattmeter.

The wattmeter has two sets of coil :

1. Current Coil :- One coil is fixed and is made in two identical parts. It is made of heavy gauge copper wire. So, it has low resistance.

Name of Lecturer :

8. Voltage coil :- The other coil which is known as voltage coil is wound from fine gauge wire. So it has relatively high resistance. It is placed between the two parts of the current coil.



→ The current coil is connected in series with the load so that the circuit current flows through it.

→ The voltage coil is connected in parallel with the load.

→ Both the coils will produce magnetic field and these fields interact to each other and produce a deflecting torque on the voltage coil.

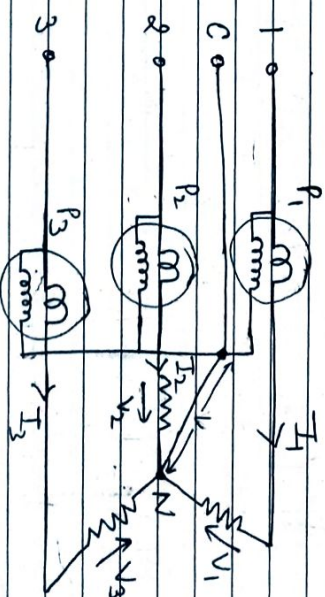
→ The interacting fields are proportional to the circuit voltage & current so the produced deflecting torque is proportional to the product of circuit voltage and current i.e. VI which is nothing but a circuit power.

→ A dynamometer wattmeter reads both ac & dc power.

* Measurement of three phase power :-

Various methods are used for measurement of three-phase power in three phase circuits on the basis of number of wattmeters used.

1. Three wattmeter method :-



The above figure shows a three phase four wire system having a current coil of all the three wattmeters joined as one, two and three connected to respective phase.

Pressure coil of all the three wattmeter are connected to common point at neutral wire. It is clearly seen that each wattmeter will give reading as product of phase current and line voltage which is a phase power.

The resultant sum of all the readings of wattmeter will give the total power of the circuit.

So,

$$P = P_1 + P_2 + P_3$$

$$P = I_1(V_1 - V) + I_2(V_2 - V) + I_3(V_3 - V)$$

$$P = V_1 I_1 + V_2 I_2 + V_3 I_3$$

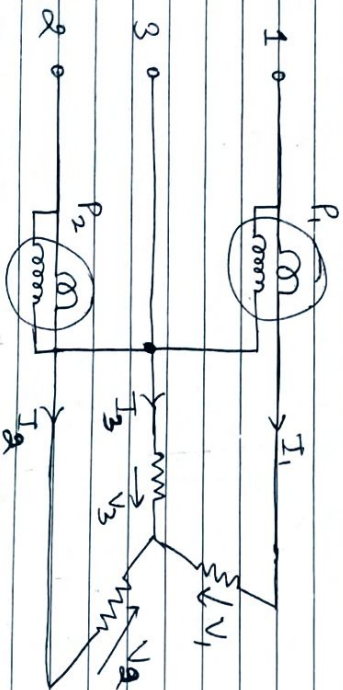
$$\therefore I_1 + I_2 + I_3 = 0$$

2. Measurement of power by two wattmeter method:

In this method, we have two types of connections:-

1. Star connection.
2. Delta connection.

When the load is star connected, the diagram is shown below:-



→ For star connected load, it is clear that the reading of wattmeter one is product of phase current and voltage difference ($V_2 - V_1$).

Similarly, the reading of wattmeter two is the product of phase current and the voltage difference ($V_3 - V_2$).

Thus, the total power of the circuit is the sum of the reading of both the wattmeter.

$$P = P_1 + P_2$$

$$P = I_1(V_1 - V_3) + I_2(V_2 - V_3)$$

$$P = I_1 V_1 - I_1 V_3 + I_2 V_2 - I_2 V_3$$

$$P = I_1 V_1 + I_2 V_2 - V_3 (I_1 + I_2)$$

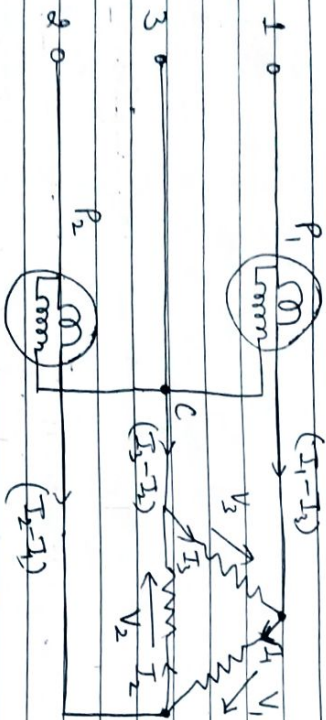
$$P = I_1 V_1 + I_2 V_2 - V_3 (-I_3)$$

$$P = I_1 V_1 + I_2 V_2 + I_3 V_3$$

According to Kirchhoff's law:-
we have, $I_1 + I_2 + I_3 = 0$

$$\text{So, } I_1 + I_2 = -I_3$$

When the load is connected in delta, the diagram is shown below:-



So, the reading of wattmeter one is given by,

$$P_1 = -V_3(I_1 - I_3)$$

and the reading of wattmeter two is given by,

$$P_2 = -V_3(I_2 - I_3)$$

So the total power measured by the two wattmeter is the sum of the reading of the both wattmeter.

$$P = P_1 + P_2$$

$$P = -V_3(I_1 - I_3) - V_3(I_2 - I_3)$$

$$P = -V_3 I_1 + V_3 I_3 - V_3 I_2 + V_3 I_3$$

$$P = -V_3 I_1 - V_3 I_2 + 2V_3 I_3$$

$$P = V_1 I_1 + V_2 I_2 + V_3 I_3$$

$$P = V_1 I_1 + V_2 I_2 + V_3 I_3$$

$$P = V_1 I_1 + V_2 I_2 + V_3 I_3$$

According to Kirchhoff's law:-

$$\begin{cases} V_1 + V_2 + V_3 = 0 \\ \text{or } V_2 + V_3 = -V_1 \end{cases}$$

(3) One wattmeter method:-

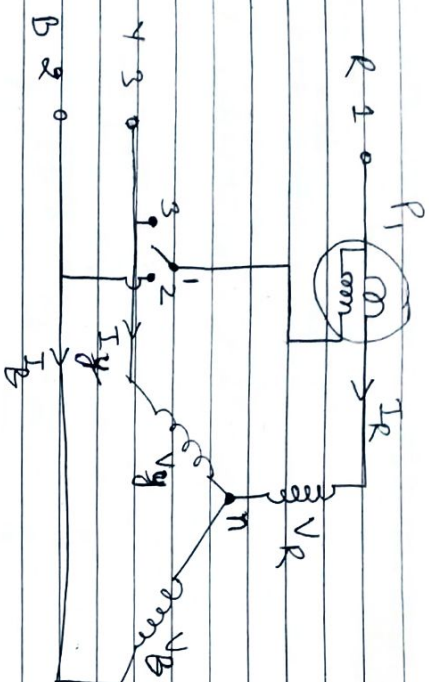
The main limitation of this method is it cannot be applied on unbalanced load.

So, in balanced condition,

$$I_1 = I_2 = I_3 = I$$

$$V_1 = V_2 = V_3 = V$$

As shown in the diagram below, the load is star connected:-



In the diagram, we can see there are two ~~one~~ wattmeters marked as (1-2) and (1-2). By closing the switch (1-3), we get the reading of wattmeter as:

$$P_1 = \sqrt{3} V_L I_L \cos(30^\circ + \phi)$$

$$P_2 = \sqrt{3} V_L I_L \cos(30^\circ - \phi)$$

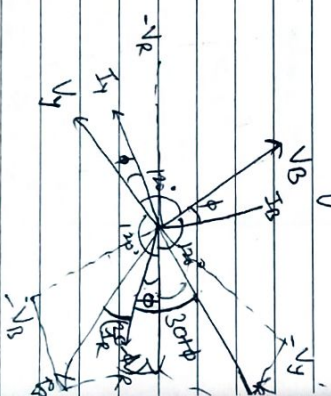
Now by closing the switch (1-2), we get the reading of wattmeter as;

$$P_3 = \sqrt{3} V_L I_L \cos(30^\circ - \phi)$$

$$P_4 = \sqrt{3} V_L I_L \cos(30^\circ + \phi)$$

Total power is given by;

$$P = P_1 + P_2$$



$$P = \sqrt{3} V_L I_L \cos \phi$$

* Energy Consumption Formula :-

Energy consumption is the consumption of energy for power of a system using supply. The consumption is done in Kilograms of oil equivalent per year (Kgoe/a), Giga-watt per year and in watts.

So, the energy ~~consumption~~ formula is given by,

$$E(\text{Kwh/day}) = \frac{P(\text{Kwh}) \times t}{1000}$$

where, E is the energy in kilowatt-hours (Kwh)
P is power in Watts.
t is time in hours per day.

Que. Calculate the energy consumption in a system that consumes 190 watts of power and works for 3 hr a day.

Solution :-

$$\text{Given : } P = 190 \text{ W} \\ t = 3 \text{ hrs/day.}$$

So, The energy consumption is given by :

$$E(\text{KWh/day}) = \frac{P \times t}{1000} \text{ KWh}$$

$$E(\text{KWh/day}) = \frac{190 \times 3}{1000} \text{ KWh.}$$

$$E(\text{KWh/day}) = 0.57 \text{ KWh}$$

* Introduction :-

An electrical installation is a combination of electrical equipment installed to fulfill a specific purpose.

In dealing with the electrical installation, it is necessary to ensure the safety of personnel as well as the ~~prop~~ protection of equipment from the electrical faults.

The most popular methods of protection are:

- (1) Earthing or grounding of equipment.
- (2) Use of fuses or Circuit Breakers (ie MCB & MCCB)
- (3) Use of earth leakage and residual current circuit breakers.

* Switchgear :-

Switchgear is the combination of electrical disconnect switches, fuses or circuit breakers used to control, protect and isolate electrical equipment.

A switchgear or electrical switchgear is a generic term includes all the switching devices associated with mainly power system protection.

High-voltage switchgear was invented at the end of the 19th century for operating motors and other electric machines and now be used with voltage up to 1,100 KV.

Switchgear in substations are located on both the high and low voltage sides of large power transformers.

* Features of switchgear :-

- (1) Complete reliability
- (2) Absolutely certain discrimination.
- (3) Quick operation
- (4) Provision for manual control.

* Classification of switchgear :-

Switchgear can be classified on the basis of voltage level into the following :-

(1) Low Voltage Switchgear :-

The electrical switchgear rated up to 1KV is termed as low-voltage switchgear. This includes low-voltage circuit breakers, switches, HRC fuses, MCCB, ELCB etc.

(2) Medium-Voltage Switchgear :-

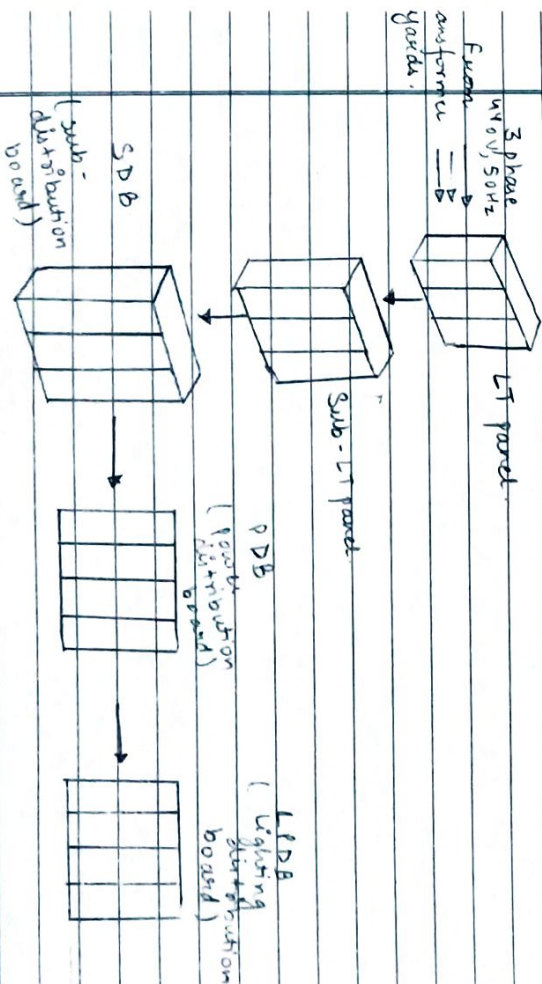
The electrical switchgear rated from 3KV to 36KV

is termed as medium-voltage switchgear. It includes metal enclosed indoor type, metal enclosed outdoor type, oil CB, SF₆ and vacuum CB.

(3) High voltage switchgear :-

The electrical switchgear rated beyond 36KV is termed as high-voltage switchgear. In these circuit breakers special care is taken during designing as the voltage level is very high.

An installation process of a switchgear is shown below :-



Switchgear installation process

* Switch Fuse Unit (SFU) :-

→ SFU is a switching device having one switch and one fuse unit. When we operate the breaker, the contacts will get close through switch and then the supply will pass through the fuse unit to the output.

→ It is made of power fuse rating (80-250-400 A) and a switch which can be operated from outside using extended rotary handle to rotate the shaft which is inside it. ↓

→ SFU is used for protecting power and control circuit depends on the current rating required.

→ It has been used to trip the circuit particularly for high capacity tripping.

→ It is used in industrial, residential and commercial buildings for electrical fittings.

Rating current : 16-32-63-100-200 Amp.

Rated Voltage : 240/415 V AC.

* Miniature Circuit Breaker (MCB) :-

MCB is an automatically operated electrical switch designed to protect the electrical circuit and human beings from electrical shocks & faults.

These days MCB is more commonly used in low voltage electrical network than fuse. MCB is used for rating upto 100 Amps in the motor feeders. In domestic sector, its rating is upto 1000 Amps.

* Working of MCB :-

Basically, MCB is operated under two effects of overcurrent.

- (1) Thermal effect
- (2) Short-circuit condition.

The thermal operation is achieved with a bimetallic strip, whenever continuous over-current flows through MCB, the bimetallic strip is heated and deforms by bending. This deflection of bimetallic strip releases mechanical latch. At this latch is attached with operating lever, it causes to open the MCB contacts.

During short circuit condition, sudden rising of current, causes electromechanical displacement of plunger associated with tripping coil or solenoid of MCB. The plunger strikes the trip lever causing immediate release of latch and consequently opens the MCB contacts.

* Construction of MCB :-

MCB construction is very simple, robust and, maintenance free. Generally a MCB is not repaired or maintained, it just replaced by new one when required. A MCB has normally three main parts

- (1) Frame
- (2) Operating lever
- (3) Trip unit.
- (4) Bimetallic strip
- (5) Fixed and moving contacts.

During the operation of MCB, the electric current carrying path of a MCB is arranged in series like firstly bimetallic strip — latch mechanism. Trip unit = fixed and moving contacts and lastly the operating lever is outside the MCB.

So, whatever may be the condition overheat, may be due to deformation of bi-metallic strip, due to increased or sudden over current by short circuit condition of trip coil or may be due to manual operation, the same latch point is displaced and same spring is released which ultimately responsible for movement of the MCB.

* Advantages of MCB over fuse :-

- (1) It automatically switches off the electrical circuit during abnormal condition but fuse does not sense.
- (2) MCB is much more sensitive to over current than to

- (3) Quick restoration of supply cannot be possible in case of fuse.

- (4) Handling MCB is more electrically safe than fuse.

- (5) In MCB, the switch operating knob comes at its off position during tripping and can be easily identified but in case of fuse, the fuse wire should be checked.

The only disadvantage of MCB over fuse is that this system is more costlier than fuse unit system.

* Moulded Case Circuit Breakers (MCB) :-

MCBs are electromechanical devices which protect a circuit from over-current and short circuit with in the range of 32 amps to 3000 amps.

Their primary functions are to provide a means to manually open a circuit and automatically open a circuit under overload or short circuit conditions.

MCBs can be easily reset after a fault and offers improved operational safety and convenience without involving operating cost.

* Working of MCCB :-

MCCB generally have a

- Thermal element for over current and
- Magnetic element for short circuit release.

Like MCB, the MCCB will operate under thermal effect and short-circuit effect.

The thermal effect is achieved by deformation of bimetallic strip by releasing the spring-loaded trigger. In short-circuit effect, all shut off the electrical circuit.

During short-circuit condition, sudden rising of current causes the displacement of the bimetallic strip and releasing the spring loaded trigger which in turn will shut off the MCCB (breaker).

MCCBs are now available with a variety of releases and these are given below :-

(1) Thermal Magnetic Release :-

It is used bimetallic and electromagnet assemblies to provide over-current protection. For higher level over-currents, instantaneous trip allows MCCBs to interrupt suddenly.

(2) Electronic Release MCCB :-

(Electronic release)

MCCBs uses power electronic circuitry to provide over-current protection. Mainly a current transformer (CT) is used as an MCCB release in this type which act first during the short-circuit condition.

(3) Microprocessor Release MCCB :-

Microprocessor

release MCCBs use microprocessor to provide the over-current protection. It works on monitoring of current with true value of current and give the tripping effect. It is highly flexible and reliable in comparison to other releases.

* Earth Leakage Circuit Breaker (ELCB) :-

Earth leakage circuit breaker (ELCB) is a device used to directly detect currents leaking to earth from an installation and cut the power.

There are two types of ELCBs :-

- (1) Voltage Earth leakage Circuit Breaker (V-ELCB).
- (2) Current Earth leakage Circuit Breaker (C-ELCB).

(1) Voltage Base ELCB :-

Voltage ELCB is a voltage operated circuit breaker. The device will function when the current flows through the ELCB.

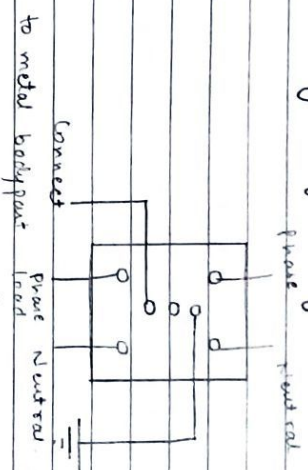
If it contains a relay coil which it being connected to the metallic load body at one end and it is connected to ground wire at the other end.

If the voltage of the equipment body is rises which could cause the difference between earth & load body voltage, as a result the danger of electric shock will occur. This voltage difference will produce an electric current from the load metallic body passes the relay coil and to the earth.

When the voltage on the equipment metallic body rises to the danger level which exceed to 50 V the flowing current through relay coil could move the relay contact by disconnecting the supply current to avoid from any danger electric shock.

The ELCB detects faults from live to the earth within the installation it protects. If sufficient voltage appears across the ELCB's sense coil, it will switch off the power and remain off until manually reset.

The wiring diagram of V-ELCB is shown below:-



→ The way to identify an V-ELCB is by looking for green and yellow earth wires entering the device.

The ELCB is the brain for the shock protection and the grounding is the backbone. Therefore, without a functional grounding there is no protection against electrical shocks in your house even if you have installed the ELCB.

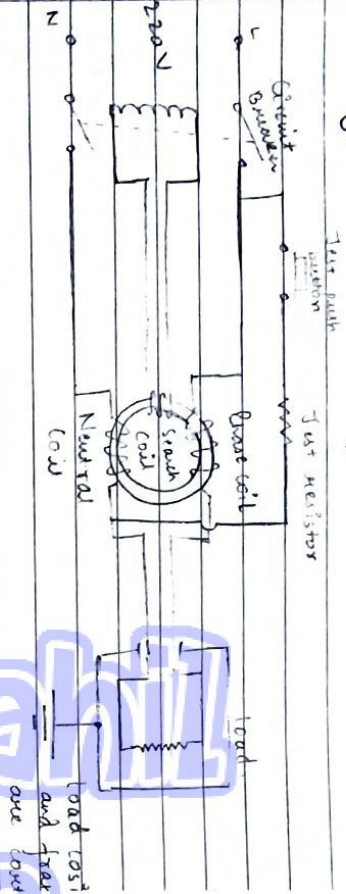
(2) Current-operated ELCB (RCD) :-

Current-operated ELCBs are generally known as Residual-current devices (RCDs). They provide protection against earth leakage. Both the in-rush currents due to run through a sensing coil, any imbalance of the currents means the magnetic field does not perfectly cancel. The device detects

the imbalance and trip the contact.

* Working of RCD :-

The typical circuit of RCD is shown below as



RCD circuit.

→ The supply coil, the neutral coil and the search coil are wound on a common transformer core.

During the normal operation, the same current pass through the phase coil, the load and return back through the neutral coil. Both the phase and neutral coils are wound in such a way that they will produce an opposing magnetic flux. With the same current passing through both coils, their magnetic effect will cancel but each other under a normal condition.

During the faulty operation, or a leakage to earth in the load circuit, or anywhere between the load circuit and the output connection of the RCD circuit

The current returning through the neutral coil has been reduced. Then the magnetic flux inside the transformer core is not balanced anymore. So, the total sum of the opposing magnetic flux is no longer zero. This net remaining flux is what we call a residual flux.

The periodically changing residual flux inside the transformer core crosses with the winding of the search coil. This action produces an EMF and across the search coil which in turn produces a current inside the winding of the trip circuit. This current operates the trip coil of the circuit breaker.

Sensitivity of RCB :-

→ High sensitivity :- 6-10-30 mA (for direct contact life injury protection)

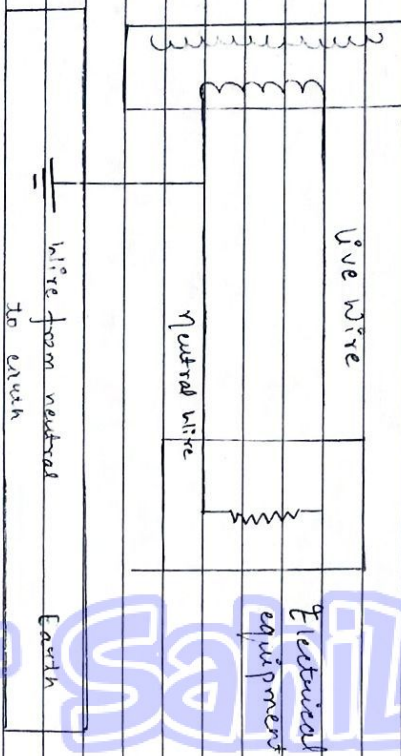
→ Medium sensitivity :- 100-300-500-1000 mA (for fire protection)

→ Low sensitivity :- 3-10-30 A (for protection of machine)

* Earthing System / Grounding System :-

In an electrical installation, an earthing system or grounding system connects specific parts of the installation with the earth's conductive surface for safety and functional purposes.

The diagram of earthing system is shown below:



→ To connect the metallic parts of an electric appliance to the earth is called earthing.

→ Or, in other words, to connect the metallic parts of electric machinery and devices to the earth plate through a thick conductor wire of low resistance is known as earthing.

→ The process of transferring the immediate discharge of the electrical energy directly to the earth by the help of the low resistance wire is known as earthing.

Mostly, the galvanised iron is used for the earthing. Earthing provides the simple path to the leakage current.

* Types of earthing :-

The electrical equipment mainly consists of a non-metallic parts or the body of the equipment which does not carry current and a neutral wire or system of the equipment. So, the earthing is classified according to these two parts of the electrical equipment and they are :-

(1) Neutral earthing :-

In neutral earthing, the neutral of the system is directly connected to the earth by the help of the GI wire. Such type of earthing is mostly provided to the system which has the star winding for example, generators, transformers, motors etc.

(2) Equipment earthing :-

In equipment earthing, the electrical equipment body is directly connected to the earth by a conducting wire having low resistance. If any fault occurs in the apparatus the short-circuit current will pass through the conducting wire to the earth and thus protect

the system from damage and shock.

* Importance of earthing :-

The earthing is essential because of the following reasons :-

- Protects the personnel from the short circuit current
- Provides the easiest path to the flow of short circuit current.
- Protects the machinery and personnel from the high voltage surges and lightning discharge.

* Difference between earthing and grounding

	<u>Grounding</u>	<u>Earthing</u>
Definition:	The current carrying part is connected to the ground.	The body of the equipment connected to the ground
Location:	Between neutral & Earth.	Between equipment body & ea
Symbol:		
Potential:	It does not have zero potential	It has zero potential.
Applications:	Provide the return path to current	Discharges the electrical energy to the earth.
Examples:	Neutral of Generator and Power transformers. Connected to the ground.	The enclosure of the apparatus connected to the ground